
Using Simulation to Estimate MOSFET Junction Temperature in a Circuit Application

Presented by:

David Divins

Senior Staff Field Applications Engineer

International Rectifier

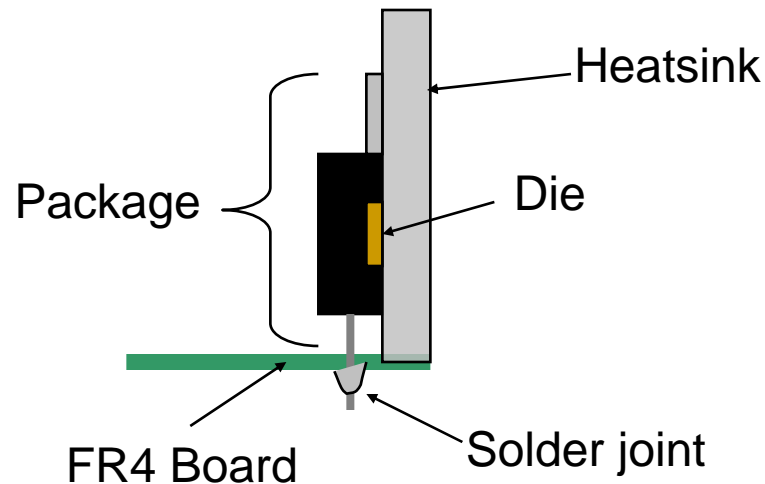
ddivins1@irf.com

Agenda

- Definition of Electro-Thermal Simulation
 - Simulation Tools and Methods
 - Methods of Estimating Die Temperature
 - Creating Quasi-Dynamic MOSFET Model
 - Model Generation
 - Example Application
 - Conclusion
-

Electro-Thermal Simulation

- Purpose of Electro-Thermal Simulation is to predict MOSFET junction for a given application.



Electro-Thermal Simulation

■ Applications

- ❑ Solenoid drivers
- ❑ Motor drive
- ❑ Lighting ballast
- ❑ DC/DC converters
- ❑ Switch model power supplies
- ❑ Class D amplifiers



Simulation Tools and Methods

■ Method

- ❑ Implementing model in the hardware description language
- ❑ Implementing model using equations and macro modeling

```
library ieee; use ieee.std_logic_1164.all;
use ieee.electrical_systems.all;
```

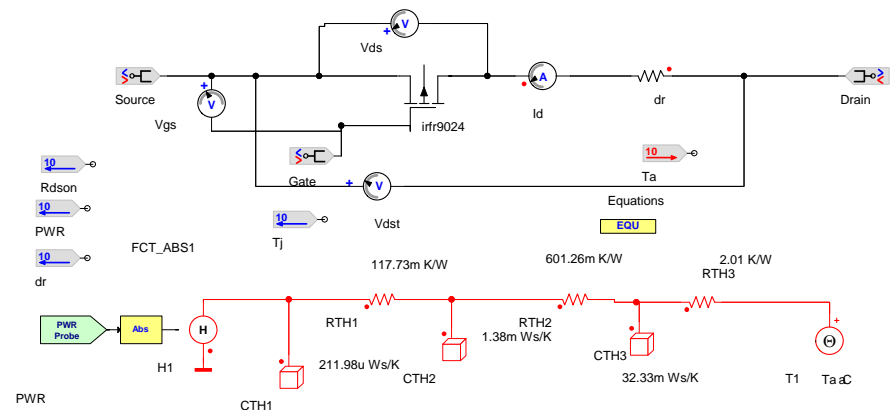
```
entity comparator is
port ( terminal a : electrical;
      signal d : out std_ulogic );
end entity comparator;
```

```
architecture ideal of comparator is
constant ref_voltage : real := 5.0;
quantity vin across a;
begin
```

```
comparator_behavior : process is
begin
if vin > ref_voltage / 2.0 then
d <= '1' after 5 ns;
else
d <= '0' after 5 ns;
end if;
wait on vin'above(ref_voltage / 2.0);
end process comparator_behavior;
```

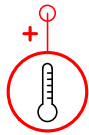
```
end architecture ideal;
```

VS.



Simulation Tools and Methods

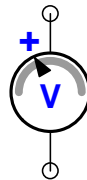
- Quasi-Dynamic MOSFET model implementation
 - Macro modeling with use of linking equations
 - Using a multi domain simulator that allows for Electro-Thermal simulation.
 - Volts and Amps
 - Heat flow (Watts) and temperature



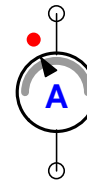
Thermometer



Heat



Voltmeter



Ammeter

Methods of Estimating Die Temperature

- Methods of estimating MOSFET die junction temperature
 - Equation based + Thermal Impedance curve

$$P = I * V * D$$

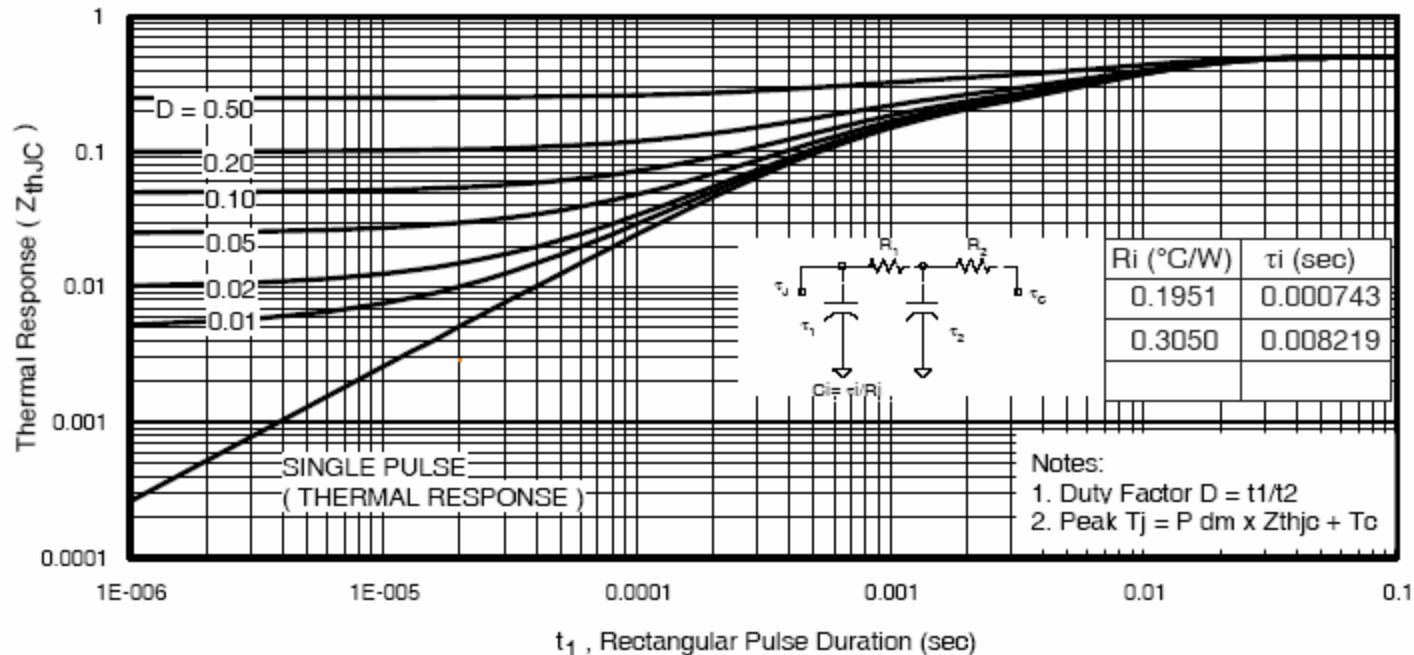
Where:

I = average current during the conduction cycle

V = equivalent voltage across the device during the conduction cycle

D = duty cycle

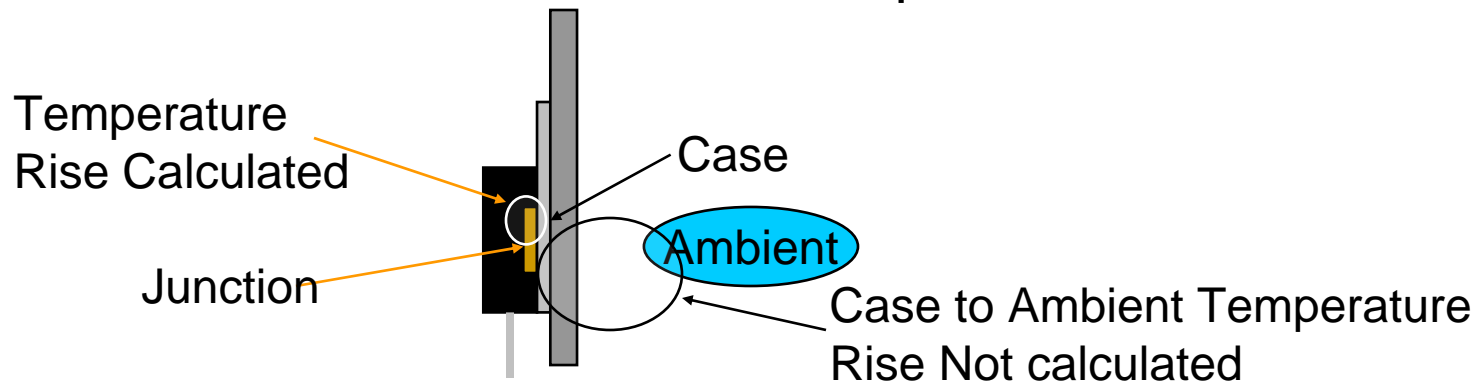
Methods of Estimating Die Temperature



- Use Power calculated with $P=I*V*D$
- Use pulse width and duty cycle to determine Z_{th} (thermal impedance) from device thermal impedance curve
- Temperature rise ($\Delta T_{junction}$)= $Z_{th} * P$

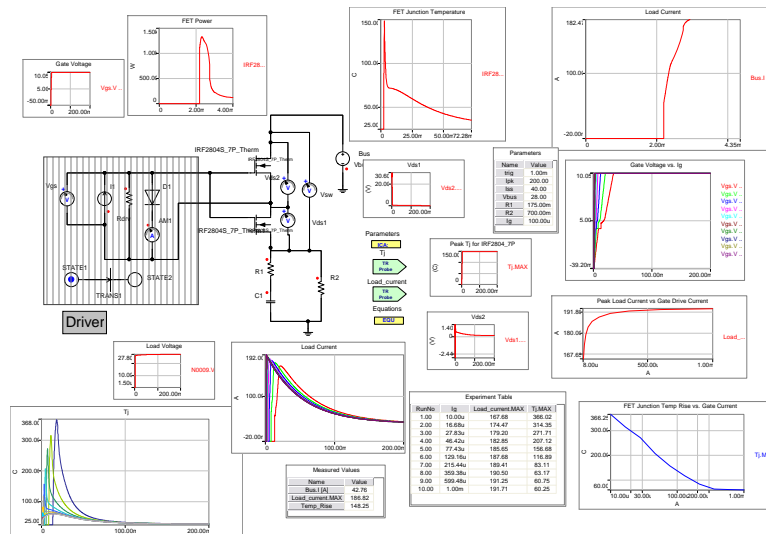
Methods of Estimating Die Temperature

- Limitations of the equation based junction temperature estimate
 - Only temperature rise from junction to case is taken into account. Neglects case to ambient temperature rise.
 - Assumes the power pulse is an ideal square edged pulse train.
 - It does not allow for transient thermal response.



Methods of Estimating Die Temperature

- Simulator based MOSFET junction temperature estimate
 - Uses circuit simulation to calculate junction temperature in an application
 - The circuit can be arbitrary
 - Transient thermal response is calculated
 - Component parameters change with temperature



Methods of Estimating Die Temperature

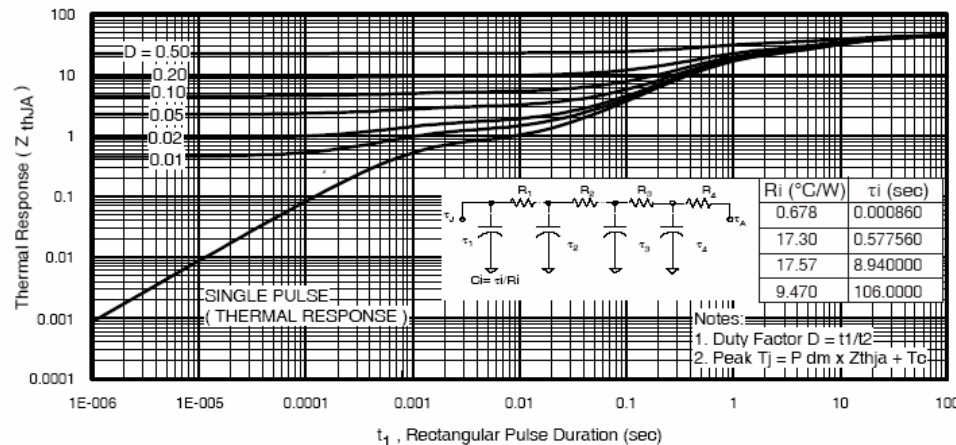
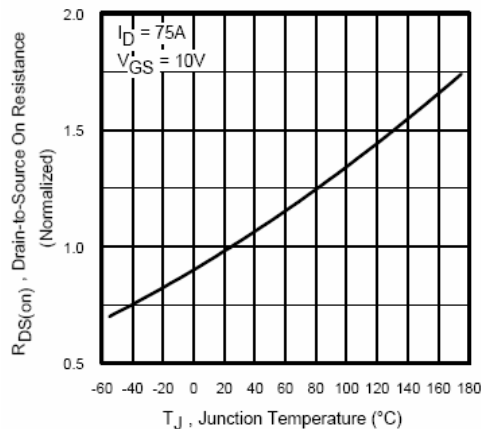
- Assumptions made for junction temperature estimates using simulation
 - No other source of heat considered (Temperature rise due to self heating only)
 - Only MOSFET R_{dson} and threshold voltage changes with temperature
 - Since simulation solves Ordinary Differential the junction is assumed to be a point source of heat.
-

Creating Quasi-Dynamic Thermal MOSFET Model

■ Gathering information:

- 25C Spice Model of MOSFET
- Datasheet information
 - Rdson vs. Temperature curve
 - Thermal Impedance Curve with thermal RC ladder network

```
.SUBCKT irf1404 1 2 3 * SPICE3 MODEL WITH
THERMAL RC NETWORK
*****
***** * Model Generated
by MODPEX * *Copyright(c) Symmetry Design
Systems* * All Rights Reserved * * UNPUBLISHED
LICENSED SOFTWARE * * Contains Proprietary
Information * * Which is The Property of * *
SYMMETRY OR ITS LICENSORS * *Commercial Use
or Resale Restricted * * by Symmetry License
Agreement * ***** * Model
generated on April 2, 01 * MODEL FORMAT: SPICE3
* Symmetry POWER MOS Model (Version 1.0) *
External Node Designations * Node 1 -> Drain * Node
2 -> Gate * Node 3 -> Source M1 9 7 8 8 MM L=100u
W=100u .MODEL MM NMOS LEVEL=1 IS=1e-32
+CGSO=7.17952e-05 CGDO=1.60578e-08 RS 8 3
0.00282867 D1 3 1 MD .MODEL MD D IS=1.89845e-
10 RS=0.00218742 N=1.20398 BV=40 +IBV=0.00025
EG-1 2 YTI=1 85712 TT=2 00014e-05
```

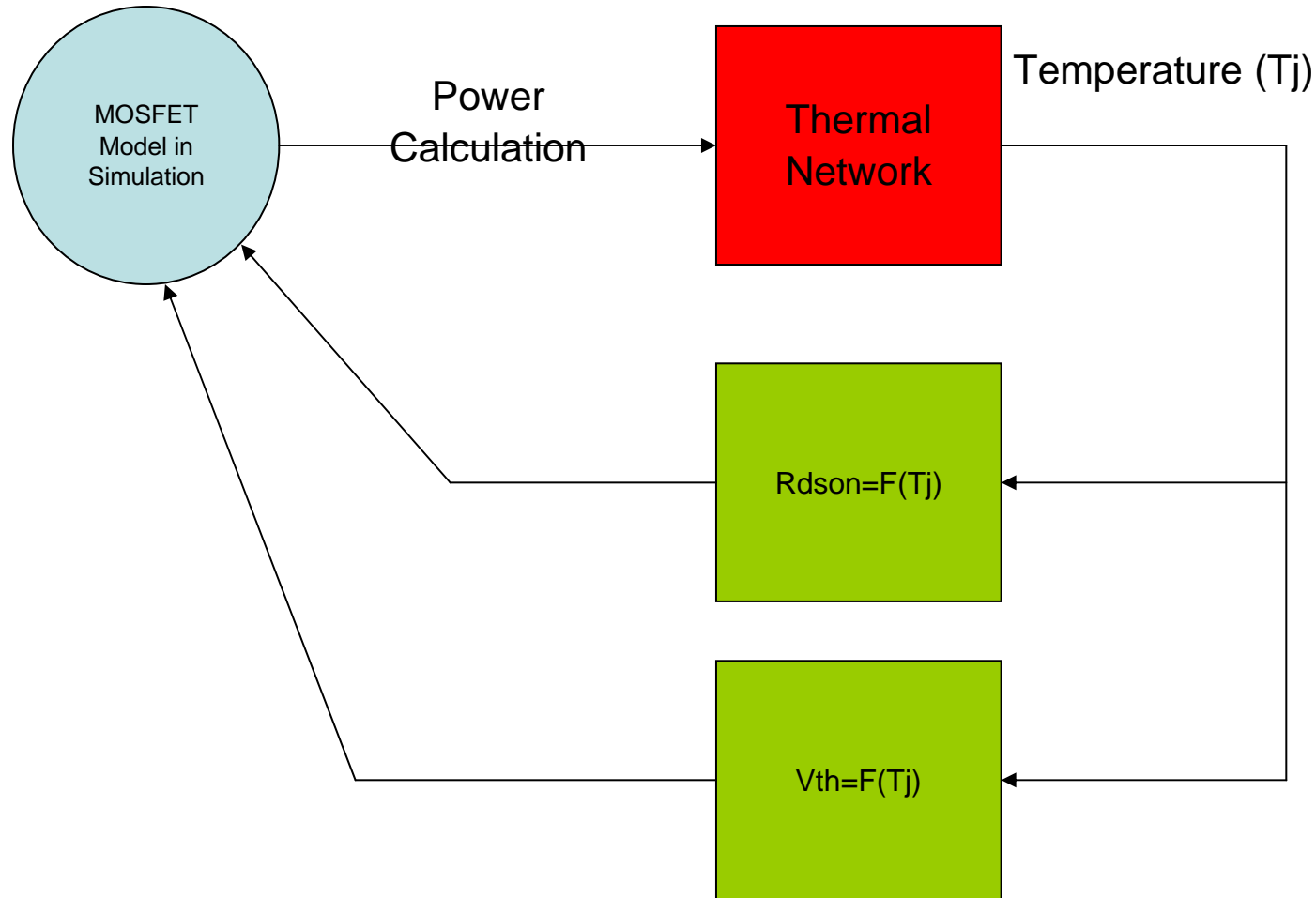


```
FC=0.1
I16781
S=0
3e-09
ID2 *
0 TT=0
) IS=1e-
VF12 4
I1 7 9
*
0 TT=0
MD3 D
```

Creating Quasi-Dynamic Thermal MOSFET Model

- 25C Spice Model
 - Characterized to the datasheet
 - Does not change performance characteristics as power is calculated
 - Used as base model for Quasi-Dynamic MOSFET model
-

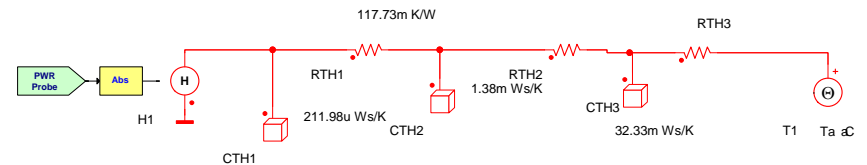
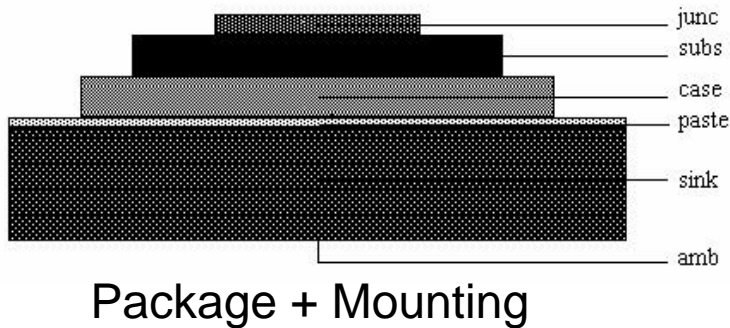
Creating Quasi-Dynamic Thermal MOSFET Model



Model Generation

■ Ladder Network

- A thermal RC network used to model the dynamic thermal behavior of the package + mounting system.



Equivalent RC Ladder Network

The ladder network can be synthesized from the thermal impedance curve or is given by the MOSFET manufacturer

Model Generation

- Tying the thermal model to the 25C Spice model
 - Create the equation that represents Rdson vs. temperature

$$R_{dson}(T_j) = R_{dson}(25C) * [a * T_j^2 + b * T_j + c]$$



$$\frac{dR_{dson}(T_j)}{dT_j} = R_{dson}(25C) * (2 * a * T_j + b)$$



$$dR_{dson} = R_{dson}(25C) * (2 * a * T_j + b) * (T_j - 25)$$

← This expression gets implemented in the model

Note: a, b and c are calculated via a curve fitting routine.
The Rdson vs Temperature curve is assumed to be quadratic.

Model Generation

- Create the voltage source that represents the temperature dependence of V_{th} (threshold voltage)

$$V_{th}(T_j) = -0.007 * (T_j - 25) \longrightarrow \text{Expression used in model.}$$

The voltage source is in series with the MOSFETs gate.

Model Generation

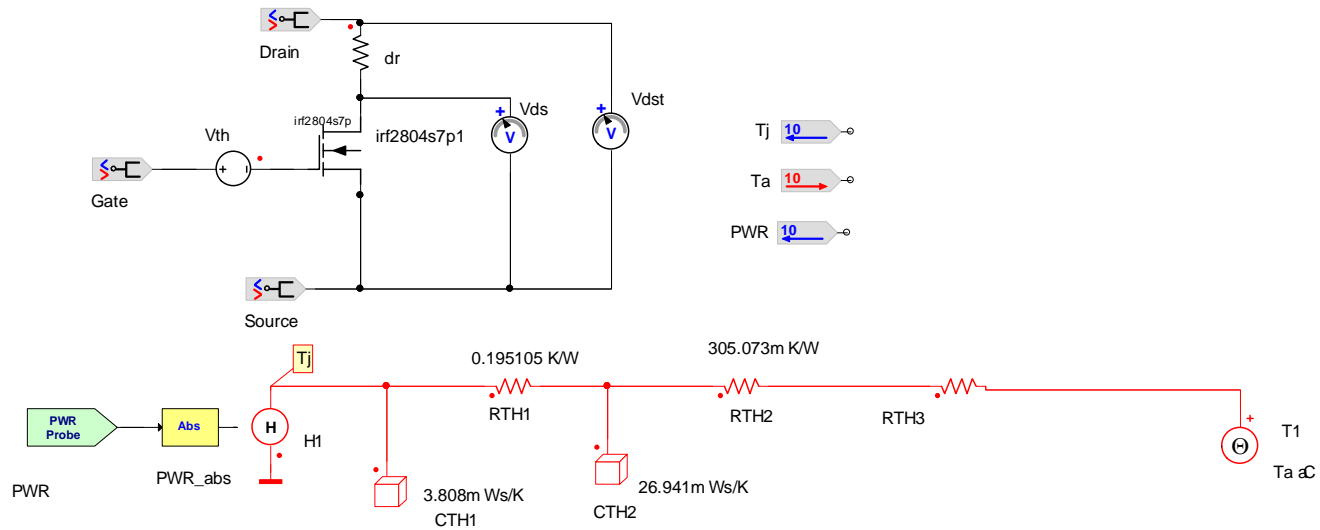
- Calculating the power in the MOSFET for use in the thermal network.

$$P = I_d * V_{ds}$$

This calculated power is the source for the thermal network.

Model Generation

■ Putting it together



Equations

EQU

$$dt := Tj.T - Ta$$

$$Tj := Tj.T - 273.15$$

$$\text{if}(Vds.V < 0.1) \{Rdson25 := \text{abs}(Vds.V / dr.I)\} \text{ else } \{dr := 1m\}$$

$$\text{if}(Vds.V < 0.1) \{dr := (7.41u * Tj + 3.519m) * dt * Rdson25\} \text{ else } \{dr := 1u\}$$

$$PWR := PWR_abs.VAL$$

$$Vth := -7m * (Tj - 25)$$

Model Generation

- Final model
 - 25C Spice model
 - Added voltage source V_{th} in gate implements $V_{th}(T_j)$
 - dR_{dson} implements $dR_{dson} = R_{dson}(25C) * (2 * a * T_j + b) * (T_j - 25)$
 - V_{ds} and the current in dR are used to calculate $R_{dson25C}$
 - V_{dst} and the current in dR is used to calculate the total power
 - PWR_abs is used to insure that the thermal network is driven with positive power.
-

Example Application

- High side switch
 - MOSFET being driven by a opto isolated drive
 - Very low drive current capability
 - Load is capacitive
 - Issue: How does driving this load effect the junction temperature of the MOSFET
-

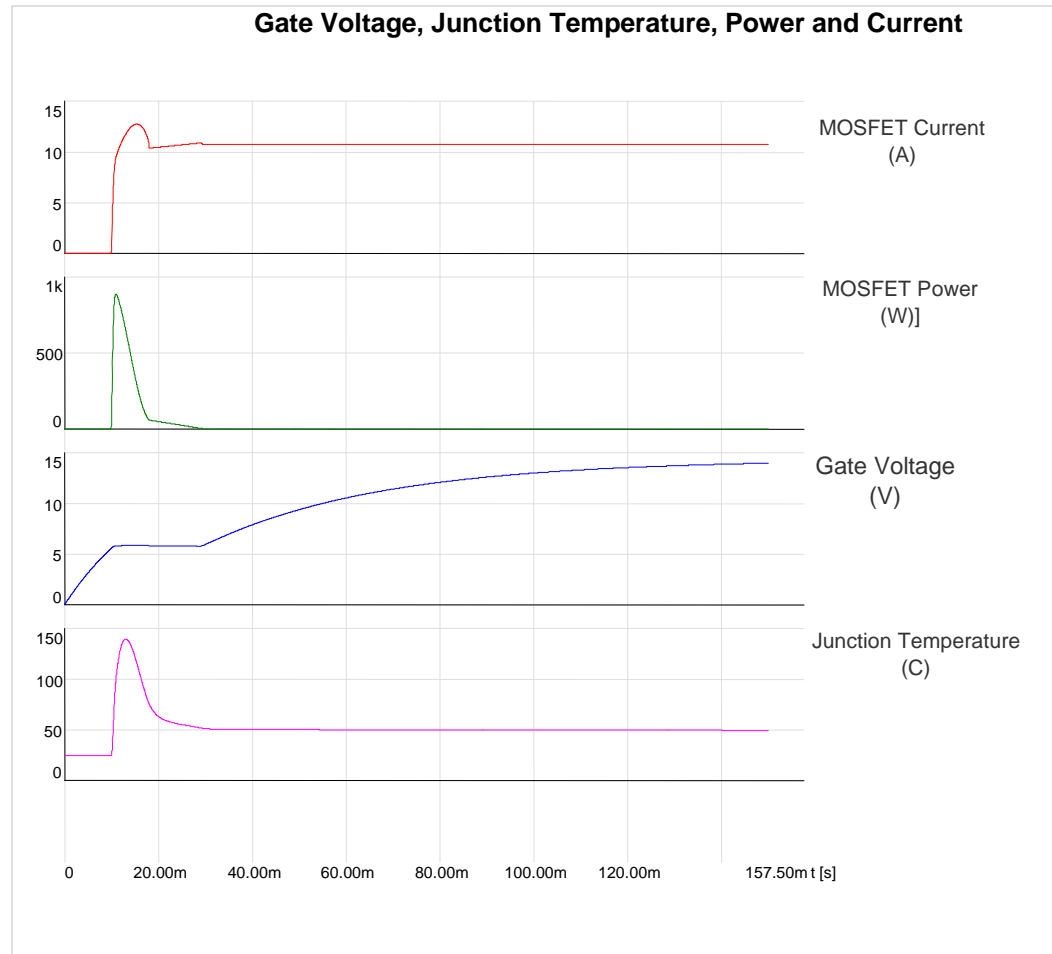
Example Application

■ Assumptions

- ❑ $T_{ambient}=25C$
 - ❑ Heat sink is modeled as just a thermal resistor
 - ❑ C1 & R2 represent a load system i.e. power supply
 - ❑ I_g , V_{gs} , PWR_FET , States 1 & 2, Trans1 and S1 are measurements, input stimulus and ideal switch
-

Example Application

■ Results



Conclusion

- Electro-Thermal simulation allows for analysis in both electrical and thermal domains
 - Quasi-Dynamic Thermal MOSFET model allows for self-heating to alter R_{dson} and V_{th} during simulation as a function of temperature
 - Quasi-Dynamic Thermal MOSFET Model generation is a data gathering task
 - The example shows why it is difficult to switch a capacitive load with an opto-driver and a MOSFET due to the excessive junction temperature spike during turn-on.
-