

# Topic 122

## Reliability Challenges in Integrated High Voltage Devices

Peter Moens  
AMI Semiconductor  
Oudenaarde, Belgium  
and  
Geert Van den bosch  
IMEC  
Leuven, Belgium

Integrated medium-to-high voltage transistors with a breakdown voltage between 20 and 100V, are widely used in automotive, industrial and even medical applications. These transistors are used as an interface between the sub-micron digital CMOS and the analog outside world. Often, the transistors have to switch relatively large currents (1-10A), and as such power dissipation is a prime concern.

In this tutorial, the challenges in device reliability for integrated smart power devices are highlighted. An overview of the applications and the resulting quality and reliability requirements is given. The most common types of devices are presented and discussed (DeMOS, LDMOS, VDMOS, IGBT, ...). Their typical device reliability problems are highlighted. These are, amongst others : DC hot carrier, robustness under ESD-like events, switching of different types of loads, operation and failure upon single and multiple power pulsing, thermal modeling, etc... The different measurement techniques used to assess the complete reliability behavior of integrated power devices, are discussed

## **Peter Moens**

Dr. Peter Moens received a M.S. and a Ph.D. in solid state physics from the University of Gent, Belgium, in 1990 and 1993 respectively. From 1993 till 1996, he worked as a post-doctoral fellow in collaboration with Agfa-Gevaert, Mortsel Belgium on the electron capture efficiency of silver halide emulsions. In 1996, he joined AMI Semiconductor, Oudenaarde, Belgium where he is involved in the development of smart power technologies and devices. His present activities are focusing on advanced device concepts and the reliability of integrated power devices. He is author or co-author of over 60 papers in international journals and proceedings and issued several patents. He is vice-chairman of the HV reliability sub-committee of IRPS.

## **Geert Van den bosch**

Dr. Geert Van den bosch received the M. Sc. in electrical and mechanical engineering in 1987 and the Ph. D. in applied sciences in 1993, both from the Katholieke Universiteit Leuven, Belgium.

In 1987, he joined the Interuniversity Microelectronics Center (IMEC) in Leuven, where he did research on basic hot-carrier degradation effects, semiconductor device physics and electrical characterization techniques. From 1993 to 1999 he has been active in the development of several generations of deep submicron mixed-signal CMOS technologies. Since 1999, he has been working as a Senior Reliability Researcher in fields such as plasma and process induced damage, hot-carrier degradation, ultrathin gate oxide integrity, Cu-low-k dielectric back end reliability, and the reliability of high voltage devices and smart power technology.

Dr. Van den bosch has served as a technical program committee member of the Plasma and Process Induced Damage (P2ID) conference, the International Conference on IC Design and Technology (ICICDT) and the International Reliability Physics Symposium (IRPS).

# ***Reliability Challenges in Integrated High Voltage Devices***

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**Peter Moens**

**AMI Semiconductor BVBA, Belgium**

**Geert Van den bosch**

**IMEC, Belgium**

- ***Introduction***
- ***Q&R Specifications***
- ***Device Overview***
- ***Reliability Measurements and Analysis Techniques***
- ***Reliability Problems in Smart Power Technologies***
- ***Conclusions***

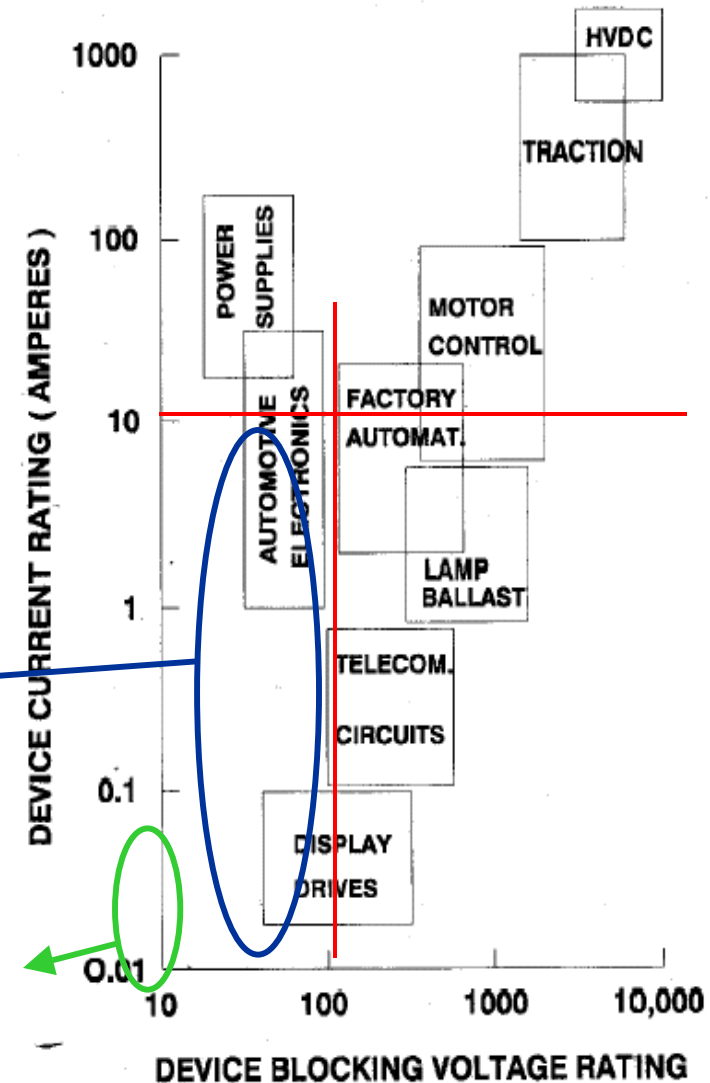
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# Introduction

- **Voltage : ~100 V**
- **Current : ~10 A**
- **Smart : integrated CMOS processing**
- **Trend from discretetes to ASICs**
- **Interface with the “real” world**

Smart HV & Power technologies

CMOS technologies

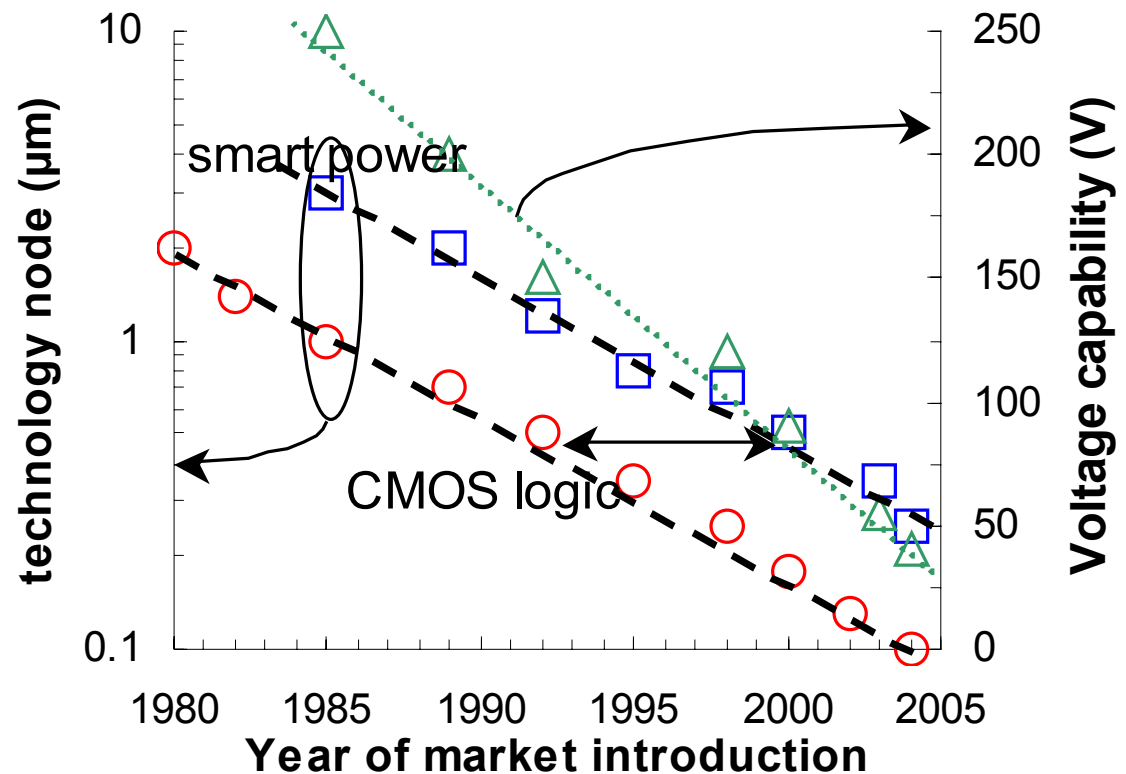


- ***Automotive***
- ***Peripherals (printers, ...)***
- ***Industrial (washing machines, ...)***
- ***Battery management***
- ***Drivers (xDSL, displays)***
- ***Medical (implantables...)***
- ***Consumer (home cinema...)***

***See refs [1-5]***

# Introduction

- **Smart Power roadmap follows ITRS, but with a delay of ~3 technology nodes.**
- **$V_{bd}$  drops**
- **Is there a need to go to lower techno dimensions for HV/power ??**



- **Smart power challenges**

- ◆ Devices integrated in sub- $\mu\text{m}$  CMOS technos :

- Low cost (limited number of extra masks, process steps, use as much as possible the standard CMOS flow)
- Small area ( $R_{on}$ ) for a high  $V_{bd}$ .
- Low metal resistance.

- ◆ Flash, EEPROM and OTP

- ◆ Heat generation: must be kept under control and has to be accurately modeled.

- ◆ Isolation and latch-up

- ◆ Reliability

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# ***Q&R Specifications at product level***

- ***Dedicated specifications and criteria at product/technology level (automotive)***
  - ◆ All standard CMOS qualifications +
    - AEC-Q100 (Stress Test Qualification for Integrated Circuits)
    - AEC-Q101 (Stress Test Qualification for Discrete Semiconductors)
  - ◆ ISO 7637 : immunity to electrical disturbance (Schaffner pulses)
  - ◆ ESD : 4 kV HBM, 750V CDM, 400V MM
  - ◆ IEC 61000-4-2 : system ESD (8 kV contact discharge ; 15 kV air discharge)
  - ◆ High temperature ( $T_{junc} \geq 175 \text{ }^{\circ}\text{C}$ )

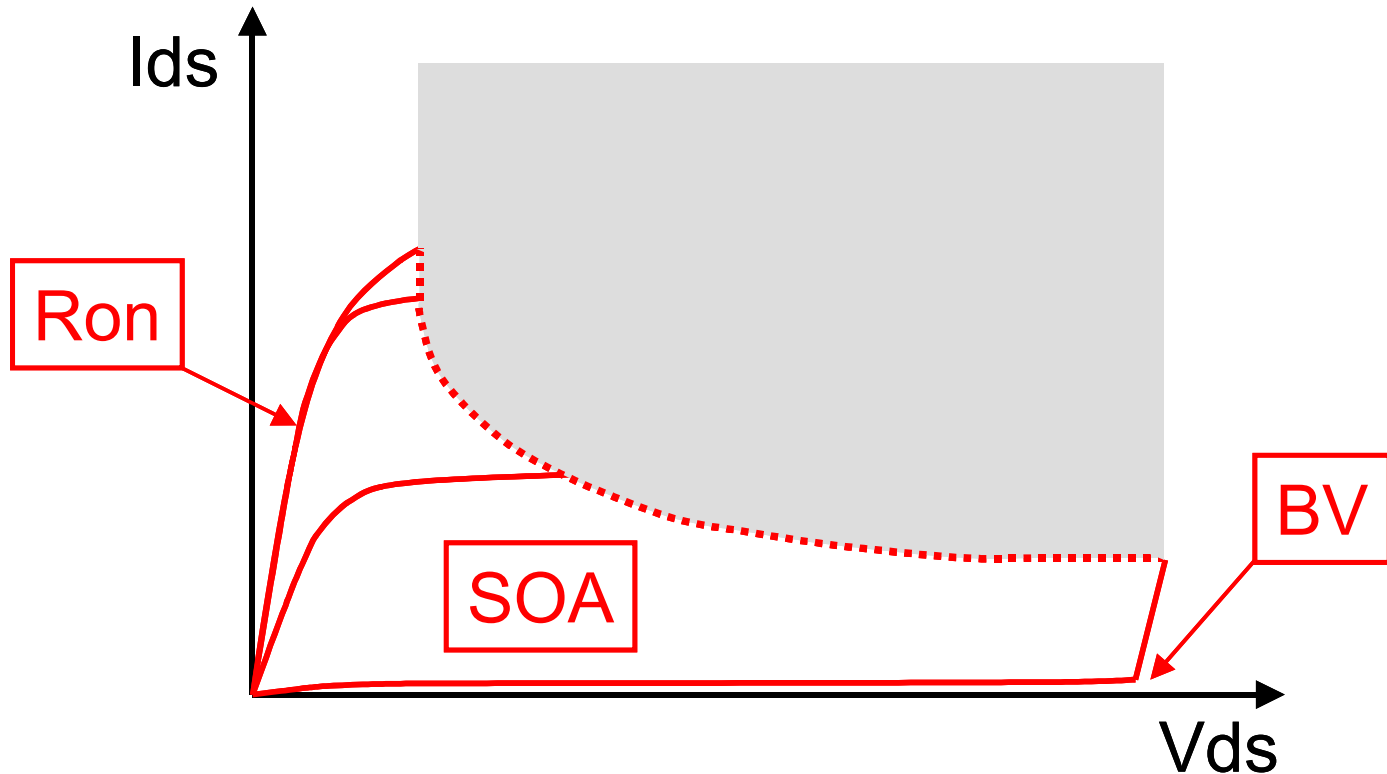
# Q&R Specifications at product level

- *High temperature applications [4,5]*

Mounting zone	Module description	Temperature zones and % of total operation time in each temperature zone			
		T1 (5%)	T2 (20%)	T3 (65%)	T4 (10%)
Engine Compartment	Temperate zone, thermally isolated	- 40 °C	25 °C	60 °C	85 °C
	Splash wall	- 40 °C	25 °C	90 °C	140 °C
	Attached to the engine or attached to the gearbox	- 40 °C	25 °C	95 °C	150 °C
	Throttle valve, close to the exhaust	- 40 °C	25 °C	120 °C	205 °C
Chassis	Locations exposed to heat sources	- 40 °C	25 °C	90 °C	120 °C
	Near breaks or hydraulics	- 40 °C	25 °C	105 °C	175 °C
Cabin	Dashboard, hat rack	- 40 °C	25 °C	60 °C	110 °C
	Roof under strong sun exposure	- 40 °C	25 °C	90 °C	115 °C

# Q&R Specifications at device level

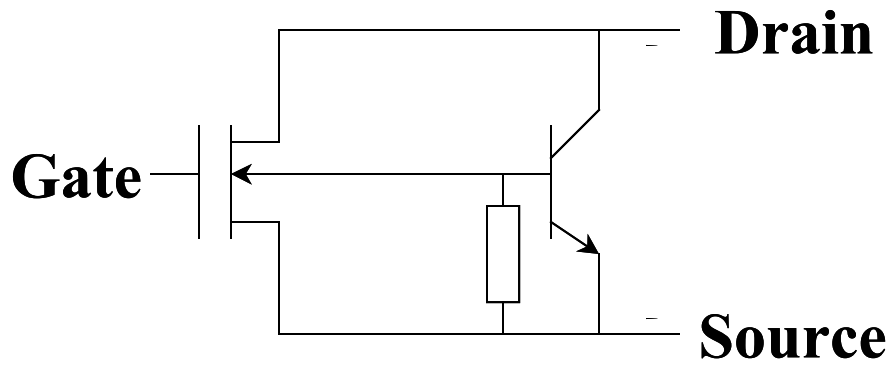
- $R_{on}$  and  $V_{bd}$  are prime device parameters



- Trade-off between  $R_{on}$ , SOA and  $V_{bd}$

# Q&R Specifications at device level

- **Every n(D)MOS has a parasitic NPN**



$$n_{ie}(T) \sim T^{3/2} \exp(-E_g / 2kT)$$

$$V_{bi}(T) \sim \frac{kT}{q} \ln\left(\frac{N_B}{n_{ie}(T)}\right)$$

$$V_{bi}(T) = I_h(T) \cdot R_{base}(T)$$

- **Bipolar can be triggered :**
  - **Electrical : hole current  $I_h$  and/or base resistor  $R_B$  are large enough to forward-bias the E-B junction**
  - **Thermal :  $T$  is high enough such that  $n_{ie} \sim N_B \rightarrow$  E/B junction is wiped out ( $T=T_{crit}(N_B)$ , typically  $T_{crit} \sim 700$  K).**

# Q&R Specifications at device level

- **long-term (s...yrs): limited by hot-carrier degradation [6-21]. Continuous operation e.g. current source.**
- **short-term (ns... $\mu$ s): limited by electrical bipolar snapback [22-30]. (system) ESD event.**
- **medium-term ( $\mu$ s...ms): limited by electro-thermal breakdown [31-41]. Switching of a load.**

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# Device Overview

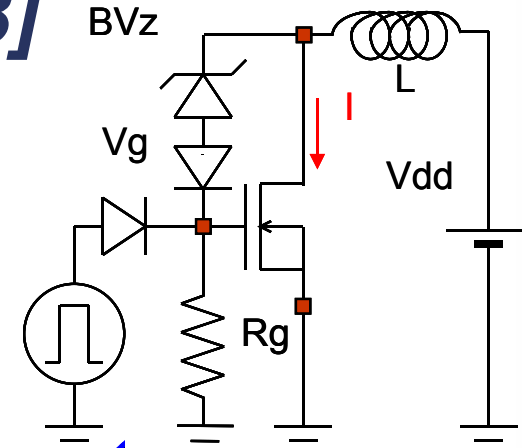
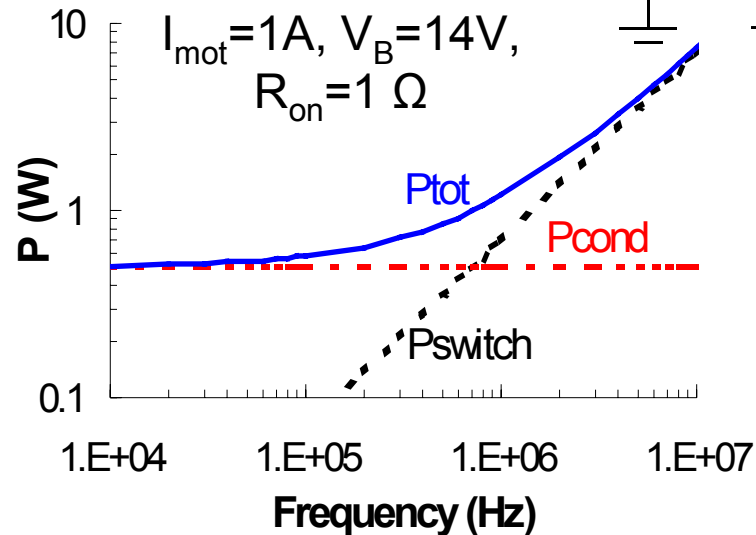
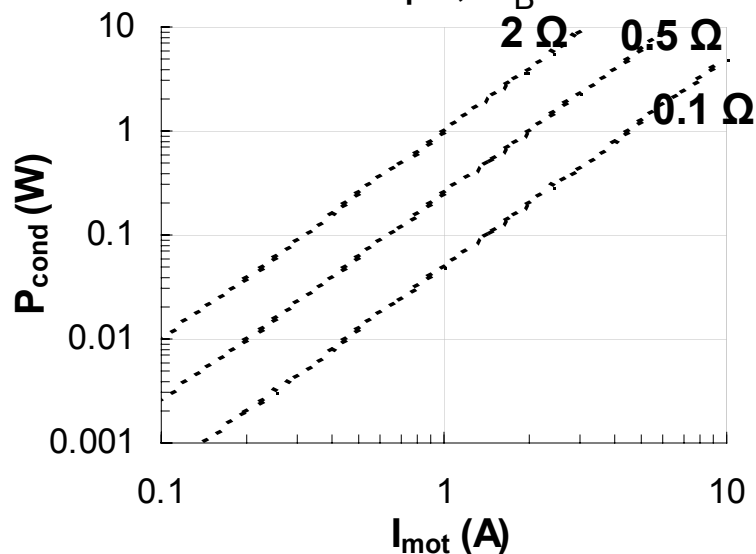
- **Figures-of-Merit :  $R_{on}$ ,  $V_{bd}$  and  $Q_{gd}$**
- **Drain extended MOS (DeMOS)**
- **Lateral DMOS (LDMOS)**
- **(quasi)-Vertical DMOS (VDMOS)**
- **Silicon limit and benchmarking**
- **Lateral IGBT (L-IGBT)**
- **SCR's and Thyristors**
- **Trenchmos**
- **...**

# Conduction and switching loss

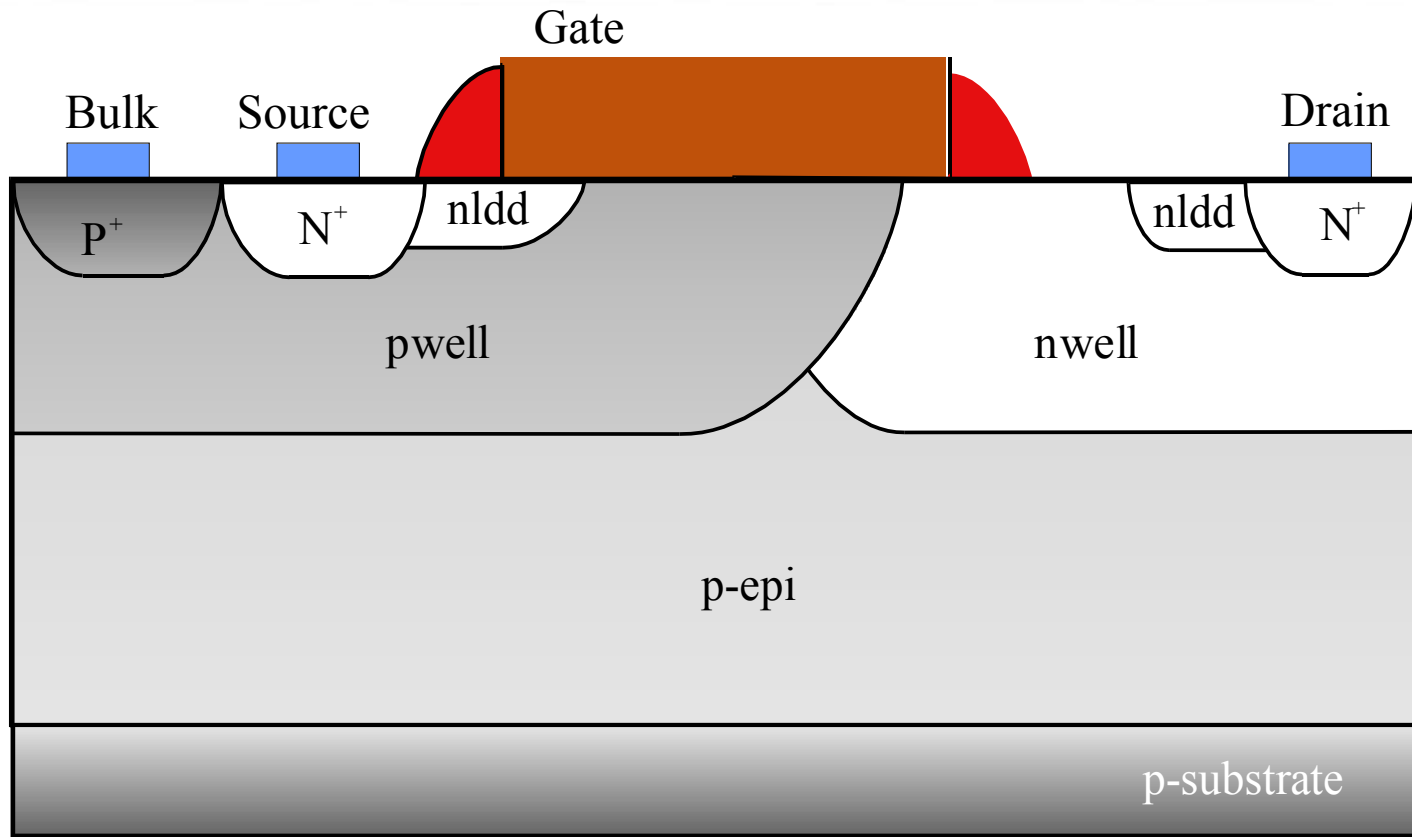
- $R_{on}$  of a switch must be as low as possible  
 $\rightarrow$  reduce power dissipation [2,3]

$$P_{tot} = \frac{t_{on}}{t} \cdot I_{mot}^2 \cdot R_{on} + \frac{t_r + t_f}{2t} \cdot V_B \cdot I_{mot}$$

$t = 40 \mu\text{s}$ ,  $V_B = 14\text{V}$

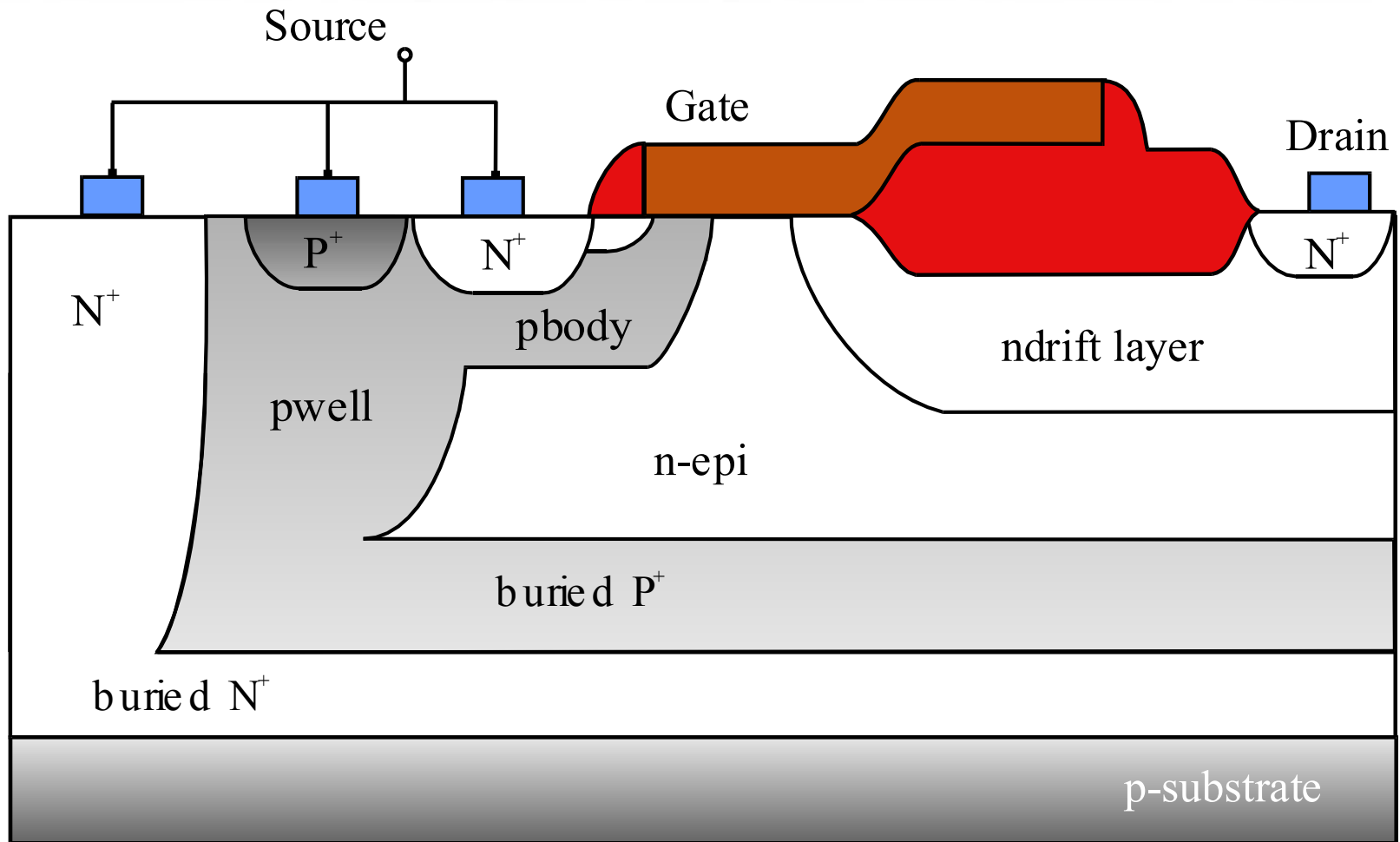


- ***Miller charge and capacitance***
  - ◆ Require energy from the gate driver
  - ◆ Delays a fast turn-on → enlarges the dissipation
- ***$Q_{gd}$  must be as low as possible in order to reduce switching losses***
- ***$R_{on} * Q_{gd}$  is an important figure of merit***



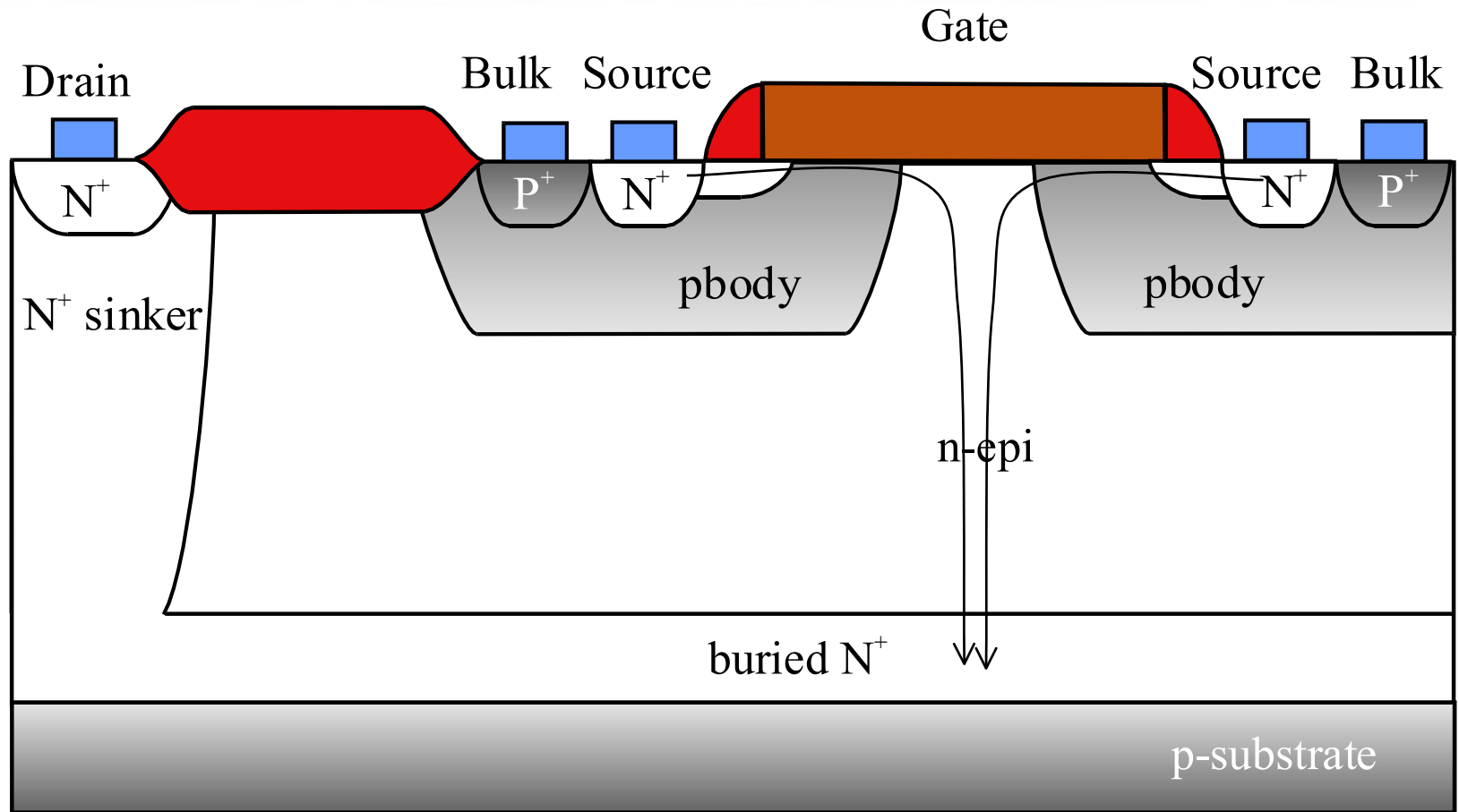
Easy to integrate.

# Resurf LDMOS



Floating RESURF LDMOS in n-epi

# Q-VDMOS

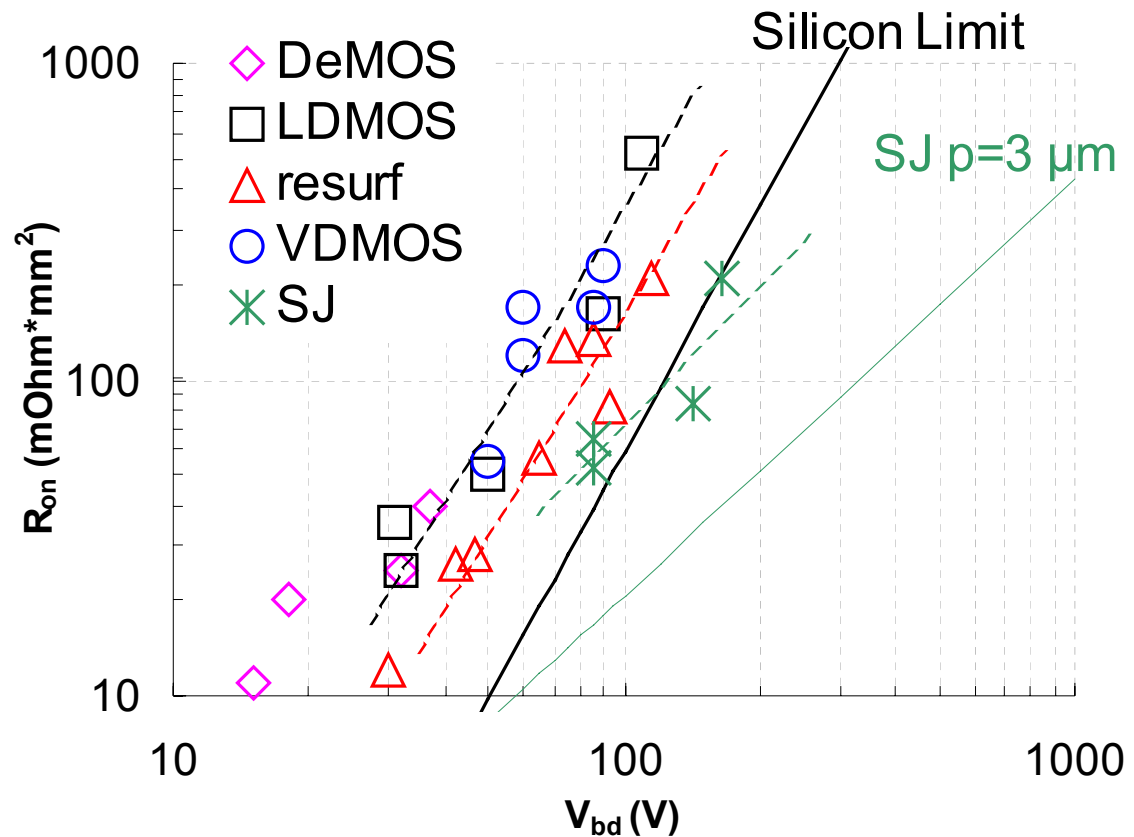


- **VDMOS:  $V_{bd}$  not scalable, depends on technology**
- **JFET principle protects the gate  $\Rightarrow$  ROBUST**

# Silicon limit & Benchmarking

- **Silicon limit expresses the relation between  $R_{on}$  and  $V_{bd}$  for an ideal vertical MOS [42]**

$$R_{on} \sim V_{bd}^{2.5}$$



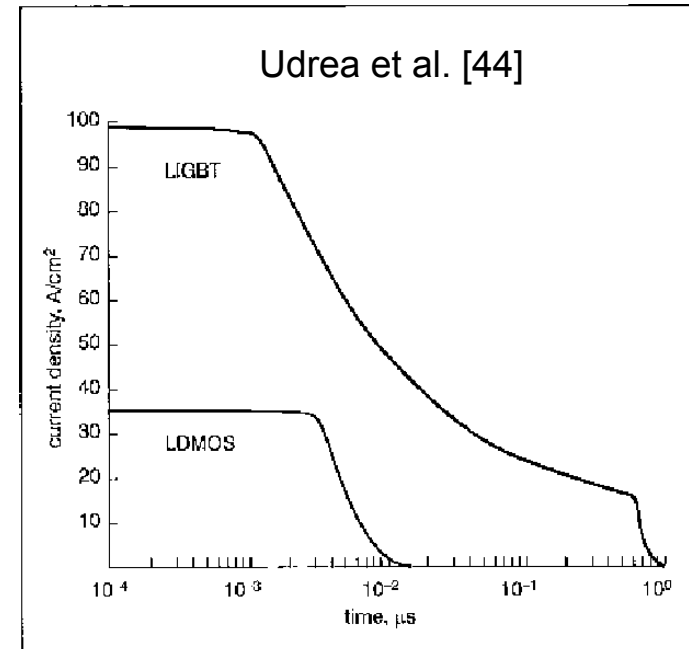
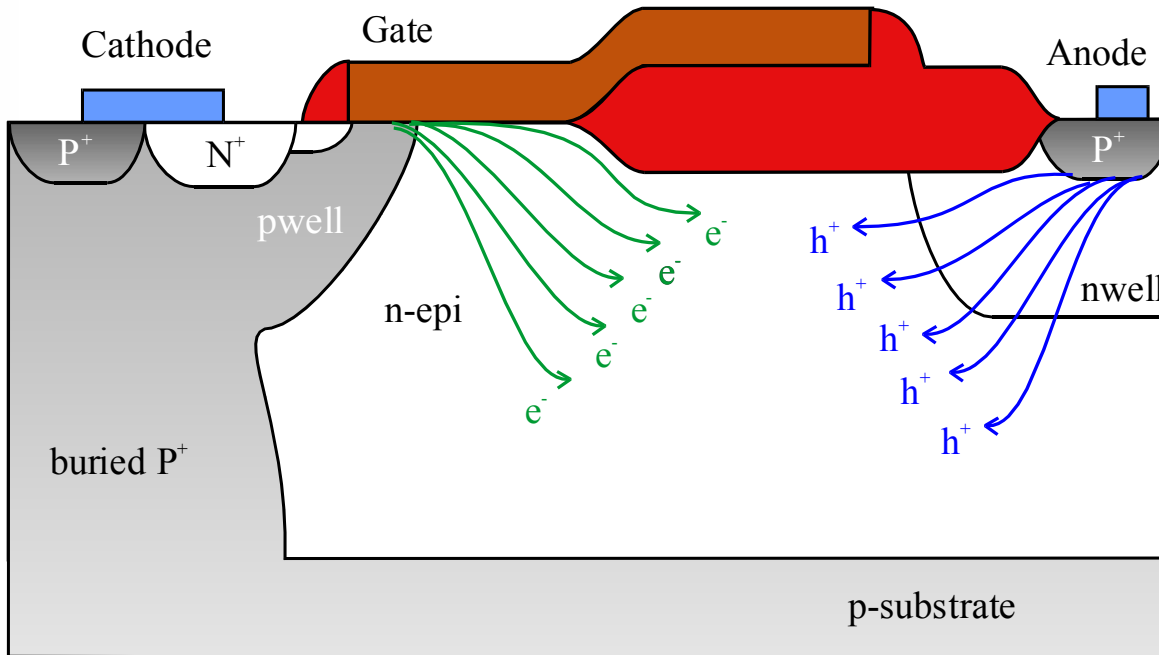


Fig. 9 Turn-off characteristics for 600 V SOI LDMOSFET and LIGBT

Conductivity Modulation  $\Rightarrow R_{on}$  drops

On-State: injection of holes & electrons results in an excess of mobile carriers: “Conductivity Modulation”

Turn-Off: slow due to excess of minority carriers in n-epi [43-45]

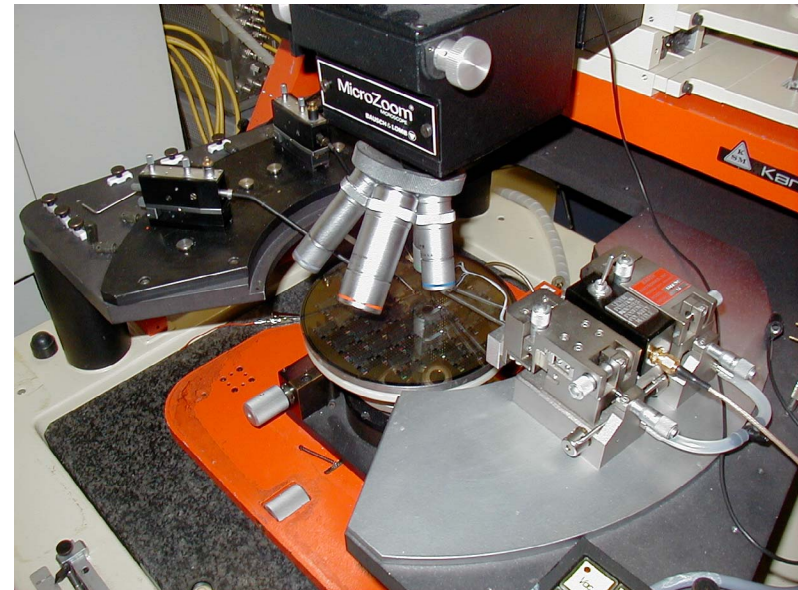
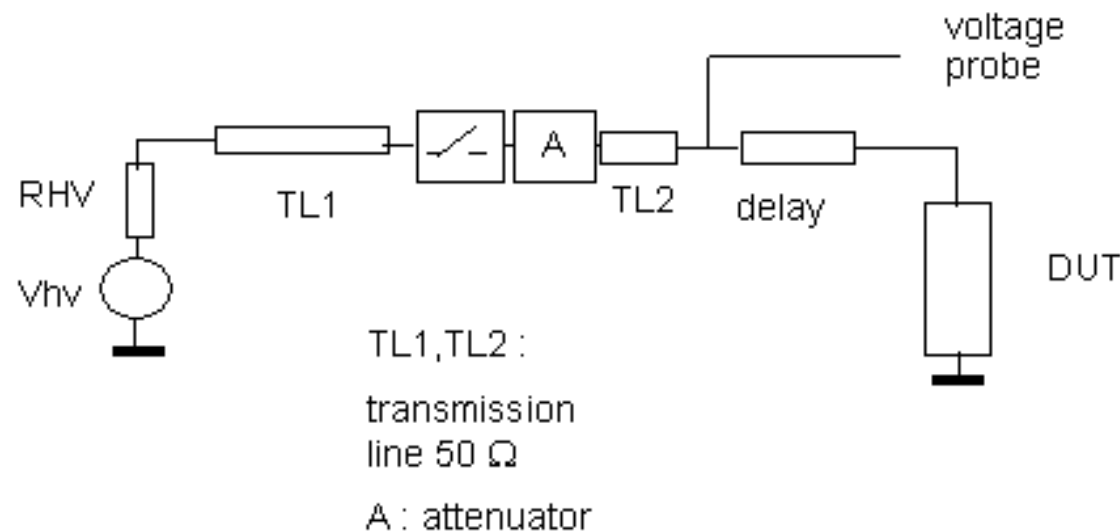
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# ***Reliability measurements & Techniques***

- ***Transmission line pulsing (TLP)***
- ***Transmission Interferometric Mapping (TIM)***
- ***Energy Capability (EC)***
- ***Charge Pumping (CP)***
- ***CV measurements***

# Transmission Line Pulsing

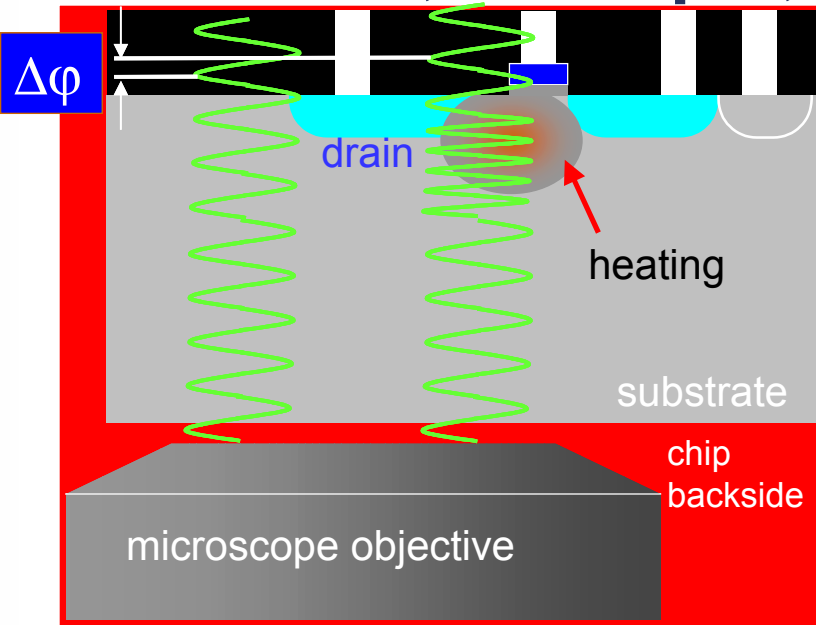
- **Measure the npn turn-on without self-heating of the device. Apply a sufficiently short (5 ns – 200 ns) power pulse [46]**



- **On wafer and/or packaged devices**

# Transient Interferometric Mapping

IR laser,  $\lambda=1.3 \mu\text{m}$ , transparent for silicon [47-49]



*Temperature and carrier concentration variation*



*Change in refractive index of the silicon*

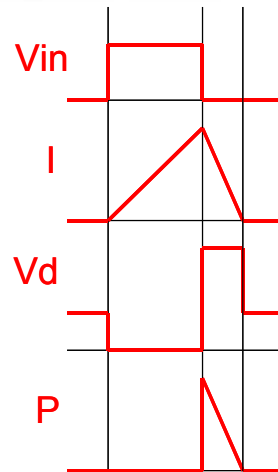
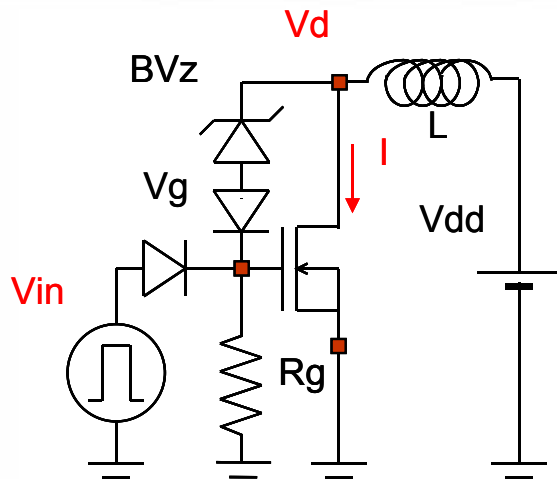


*Change in phase shift  $\Delta\phi$*

$$\Delta\phi(t) = \frac{4\pi}{\lambda} \int_0^L \left( \frac{dn}{dT} \Delta T(z,t) + (\alpha_n \Delta n(z,t) + \alpha_p \Delta p(z,t)) \right) dz$$

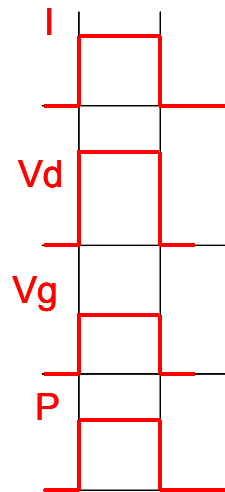
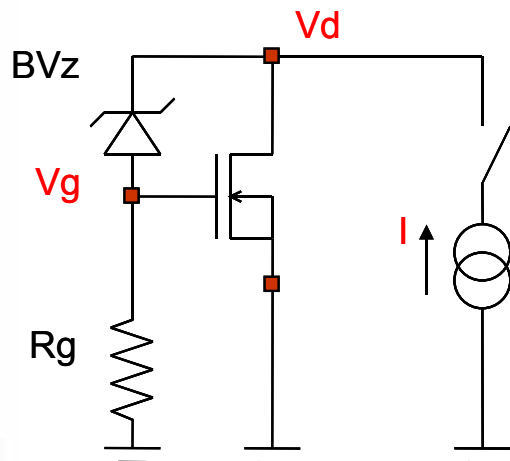
**Spatial resolution :  $1.5 \mu\text{m}$ , time resolution : 3 ns.**

# Energy Capability (1)



Clamped inductive switching (CIS) : power level and time cannot be controlled independently.

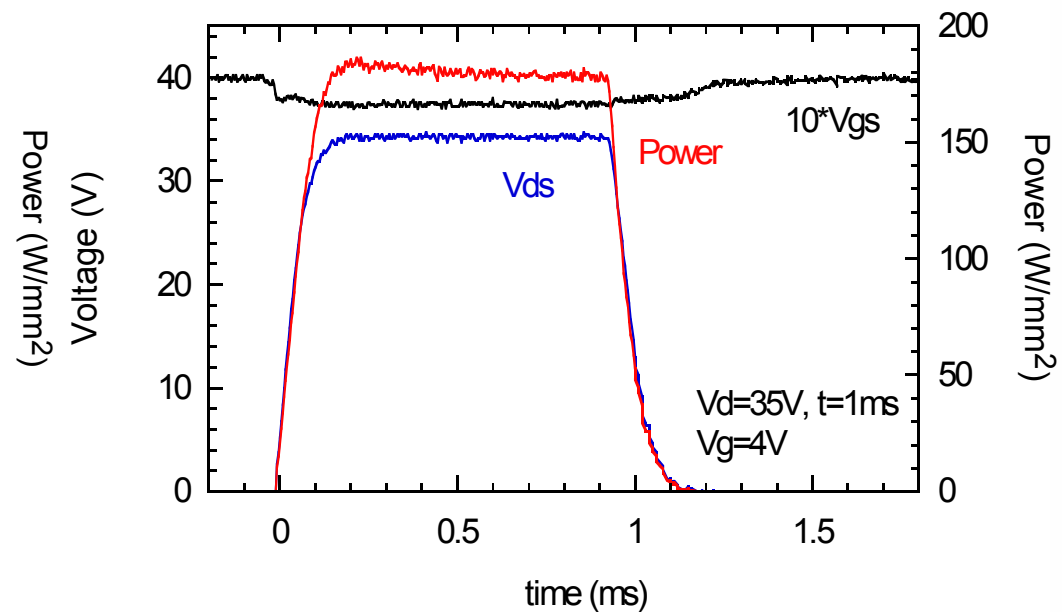
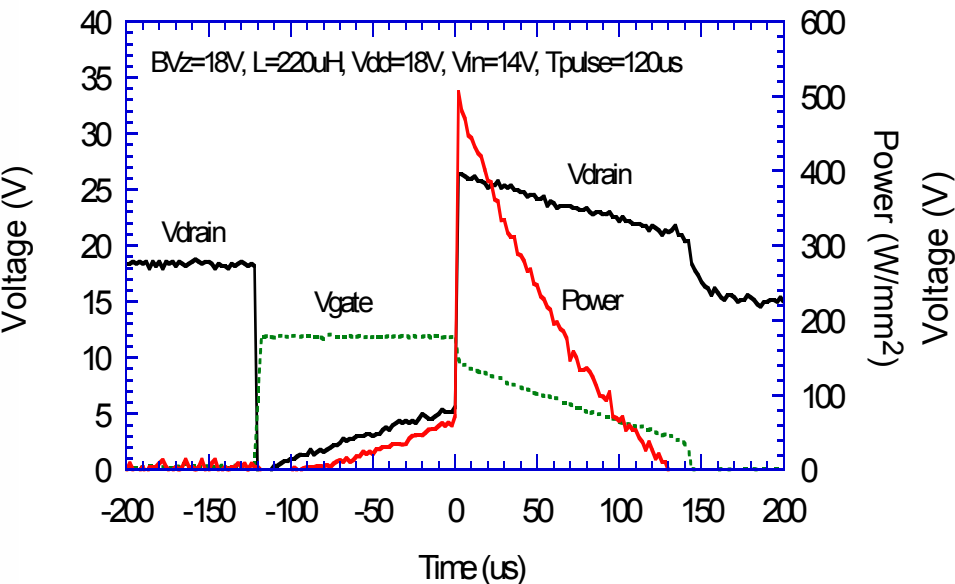
CIS and RPP yield the same results [41] (theory+exp)



Rectangular power pulsing (RPP) : independent control of Power and time.

# Energy Capability (2)

- For CIS, the power pulse shape is triangular
- For RPP, the power pulse is rectangular

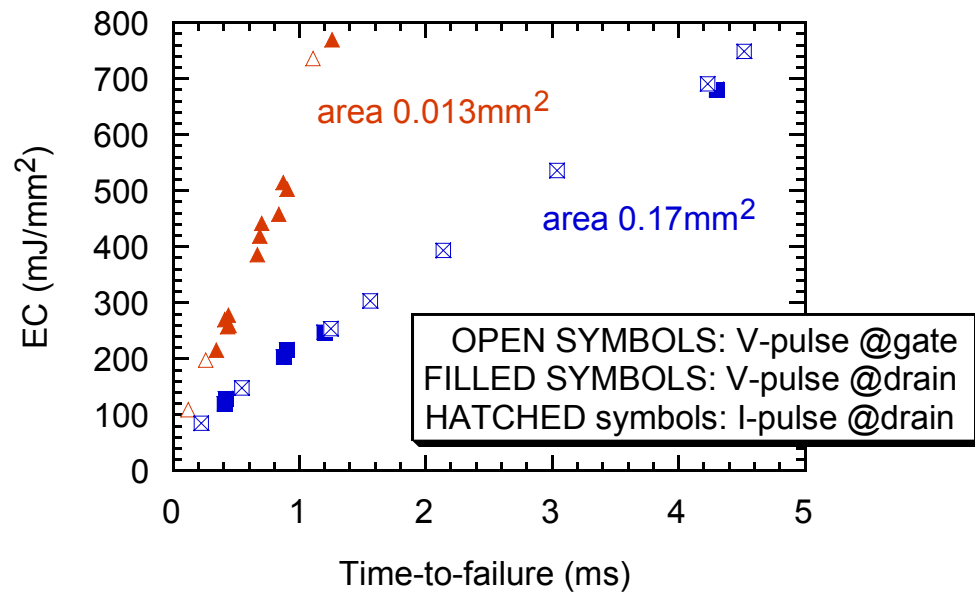
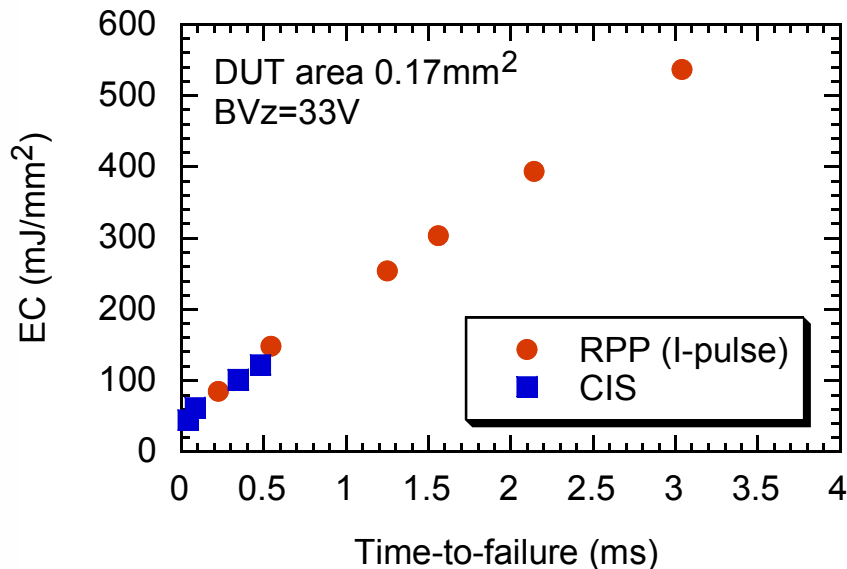


# Energy Capability (3)

- Theoretical analysis [41]**

$$P_{0,rect} \leq P_{0,tr} \leq \frac{3}{2} P_{0,rect}$$

$$EC_{rect} \leq EC_{tr} \leq \frac{9}{8} EC_{rect}$$

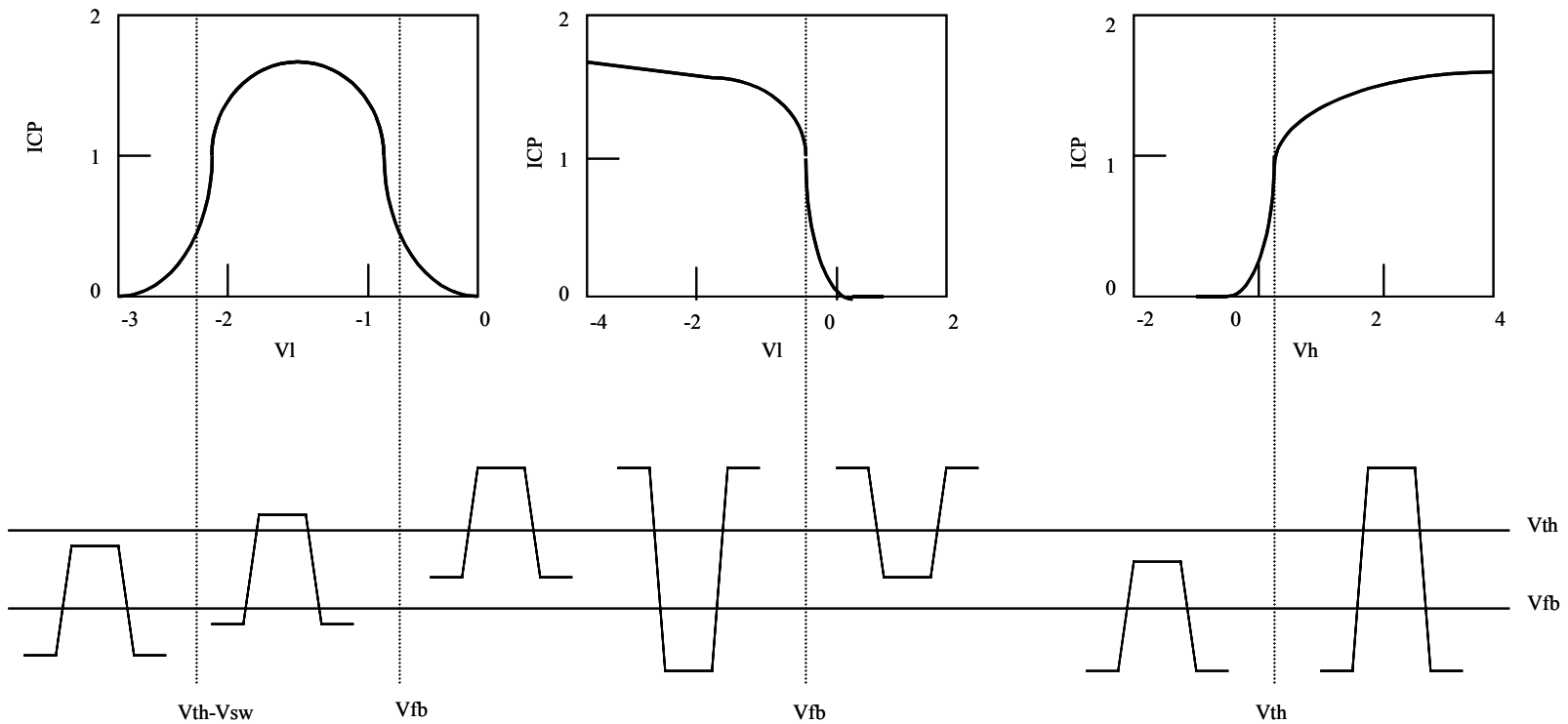


# Charge Pumping (1)

- To locate and quantify the  $N_{it}$  generation [50,51]

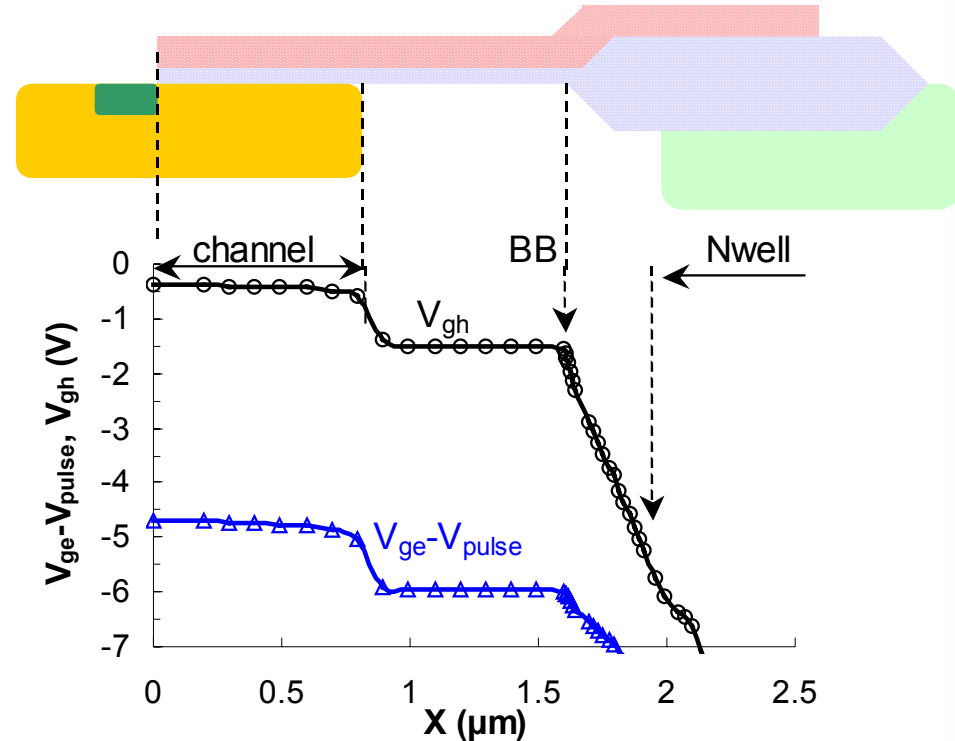
*variable base*

*variable amplitude*



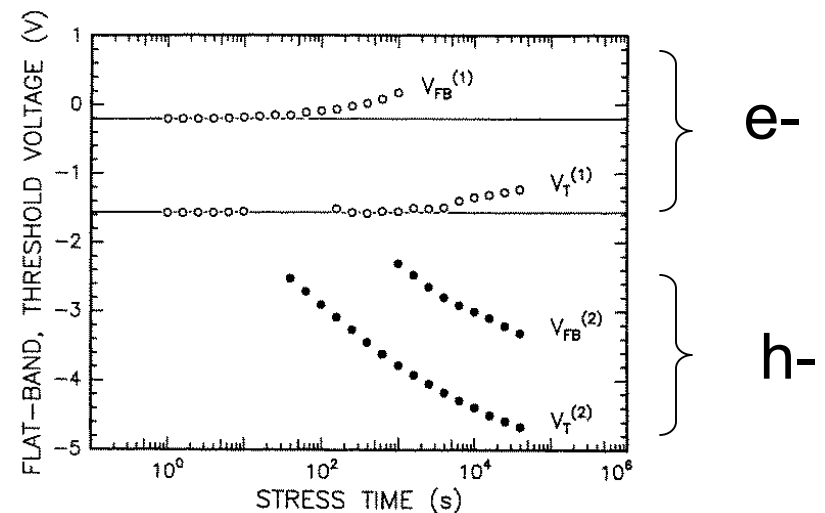
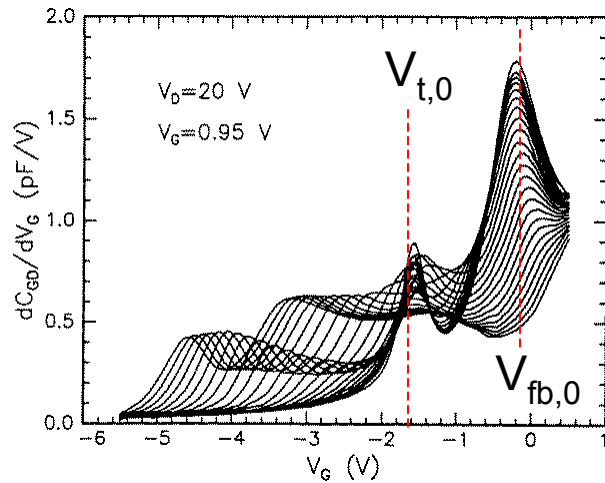
# Charge Pumping (2)

- $V_{th}$  ( $V_{g\_e}$ ) and  $V_{fb}$  ( $V_{g\_h}$ ) depend on
  - ◆ frequency of the pulse
  - ◆  $N_d$ ,  $t_{ox}$  and  $N_{ot}$  at the Si/SiO<sub>2</sub> interface
- Region that can be probed is dependent on  $V_{pulse}$ , hence on  $t_{ox}$  thin gate and on  $N_d$ .
- Very well applicable to DeMOS, LDMOS and VDMOS [19-21]



# CV measurements

- **Assessment of charge trapping in LDMOS using CV measurements [10,16].**
- **$C_{gd}$  monitored at 1 MHz.**



- **Information on charge trapping is extracted.**
- **Difficult to pinpoint location of trapping.**

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# ***Reliability Problems in Smart Power***

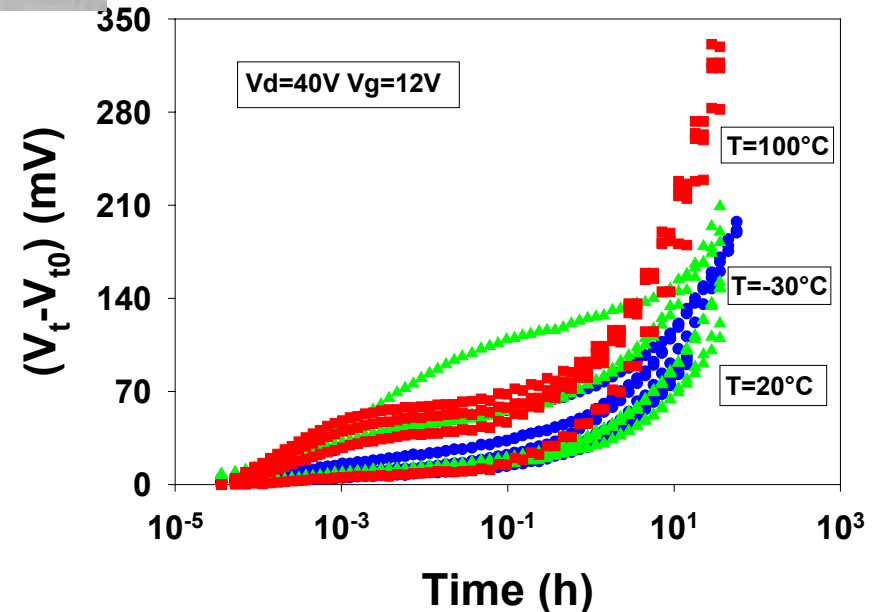
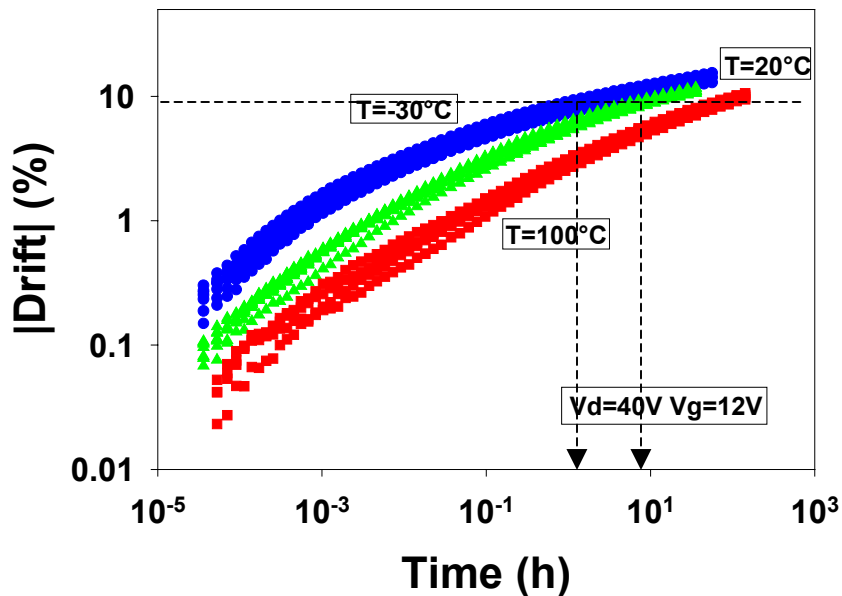
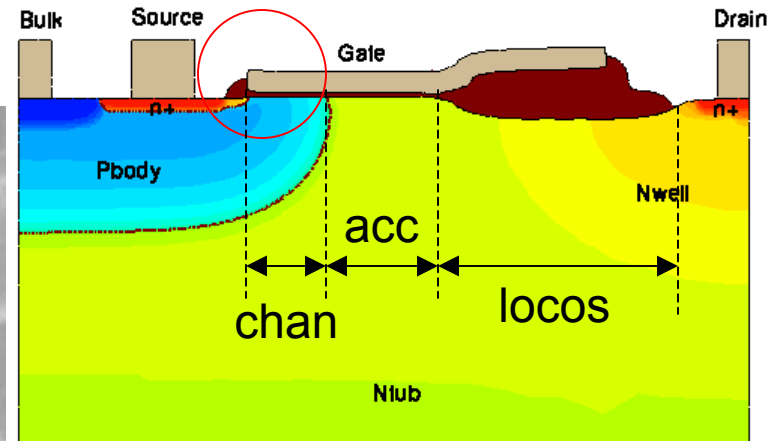
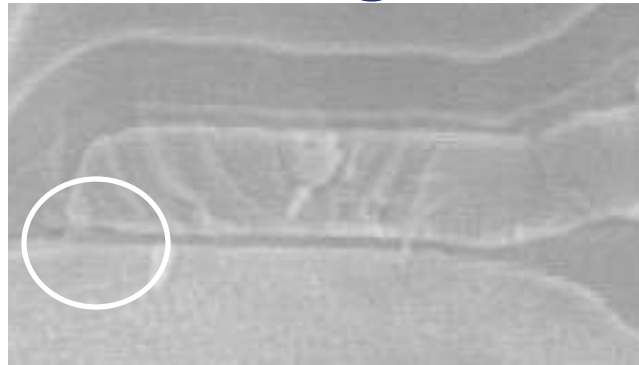
- ***DC hot carrier***
- ***AC hot carrier (switching of a load)***
- ***Electrical SOA (parasitic bipolar turn-on)***
- ***Thermal SOA (energy capability)***
- ***Simulation of thermal effects***

# ***Reliability Problems in Smart Power***

- ***DC hot carrier***
- *AC hot carrier (switching of a load)*
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- *Thermal SOA (energy capability)*
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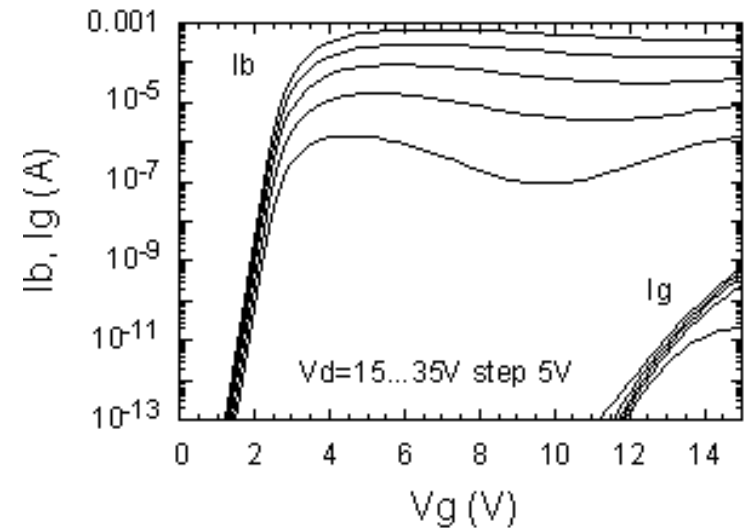
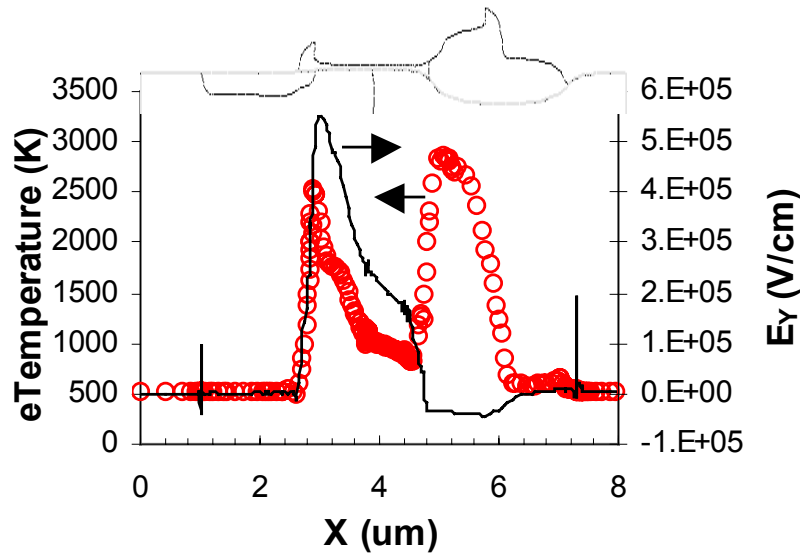
# DC hot carrier (1)

- **40V LDMOS 12V gate**



# DC hot carrier (2)

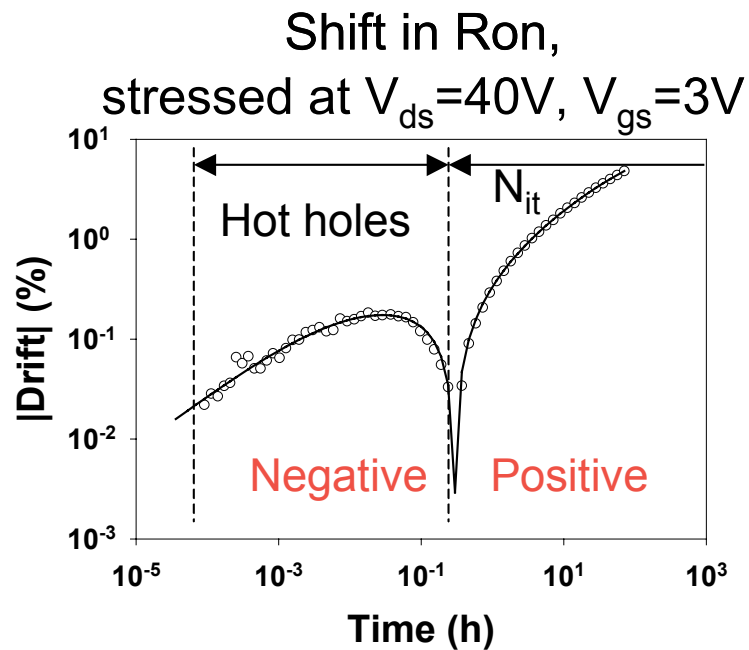
- **Source Side Injection (SSI) at high  $V_{gs}$  [52,53]**



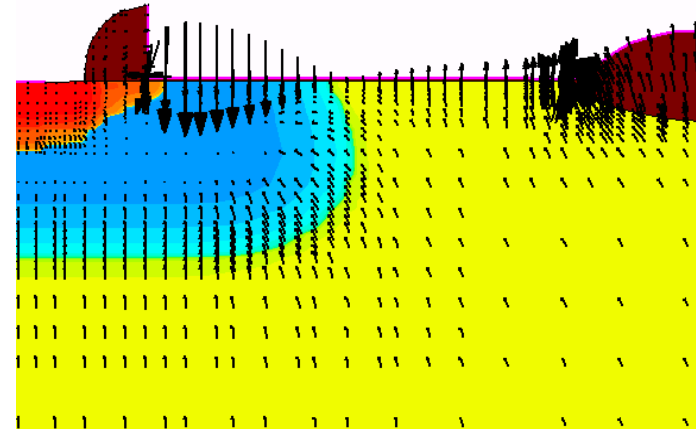
- **$n\text{LDD}$  dose is too low for self-aligned p-body dose : E-field build-up at the source**
- **Carrier Injection at the source  $\rightarrow V_{th}$  shift, independent of  $V_{ds}$**

# DC hot carrier (3)

- **Hole Injection at low  $V_{gs}$  and high  $V_{ds}$ .**
- **Occurs at birds beak ( $E$ -field direction).**

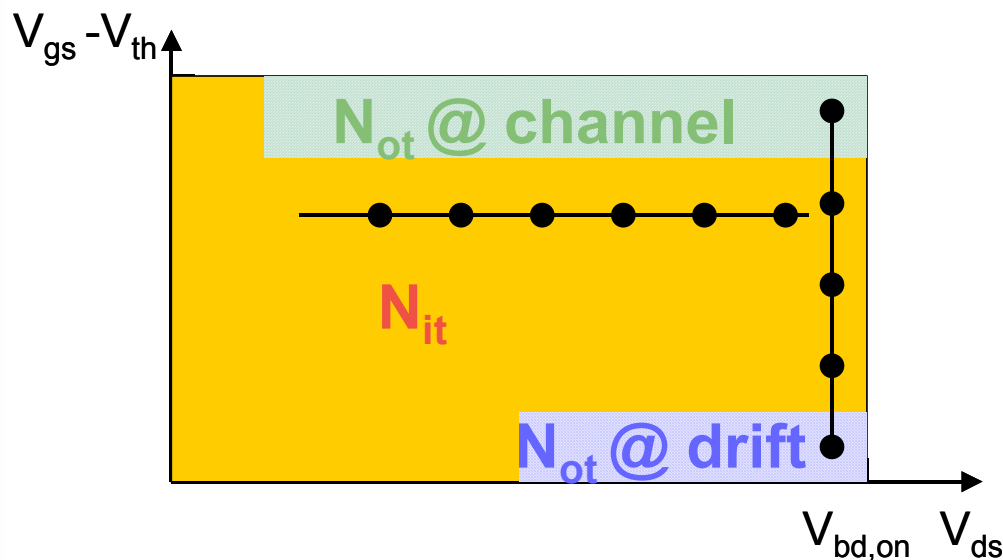


Electric field @  $V_{gs}=6V$  ;  $V_{ds}=40V$

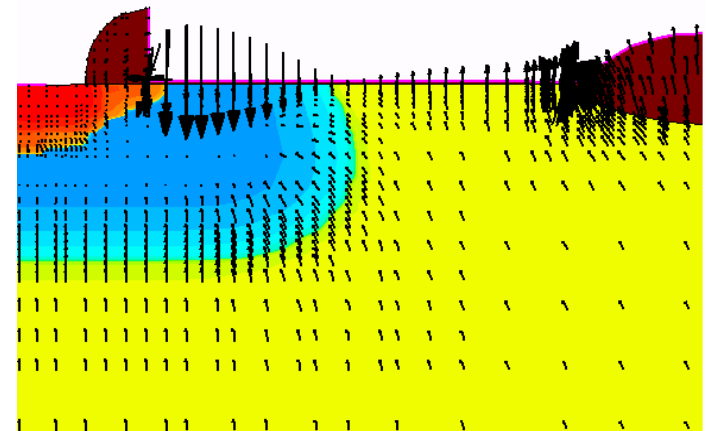


# DC hot carrier (4)

- $V_{gs}/V_{ds}$  dependency of the degradation
  - ◆ Different degradation mechanisms can occur.
  - ◆ Dependent on device specifics (layout, doping, ...)
  - ◆ At which conditions to stress for worst case ?

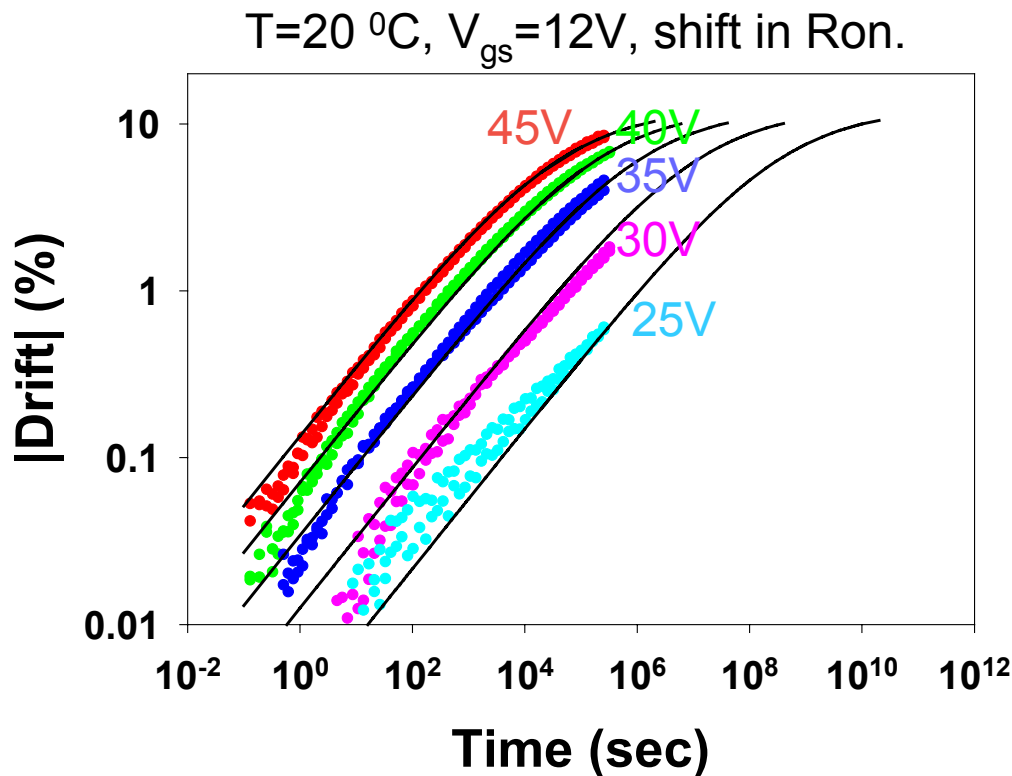


Electric field @  $V_{gs}=6V$  ;  $V_{ds}=40V$



# DC hot carrier (5)

- $V_{ds}$  dependency : simultaneous fit



$$\frac{\Delta P}{P_0} = \frac{C_1 \cdot \left[ t \cdot \frac{I_d}{W} \left( \frac{I_{sub}}{I_d} \right)^{\phi_{it}/\phi_i} \right]^n}{1 + C_2 \cdot \left[ t \cdot \frac{I_d}{W} \left( \frac{I_{sub}}{I_d} \right)^{\phi_{it}/\phi_i} \right]^n}$$

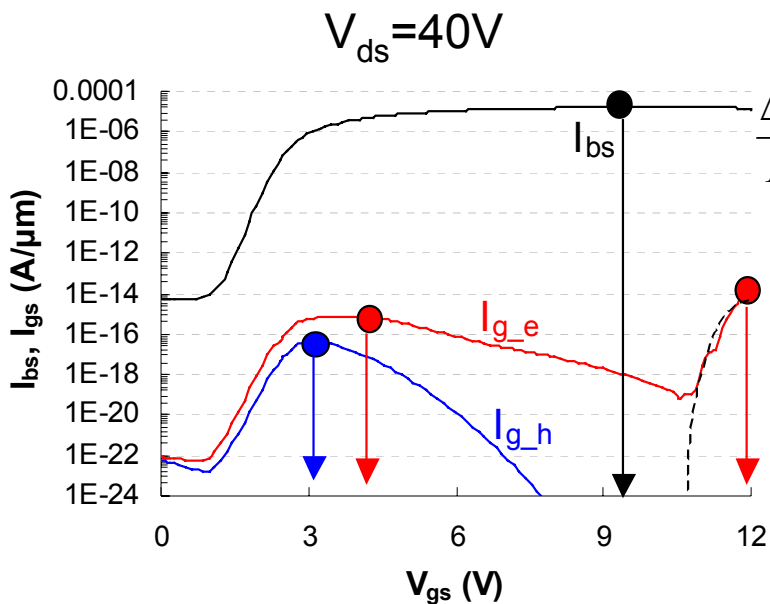
$I_{ds}$ ,  $I_{sub}$  from experiments

$C_1$ ,  $C_2$ ,  $\phi_{it}/\phi_i$ ,  $n$  : fitting parameters

$\swarrow$  3.5 eV (e)     $\searrow$  1.3 eV     $\nearrow$  ~0.4-0.6

# DC hot carrier (6)

- $V_{gs}$  dependency : different mechanisms
  - ◆ Low  $V_{gs}$  : hole injection at the birds beak
  - ◆ High  $V_{gs}$  : electron injection at source side
  - ◆ Intermediate  $V_{gs}$  :  $N_{it}$  formation



$$\frac{\Delta I_d}{I_{d0}} = \frac{C_1 \cdot \left( t \cdot \frac{I_d}{W_{eff}} \cdot \left( \frac{I_{sub}}{I_d} \right)^{\frac{\phi_{it}}{\phi_i}} \right)^n}{1 + C_2 \cdot \left( t \cdot \frac{I_d}{W_{eff}} \cdot \left( \frac{I_{sub}}{I_d} \right)^{\frac{\phi_{it}}{\phi_i}} \right)^n} + C_3 \cdot \left( t \cdot \frac{I_d}{W_{eff}} \cdot \left( \frac{I_{sub}}{I_d} \right)^{1 + \frac{\phi_{it,h} \cdot \lambda_e}{\phi_i \cdot \lambda_h}} \right)^m$$

$C_3$  negative for hole trapping

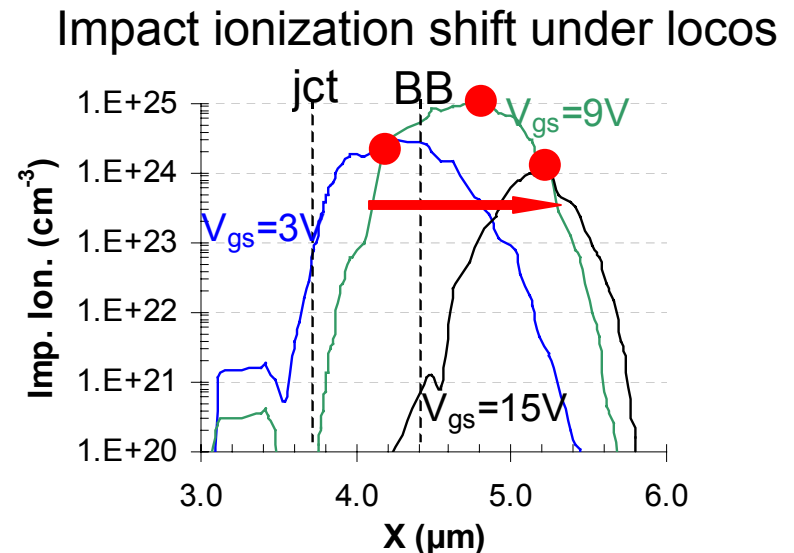
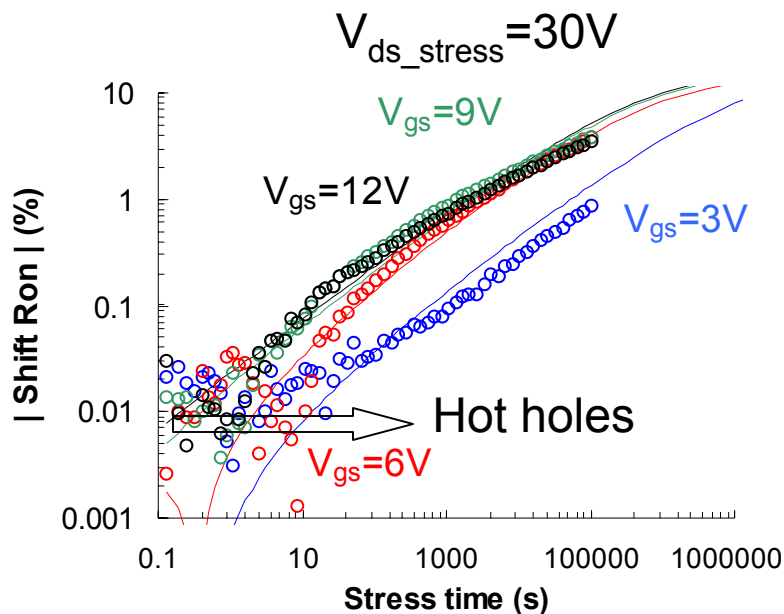
$\lambda_e, \lambda_h$  : mean free paths of electrons and holes

$\phi_{it,h}$  : 4.2 eV

# DC hot carrier (7)

- **$V_{gs}$  dependency :**

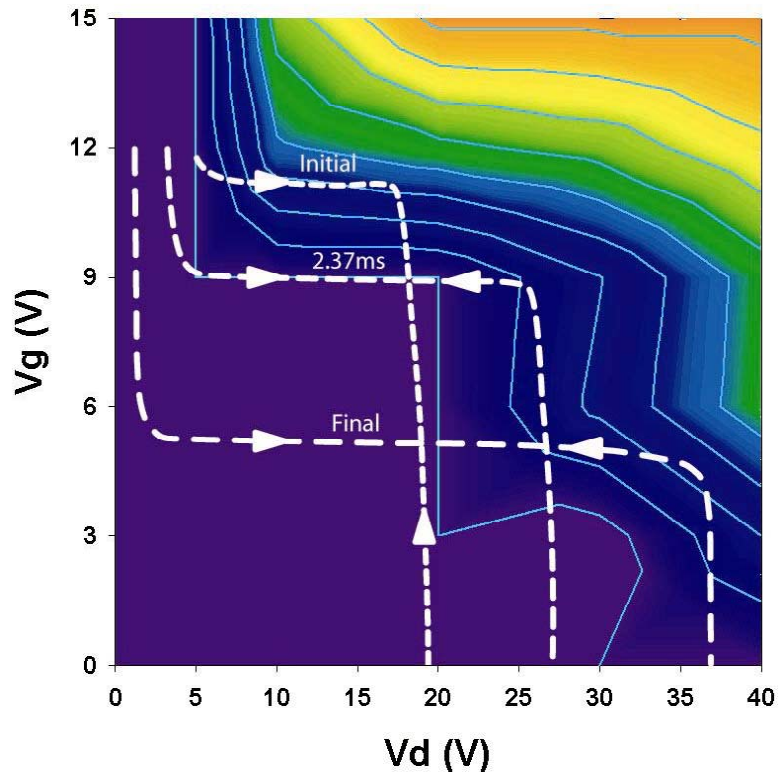
- ◆ Difficult (if not impossible) to a priori know WC  $V_{gs}$
- ◆ Measure HC degradation for different  $V_{gs}$
- ◆ Difficult to fit different  $V_{gs}$  data simultaneously, fitting yields non-physical parameters ( $\phi_{it}/\phi_l < 1$ )



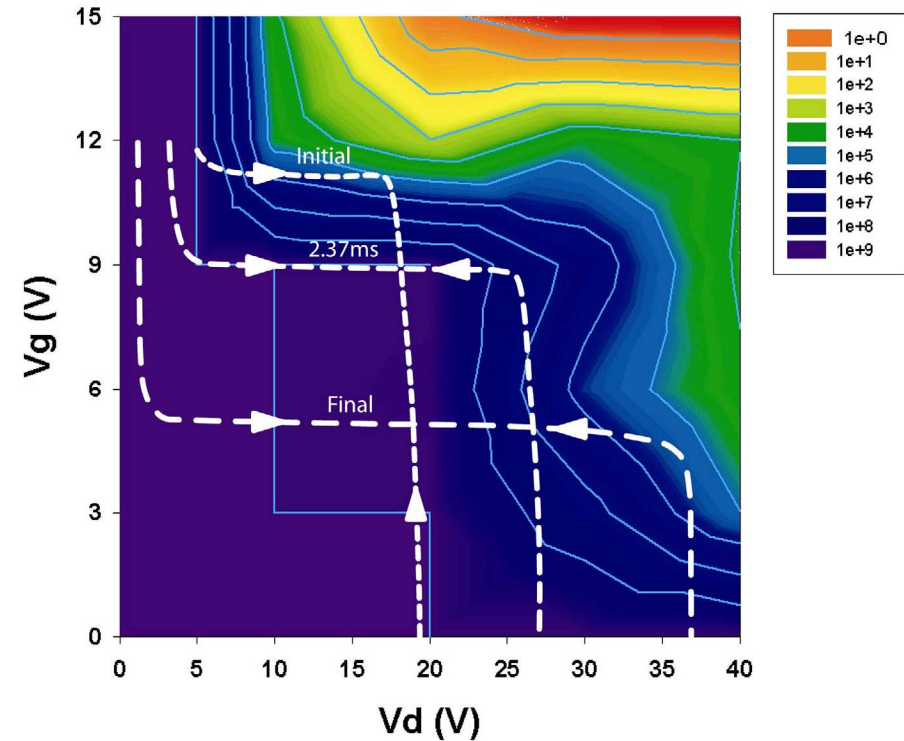
# DC hot carrier (8)

- LDMOS 40V in a 0.7  $\mu\text{m}$  smart power techno**

T=25 °C

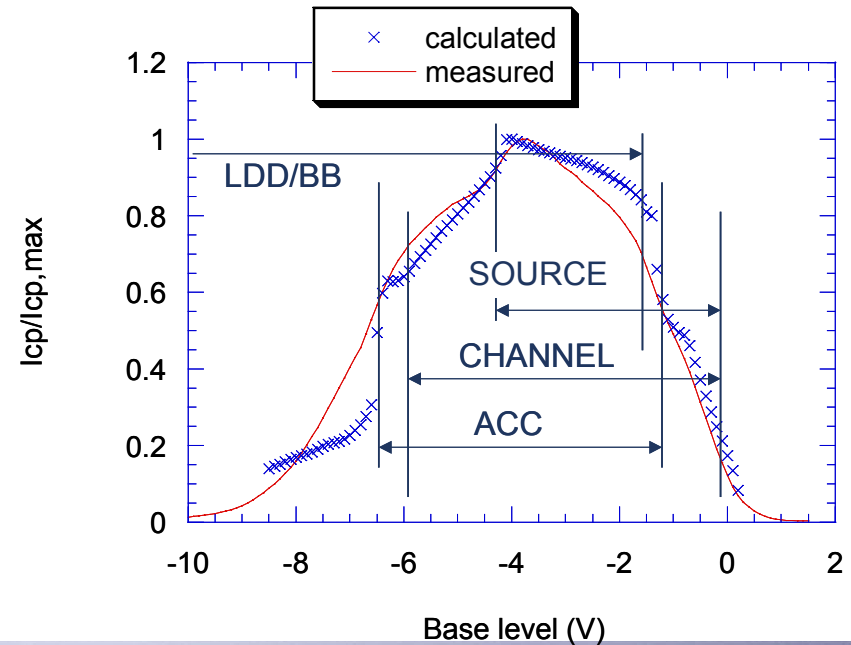
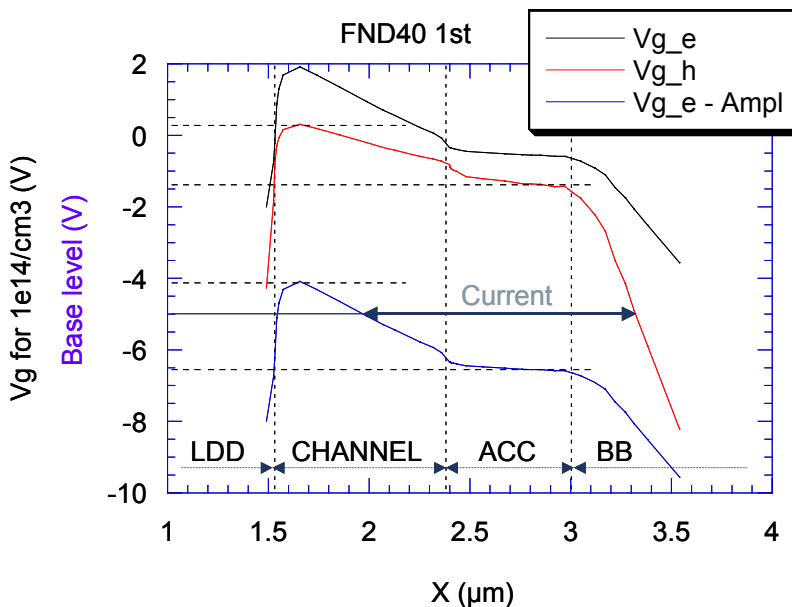
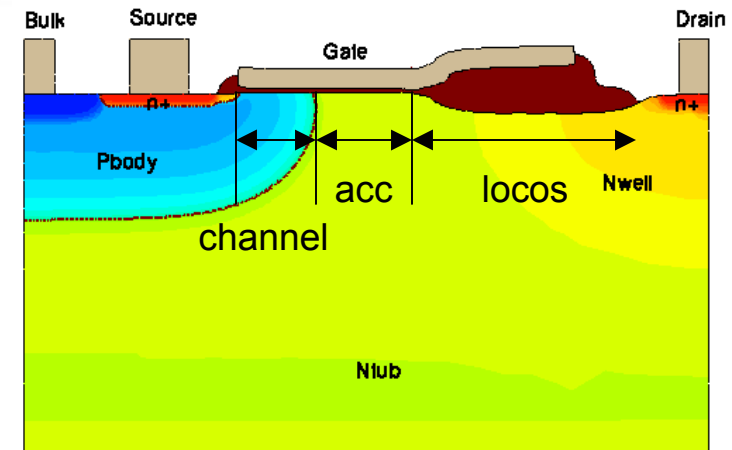


T=-30 °C



# Charge Pumping (1)

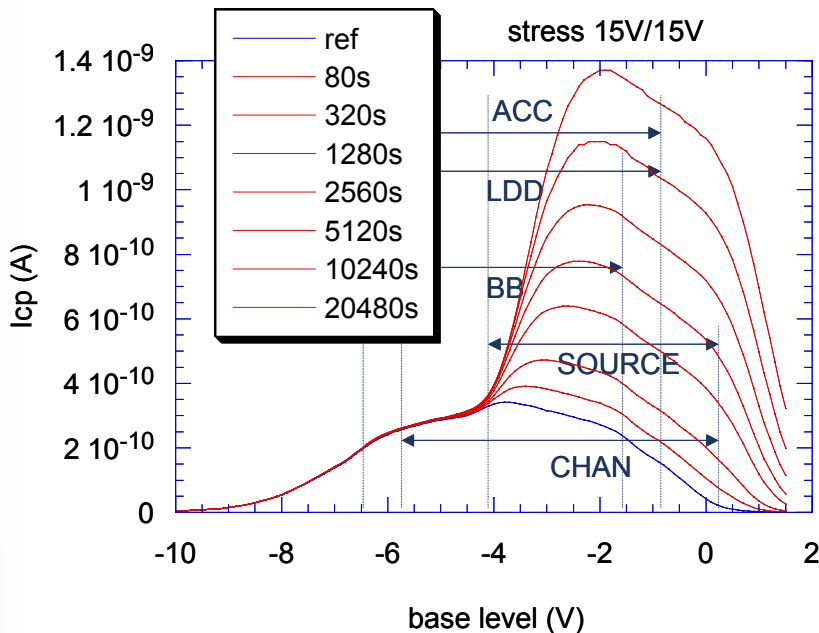
- **40V LDMOS**
- $V_{ge}$  and  $V_{gh}$  as from TCAD
- **Uniform  $N_{it}$  in thin oxide**



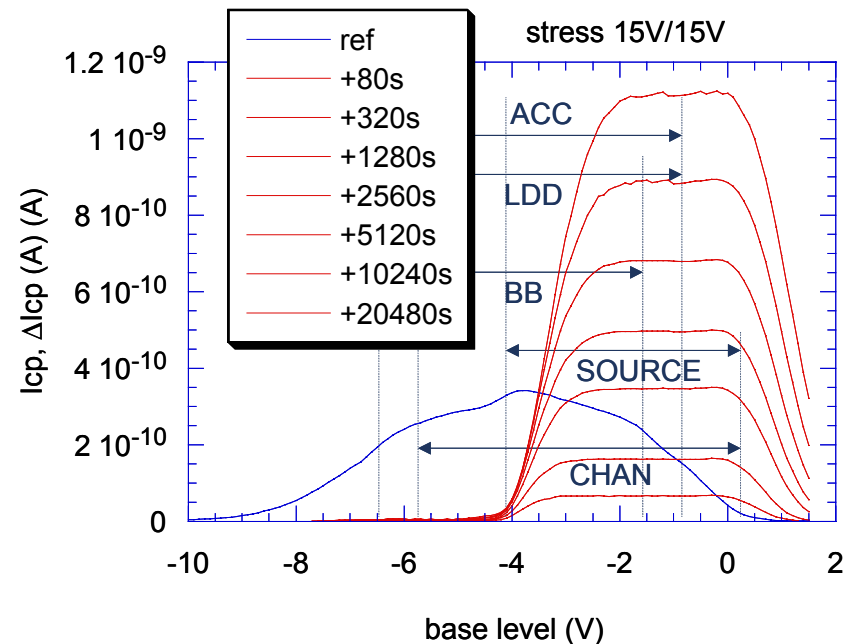
# Charge Pumping (2)

- LDMOS stressed at  $V_{ds}=V_{gs}=15V$  (low  $V_{ds}$ , high  $V_{gs}$ )  $\rightarrow N_{it}$  formation at the source.**

Charge pumping spectra

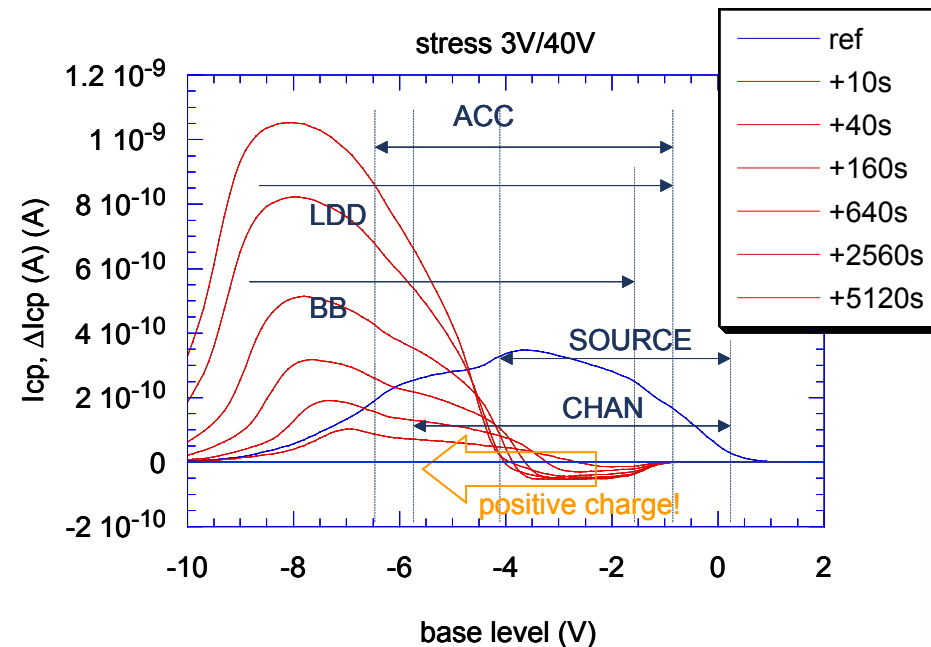
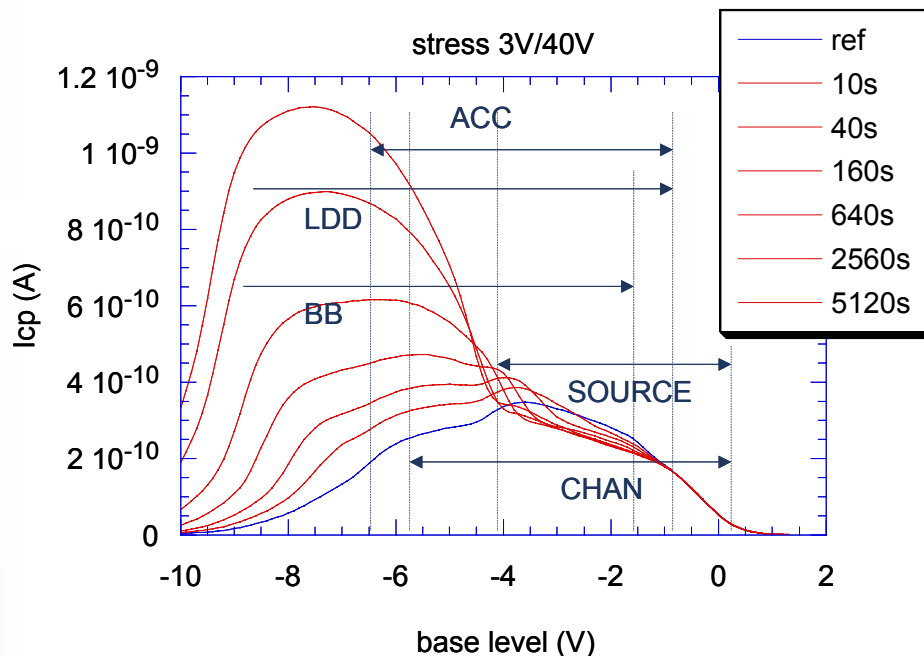


Differential CP spectra



# Charge Pumping (3)

- LDMOS stressed at  $V_{ds}=40$   $V_{gs}=3V$  (low  $V_{gs}$ , high  $V_{ds}$ )  $\rightarrow N_{it}$  formation in accumulation region or under the birds beak.**

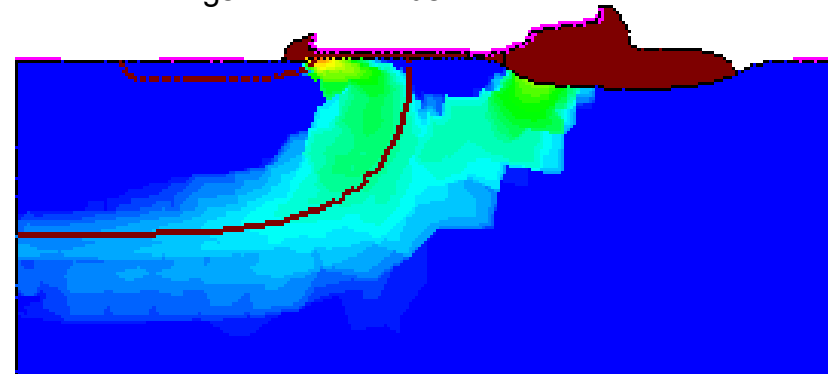
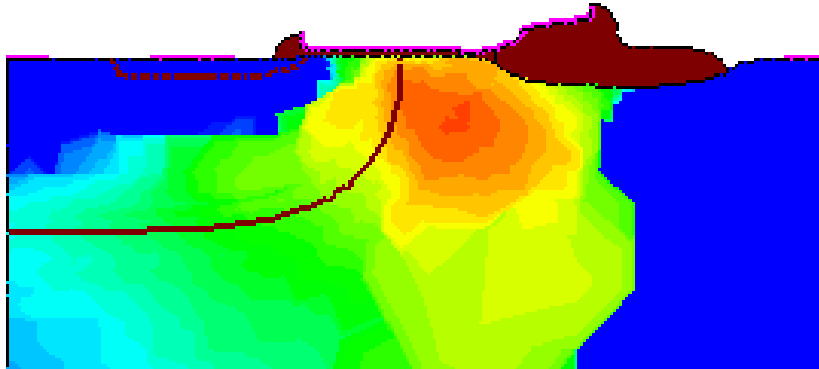


# TCAD verification on 40V LDMOS

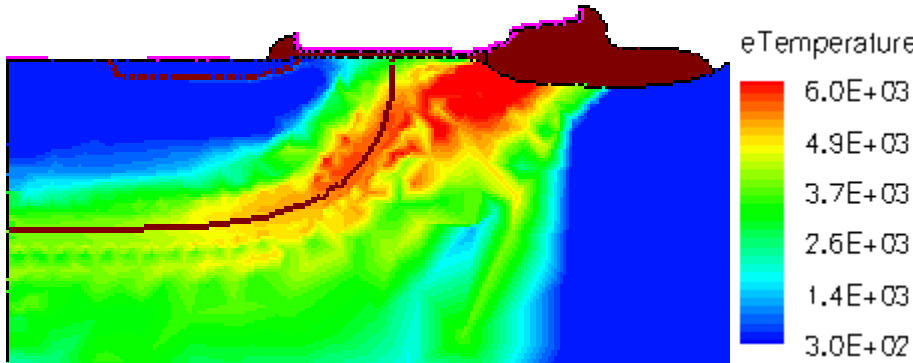
$V_{gs}=3V, V_{ds}=40V$

$V_{gs}=15V, V_{ds}=15V$

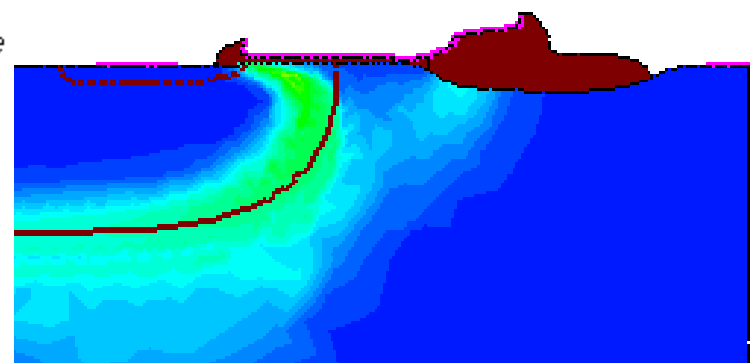
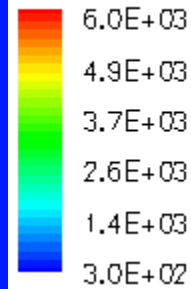
Impact  
ionization



Electron  
Temperature

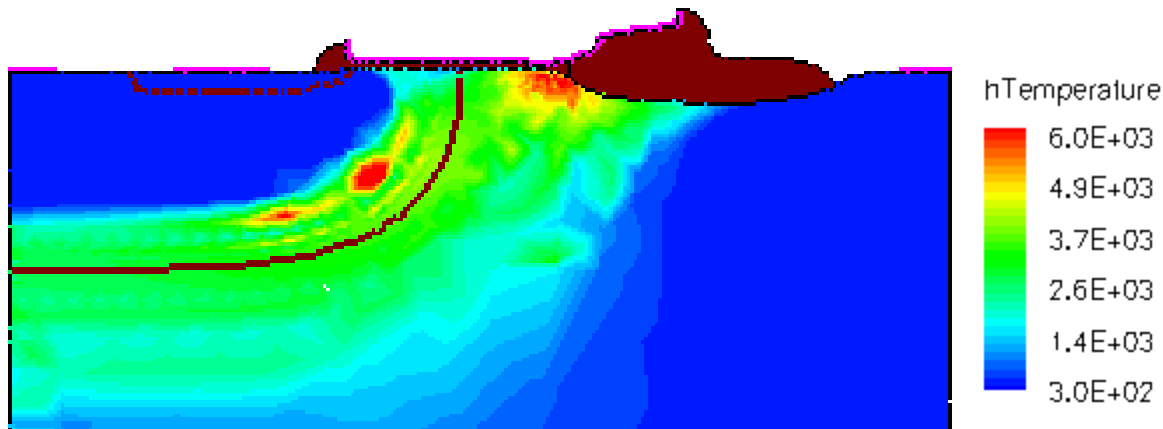


eTemperature



# TCAD verification on 40V LDMOS

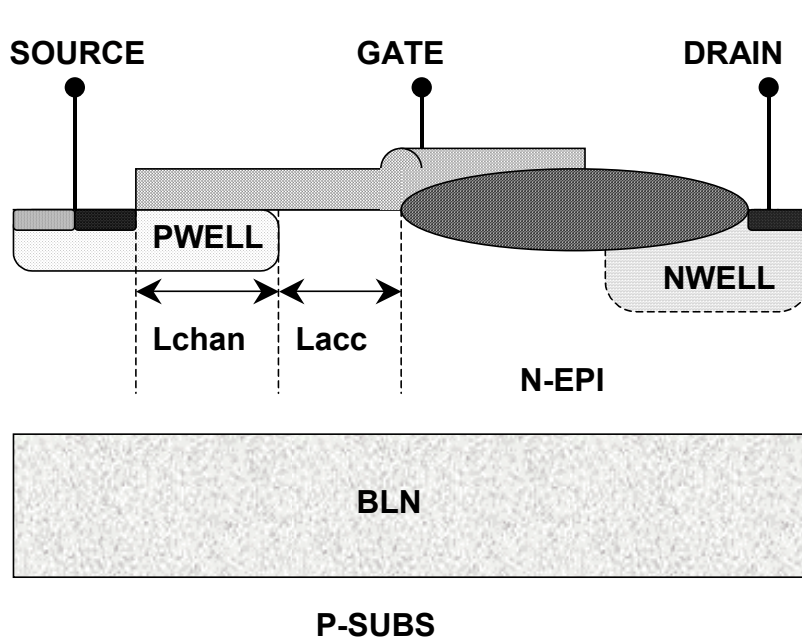
- *At low  $V_{gs}$  (3V) and high  $V_{ds}$  (40V) : hot holes are present at the device birds beak.*



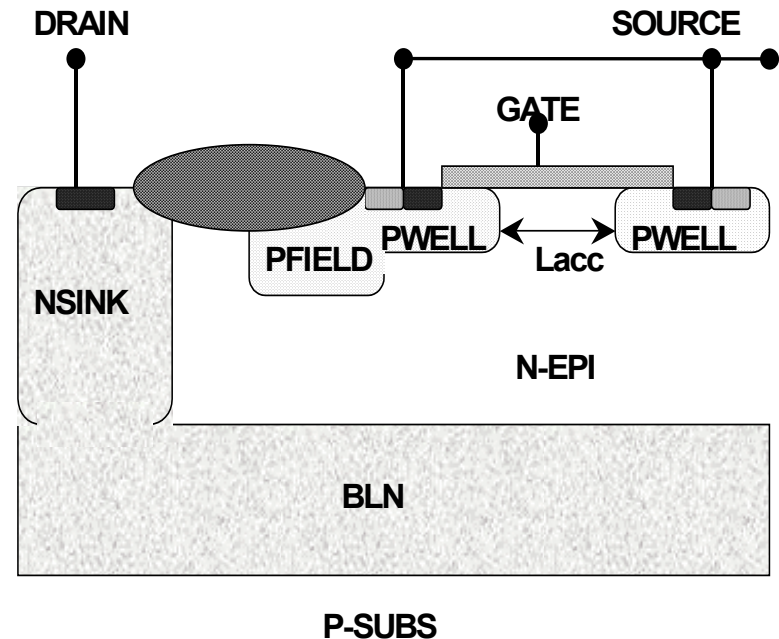
- *TCAD data confirm analysis results.*

# DC hot carrier : LDMOS vs VDMOS (1)

- 80V LDMOS and VDMOS in 0.35  $\mu\text{m}$  SPT [54]



**LDMOS**

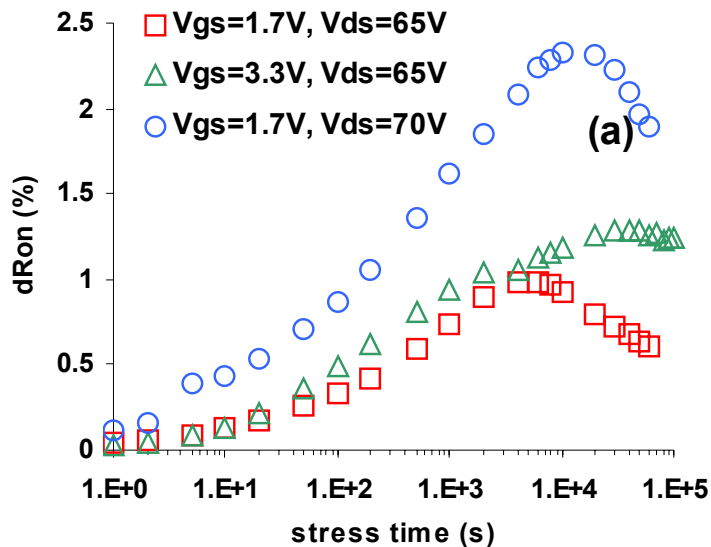


**VDMOS**

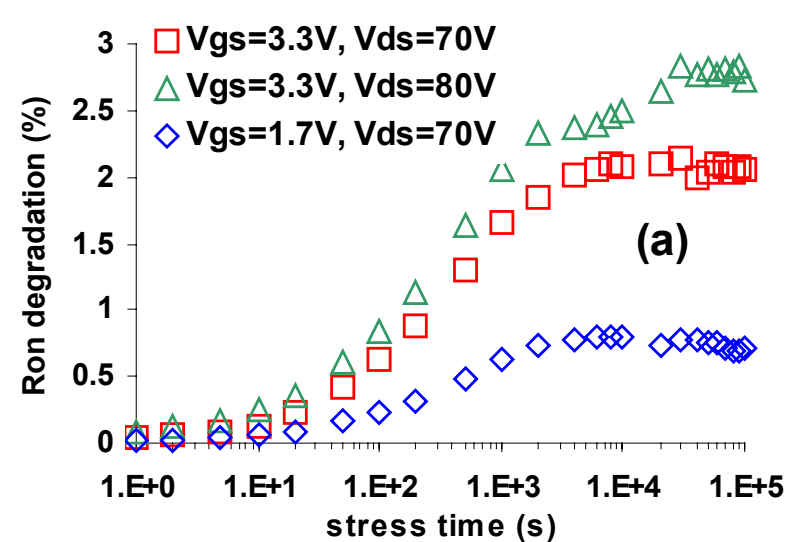
# DC hot carrier : LDMOS vs VDMOS (2)

- **Primary degradation point in LDMOS is birds beak (tip).**
- **No birds beak is present in VDMOS drift region  $\rightarrow$  expected to yield better for HC.**

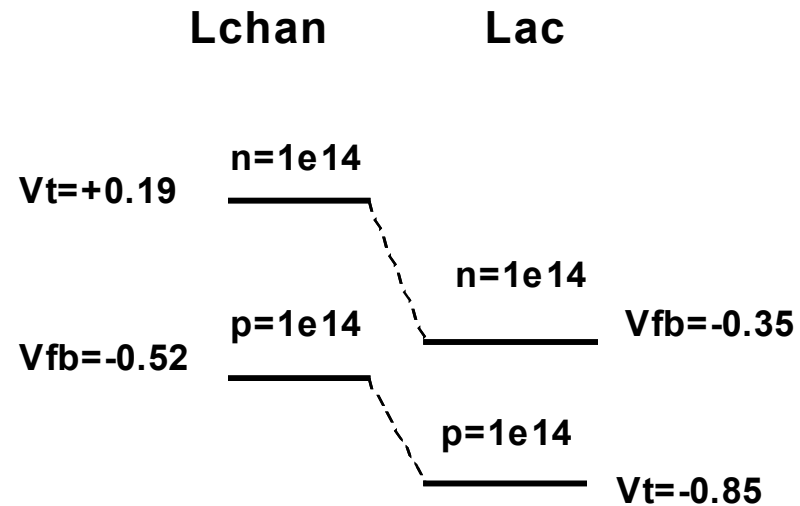
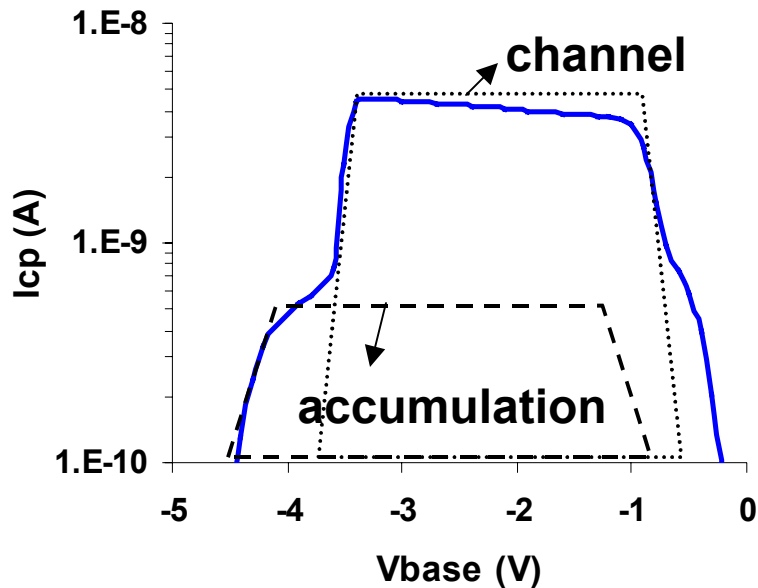
80V LDMOS in 0.35 $\mu$ m SPT



80V VDMOS in 0.35 $\mu$ m SPT



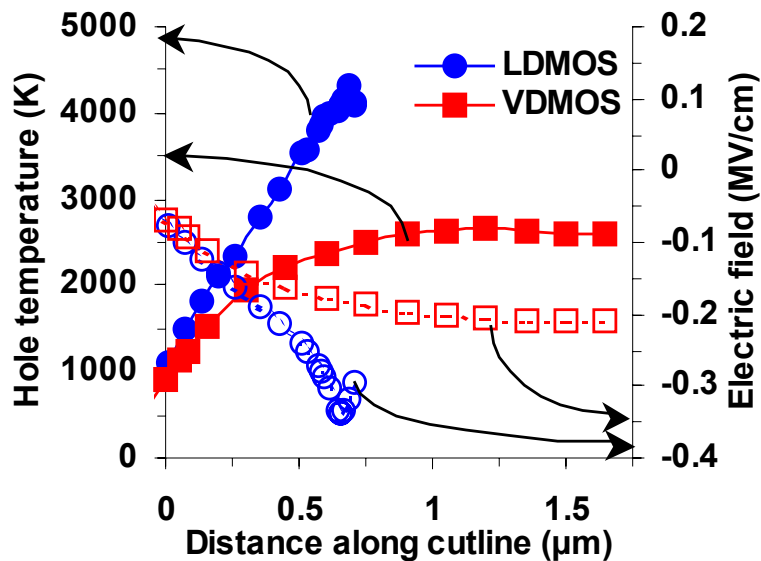
# DC hot carrier : LDMOS vs VDMOS (3)



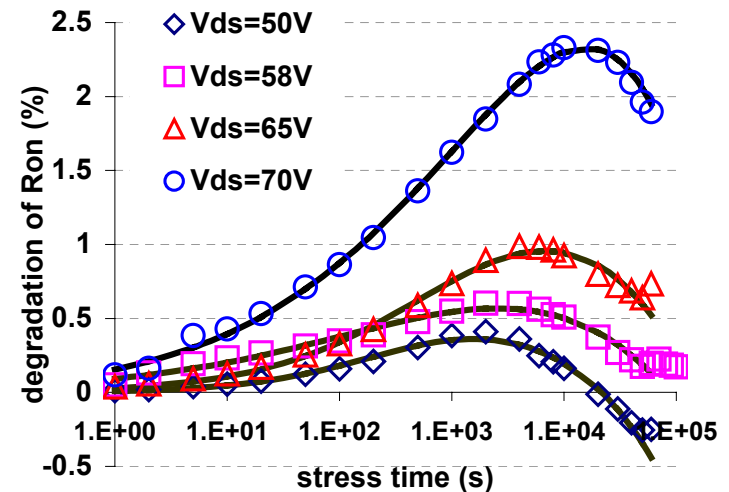
- **VDMOS** :  $N_{it}$  formation only in channel
- **LDMOS** :  $N_{it}$  formation in channel and accumulation region (birds beak).

# DC hot carrier : LDMOS vs VDMOS (4)

- **LDMOS birds beak tip :**
  - ◆ Electric field encroachment
  - ◆ Impact ionization spot : hot holes injection

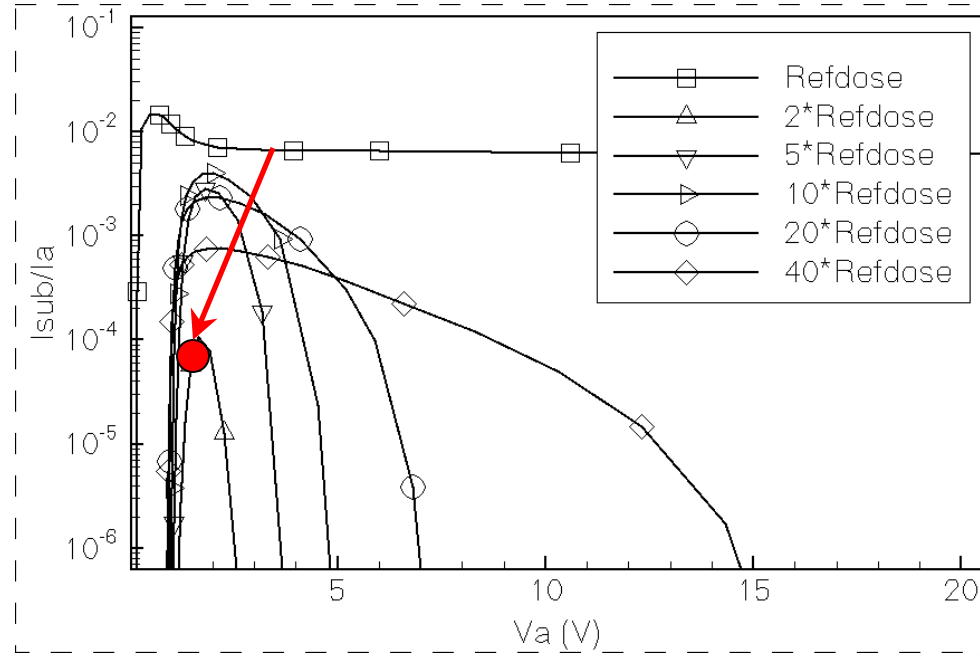
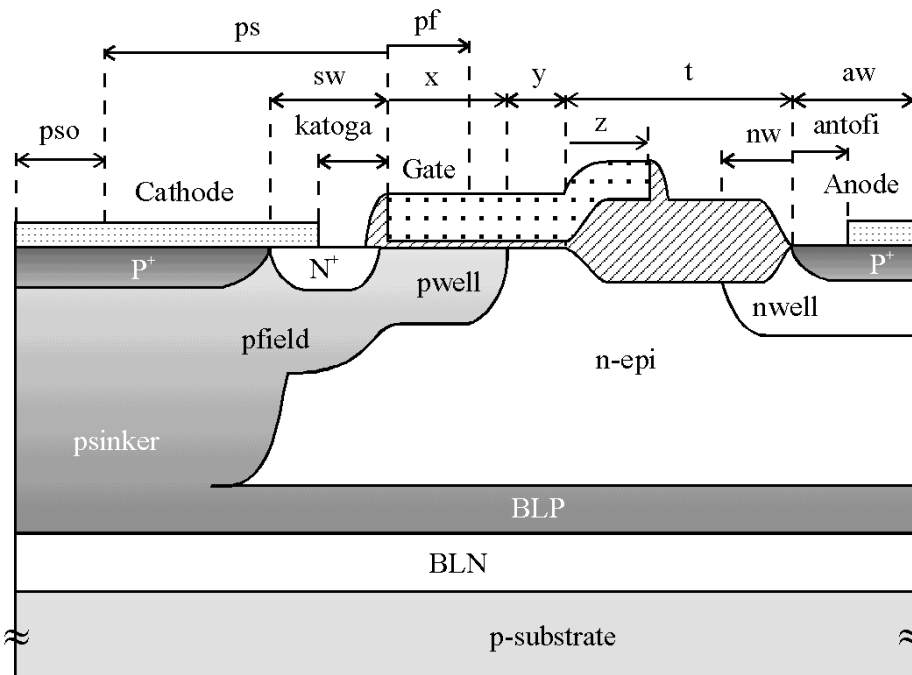


$$\Delta R_{on} = \frac{A.t^n}{1 + B.t^n} + C.t^m$$



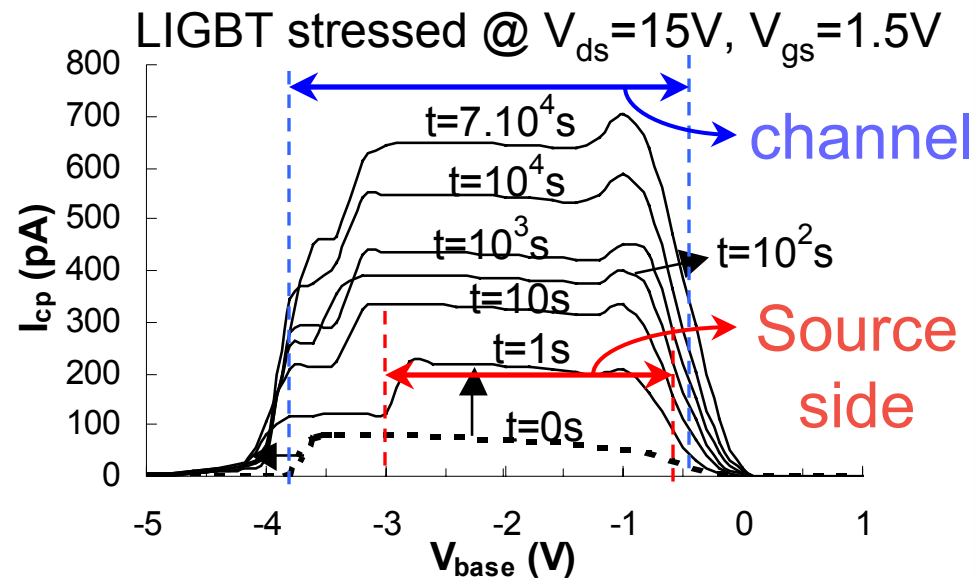
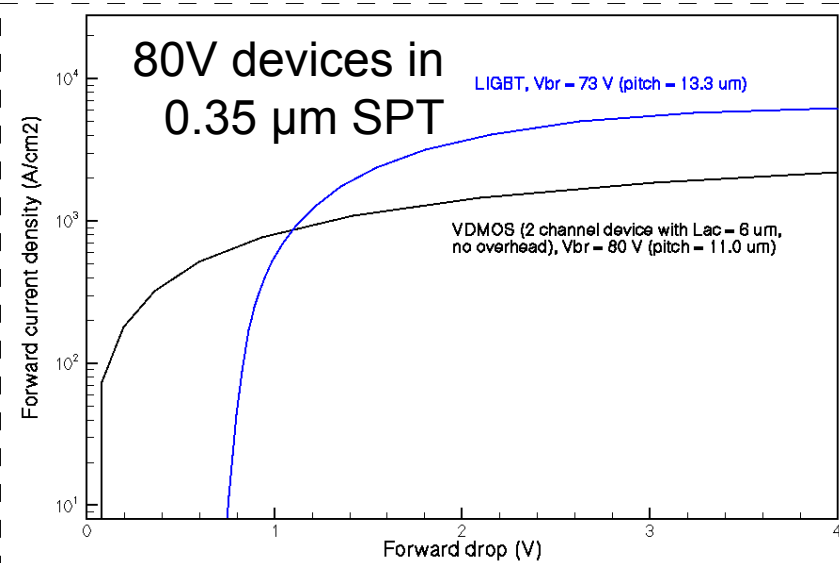
# *n*-type L-IGBT (1)

- *L*-IGBT in JI techno → substrate injection
- Use double isolation : BLN & BLP [45]



# *n*-type L-IGBT (2)

- *LIGBT has a much higher current density in on-state compared to MOS → power dissip*
- *During operation : injection of charges can be monitored with CP*

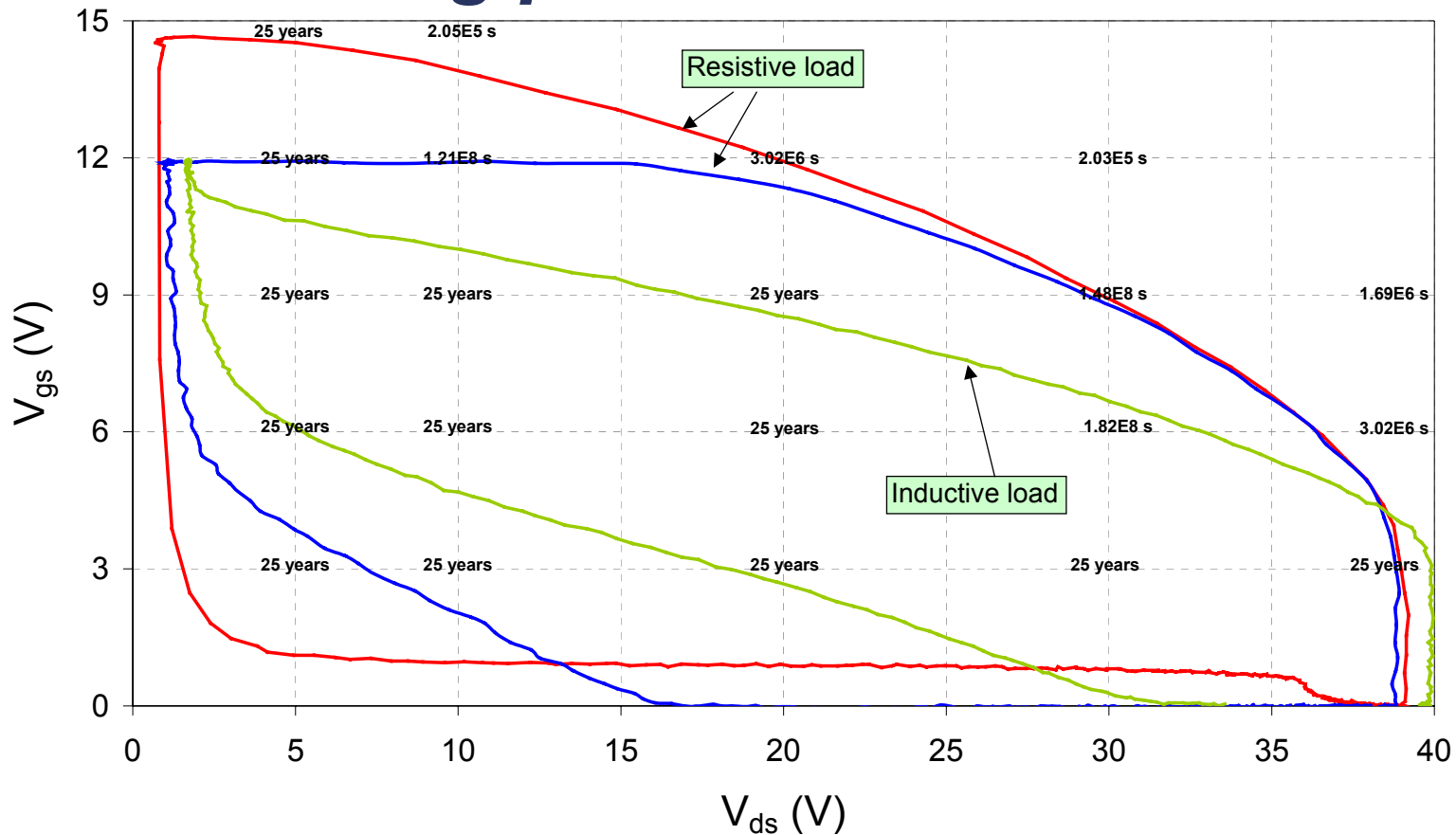


# ***Reliability Problems in Smart Power***

- *DC hot carrier*
- ***AC hot carrier (switching of a load)***
- *Electrical SOA (parasitic bipolar turn-on)*
- *Thermal SOA (energy capability)*
- *Simulation of thermal effects*

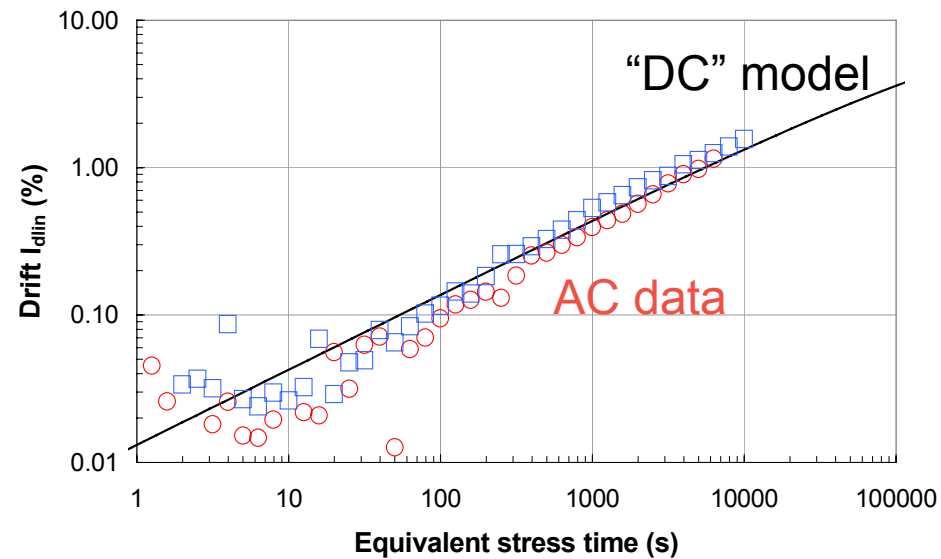
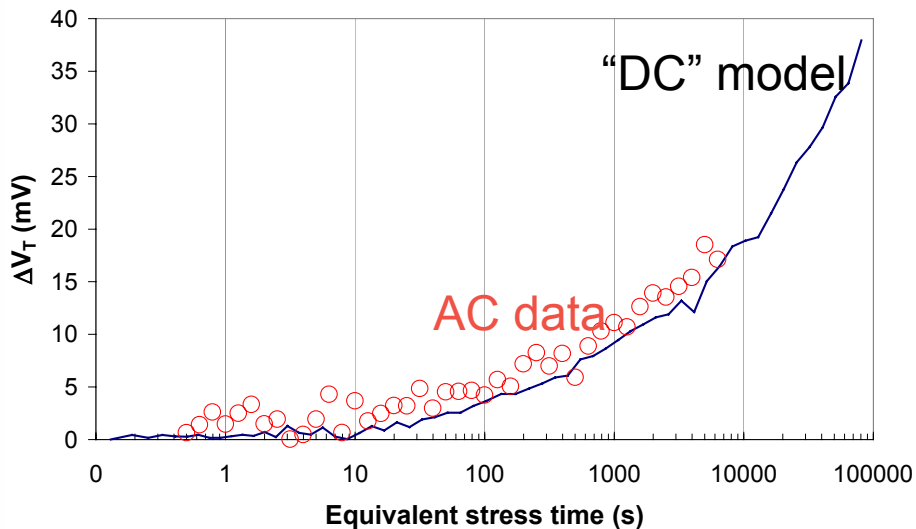
# Switching of a load-Link to DC (1)

- Calculate total equivalent stress time from AC switching path.



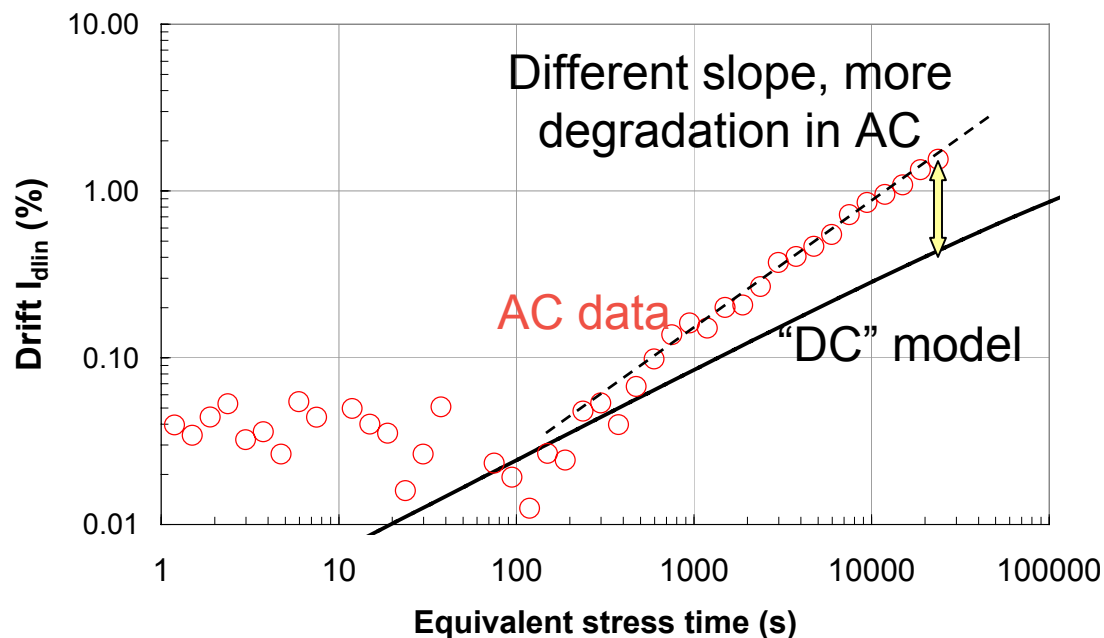
# Switching of a load-Link to DC (2)

- **Resistive load : good agreement AC-DC**
- **Problem : switching through regions where different parameters are degrading.**

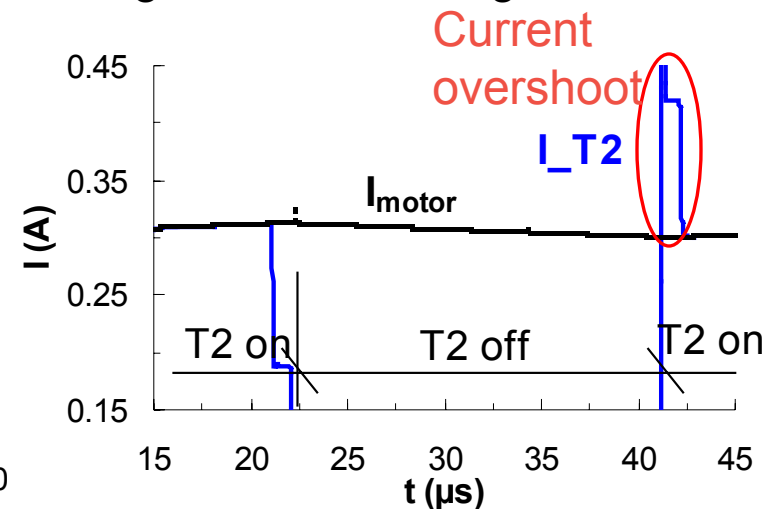


# Switching of a load-Link to DC (3)

- **Inductive load : deviation from “DC” model.**
- **Due to overshoots upon switching.**
- **Overshoots to be reduced by design/circuit.**



Low side switch of a H-bridge configuration switching a motor



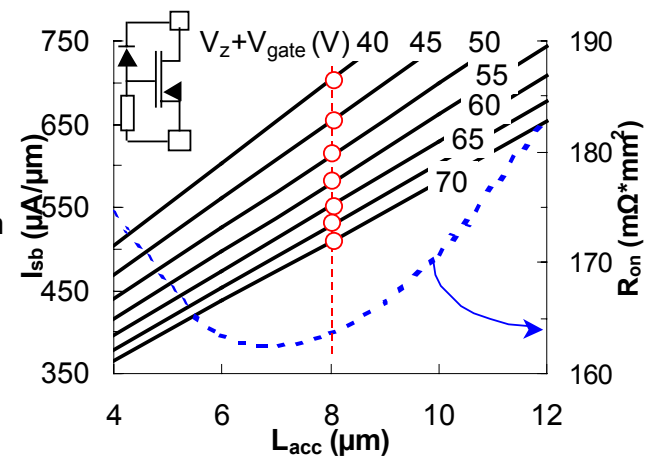
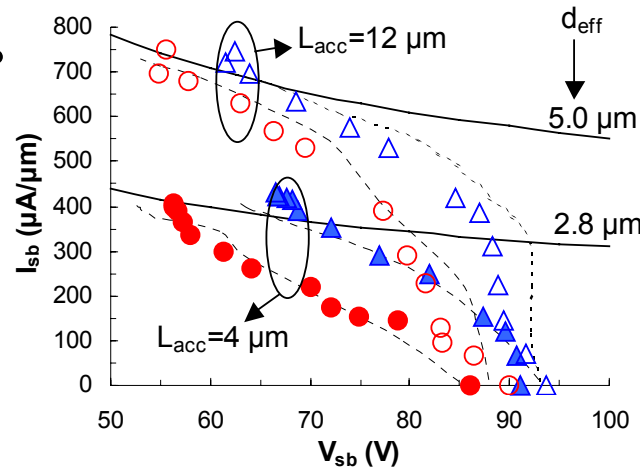
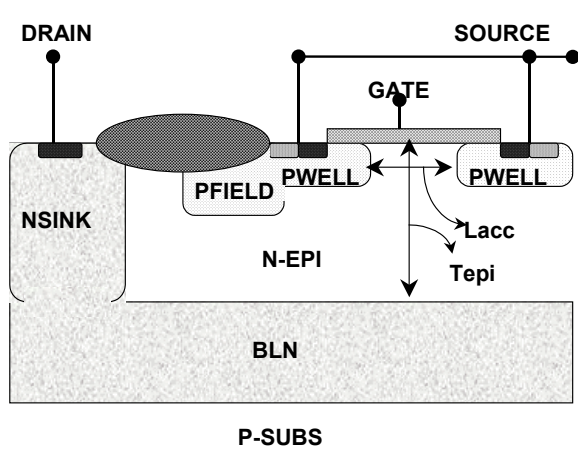
# ***Reliability Problems in Smart Power***

- *DC hot carrier*
- *AC hot carrier (switching of a load)*
- ***Electrical SOA (parasitic bipolar turn-on)***
- *Thermal SOA (energy capability)*
- *Simulation of thermal effects*

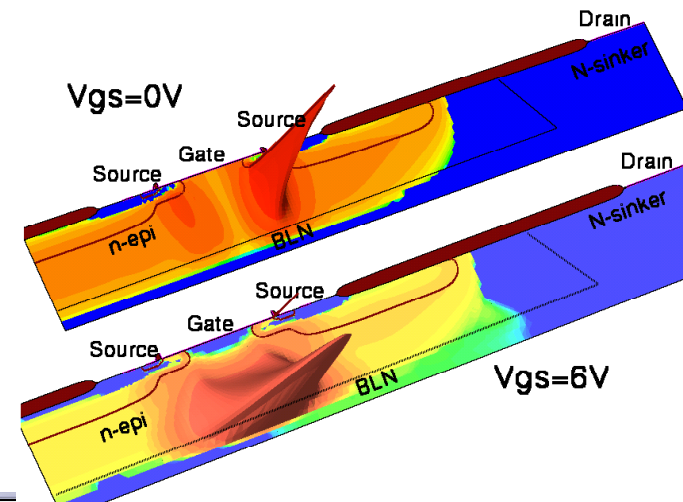


# Electrical SOA (2)

- **Electrical SOA VDMOS [55]**

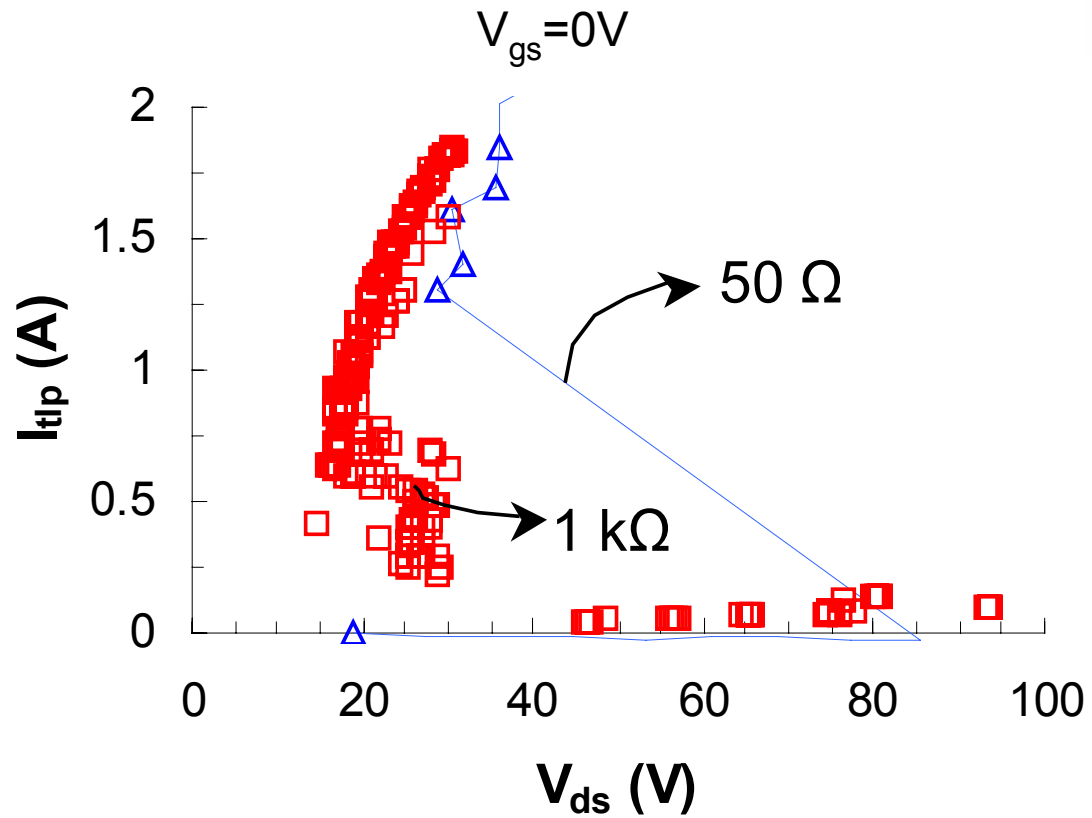
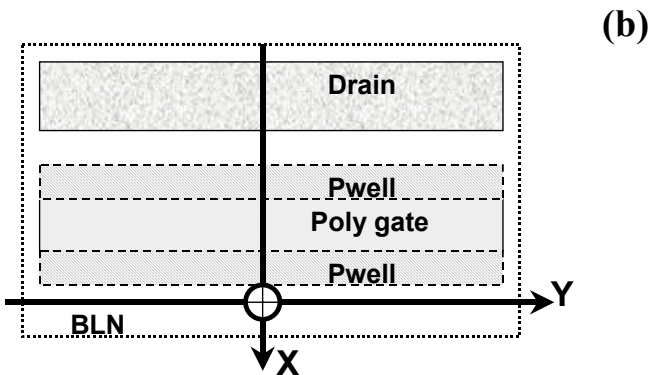
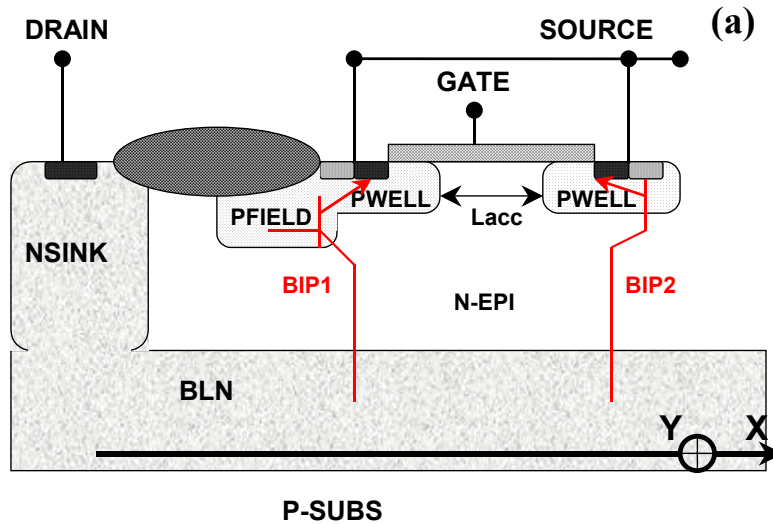


- **Flux tube model predicts well limiting  $I_{sb}$ .**
- **Trade-off  $I_{sb}$  -  $R_{on}$**



# TIM on VDMOS (1)

- Hot spot hopping in VDMOS [28,29]





# ***Reliability Problems in Smart Power***

- *DC hot carrier*
- *AC hot carrier (switching of a load)*
- *Electrical SOA (parasitic bipolar turn-on)*
- ***Thermal SOA (energy capability)***
- *Simulation of thermal effects*

# Thermal SOA (1)

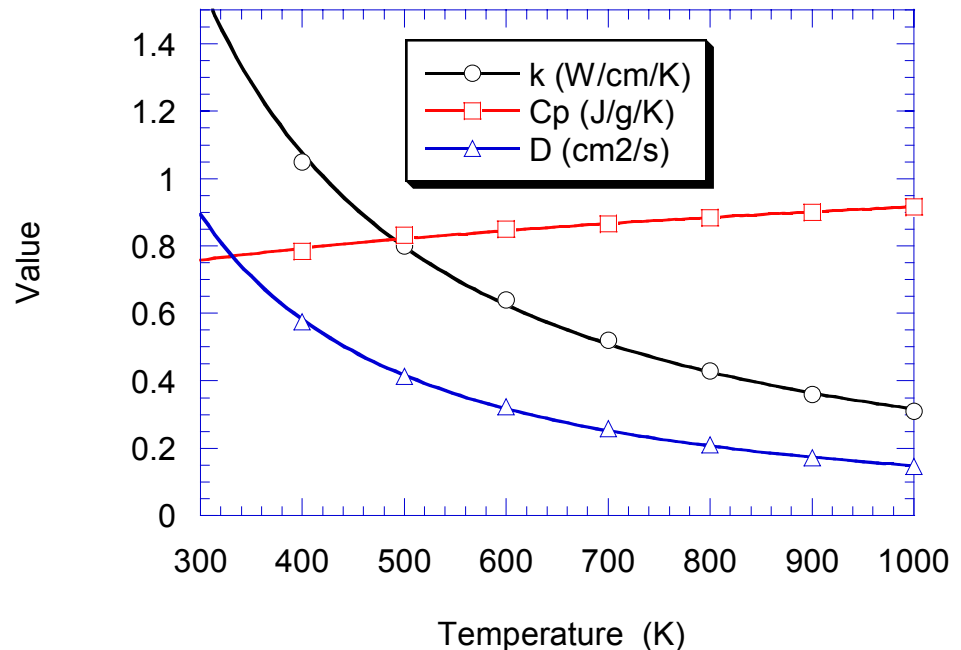
- **Wunsch-Bell model for a semi-infinite piece of silicon subject to rectangular power pulse  $P$**

$$P(t) = \frac{k\sqrt{\pi}\Delta T_{crit}}{\sqrt{4Dt}}; E(t) = P(t) * t \sim \sqrt{t}$$

Thermal conductivity  $k$  in W/cm.K

$D = k/\rho.C_p$ ,  $\rho$  being the density (g/cm<sup>3</sup>) and  $C_p$  the specific heat (J/gK)

$\Delta T_{crit}$  is the critical temperature at thermal failure.



# Thermal SOA (2)

- **Dwyer model for heat source of finite dimensions ( $a > b > c$ )**

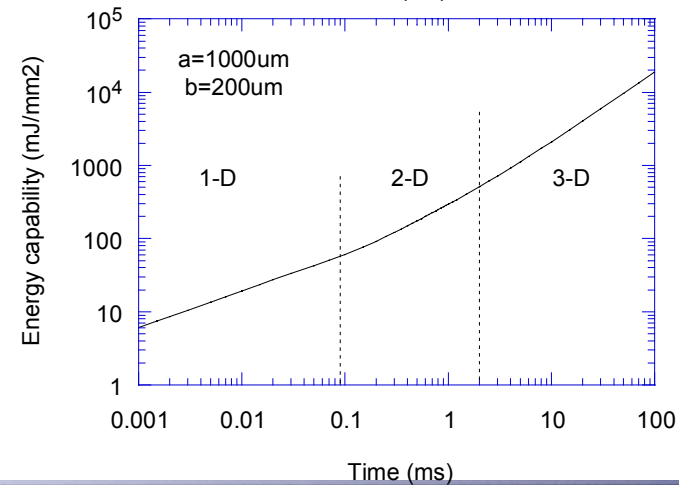
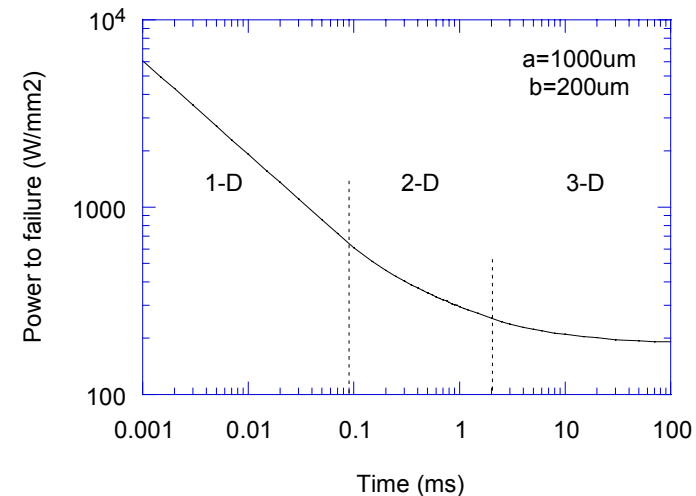
[31,41]

$$P_{1D} = \frac{k\sqrt{\pi}\Delta T_{crit}}{\sqrt{4D}\left(\sqrt{t} - \frac{\sqrt{t_c}}{2}\right)}, \quad t_c < t \leq t_b = \frac{b^2}{4\pi D}$$

$$P_{2D} = \frac{2k\pi\Delta T_{crit}}{b\left[\ln\left(\frac{t}{t_b}\right) + 2 - \frac{c}{b}\right]}, \quad t_b \leq t \leq t_a = \frac{a^2}{4\pi D}$$

$$P_{3D} = \frac{k\pi\Delta T_{crit}}{b\left[\ln\left(\frac{a}{b}\right) + 2 - \frac{c}{2b} - \sqrt{\frac{t_a}{t}}\right]}, \quad t_a \leq t$$

$$P_{ss} = \frac{k\pi\Delta T_{crit}}{b\left[\ln\left(\frac{a}{b}\right) + 2 - \frac{c}{2b}\right]}, \quad t \rightarrow \infty$$

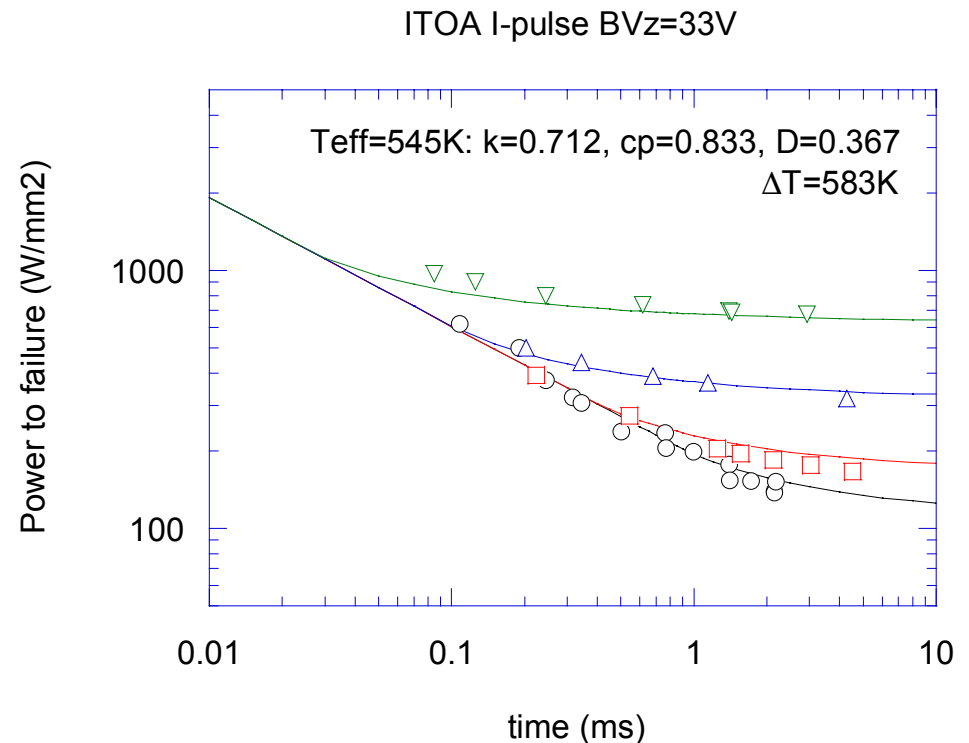


# Thermal SOA (3)

- **Motivation : construct thermal SOA of 40V square power drivers of different size.**
- **Measurement data and model fitting.**

- $D$ ,  $k$  and  $\Delta T_{\text{crit}}$  are not independent
- $K$  and  $D$  are dependent on  $T_{\text{eff}}$
- $T_{\text{eff}}$  and  $\Delta T_{\text{crit}}$  are to be fitted to the data

$\Delta T_{\text{crit}}$  (~580K) is rather large. From simulations and measurements values between 400-500 are found [39,40]



# Thermal SOA (4)

- **Dependency of  $P_{fail}$  on  $V_{ds}$** 
  - ◆ Pure thermal ?
  - ◆ Electro-thermal coupling ?
  - ◆ Data for constant time-to-failure ? → Normalize the data to a fixed time (e.g. 1 ms).

$$P_{3D}(t^*) = P_{1D}(t) \frac{\sqrt{4\pi Dt}}{b \left[ \ln\left(\frac{a}{b}\right) + 2 - \sqrt{\frac{t_a}{t^*}} \right]}, \quad t \leq t_b$$

$$P_{3D}(t^*) = P_{2D}(t) \frac{\left[ \ln\left(\frac{t}{t_b}\right) + 2 \right]}{2 \left[ \ln\left(\frac{a}{b}\right) + 2 - \sqrt{\frac{t_a}{t^*}} \right]}, \quad t_b \leq t \leq t_a$$

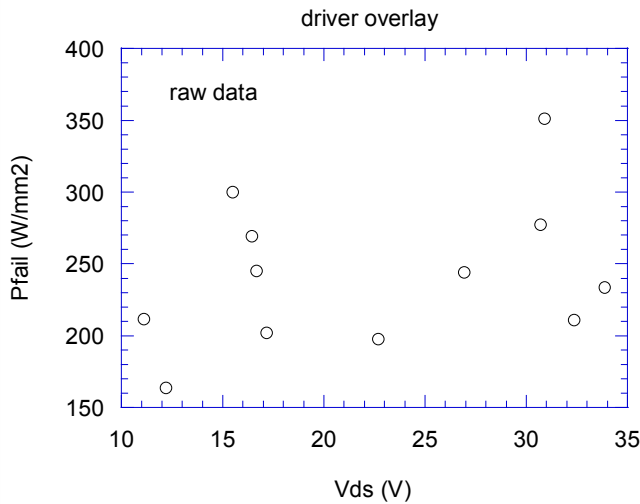
$$P_{3D}(t^*) = P_{3D}(t) \frac{\left[ \ln\left(\frac{a}{b}\right) + 2 - \sqrt{\frac{t_a}{t}} \right]}{\left[ \ln\left(\frac{a}{b}\right) + 2 - \sqrt{\frac{t_a}{t^*}} \right]}, \quad t_a \leq t$$

Only dependent on  $D$ , not on  $\Delta T_{crit}$ .

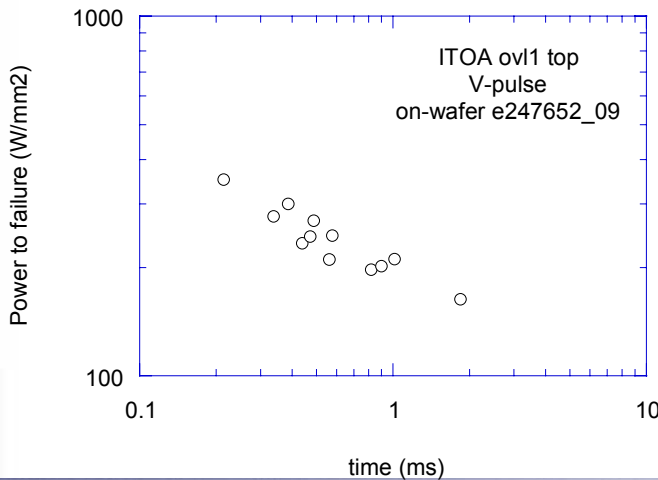
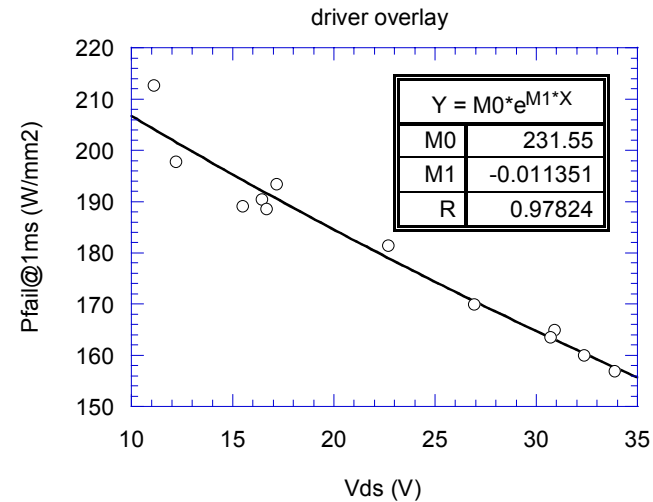
$D$  assumed to be independent of  $V_{ds}$ .

# Thermal SOA (5)

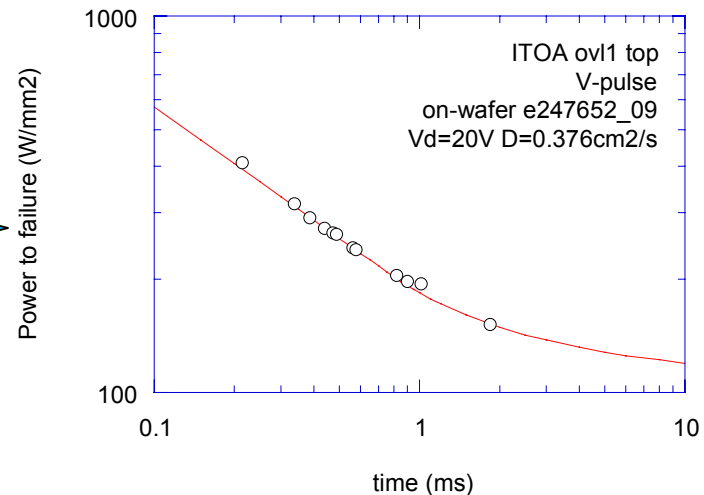
## Raw data



normalized to  
 $t^* = 1\text{ms}$

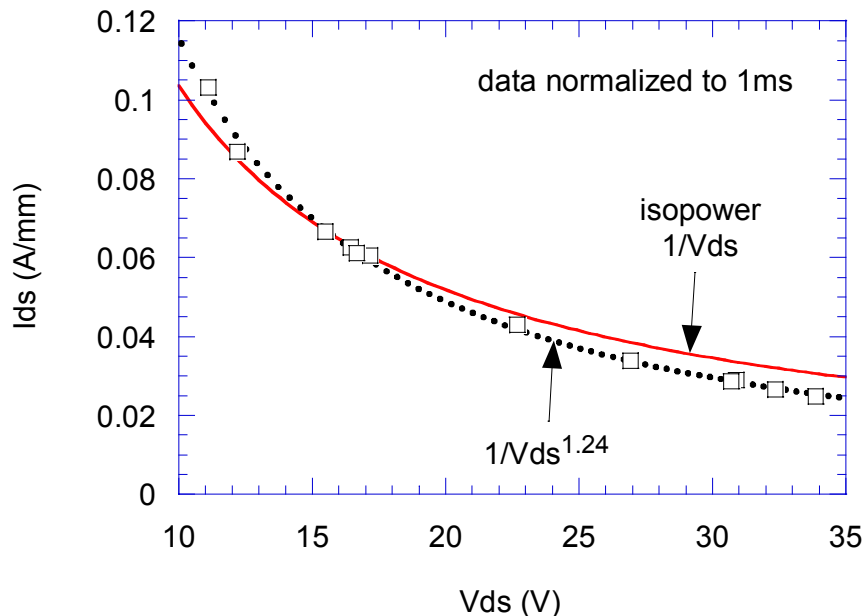


normalized  
to  $V_{ds} = 20\text{V}$



## • Thermal SOA

- ◆ Plot equi- $t_{\text{fail}}$  lines in  $V_{\text{gs}}-V_{\text{ds}}$  or  $I_{\text{ds}}-V_{\text{ds}}$  plot.
- ◆ Problem :  $t_{\text{fail}}$  is not an independent variable.
- ◆ Use normalization procedure.



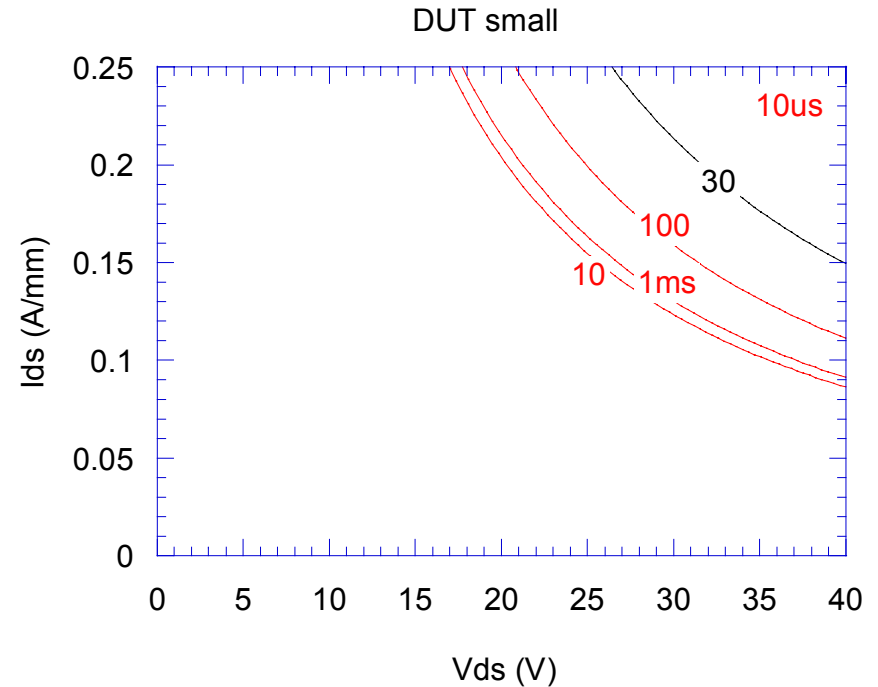
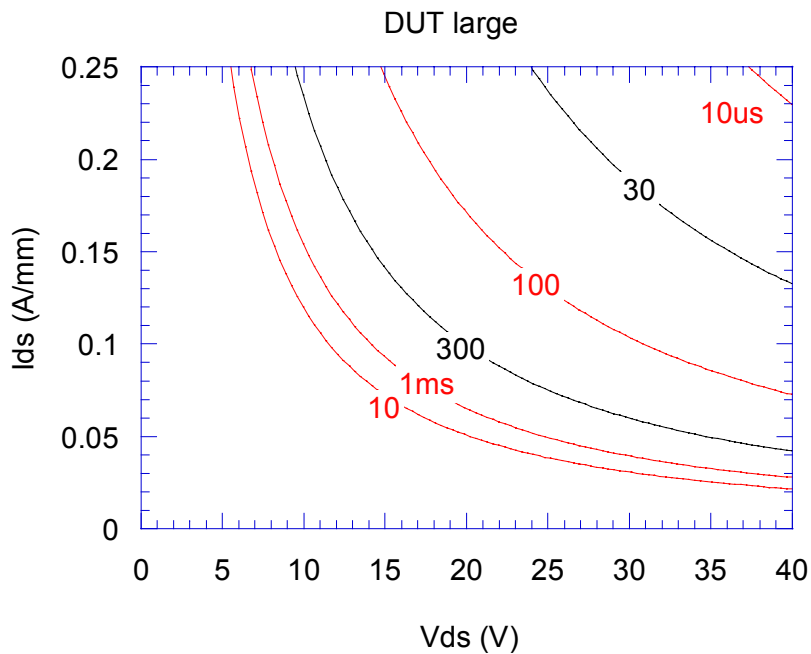
A generalized fit  $I_{\text{ds}}(V_{\text{ds}}) \sim 1/V_{\text{ds}}^{1.24}$  yields the best results, consistent with electro-thermal coupling.

Cross-checked on other sets of data.

Similar results obtained by Krabbenborg et al. [35]

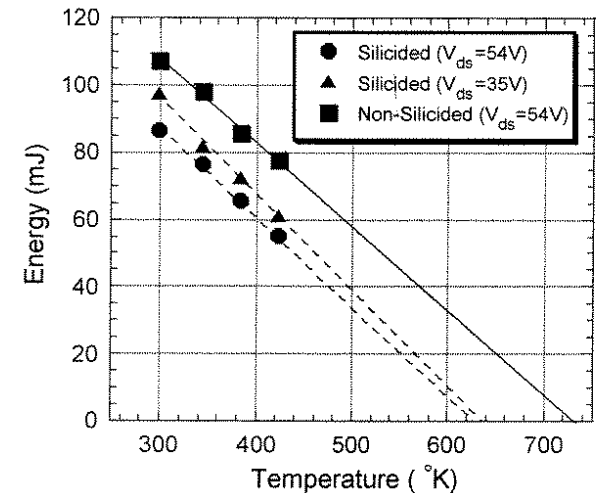
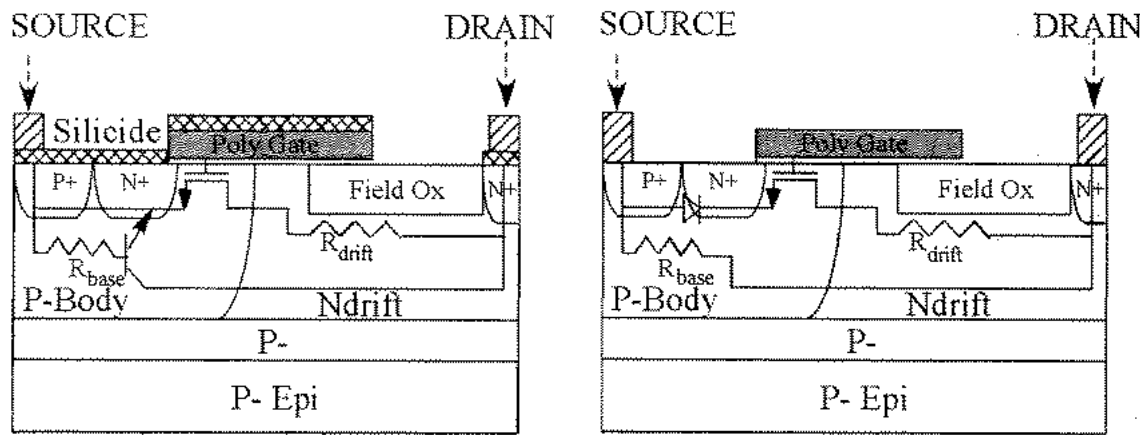
# Thermal SOA (7)

- **Thermal SOA :  $equi-t_{fail}$  as obtained from the Dwyer model.**



# Thermal SOA (8)

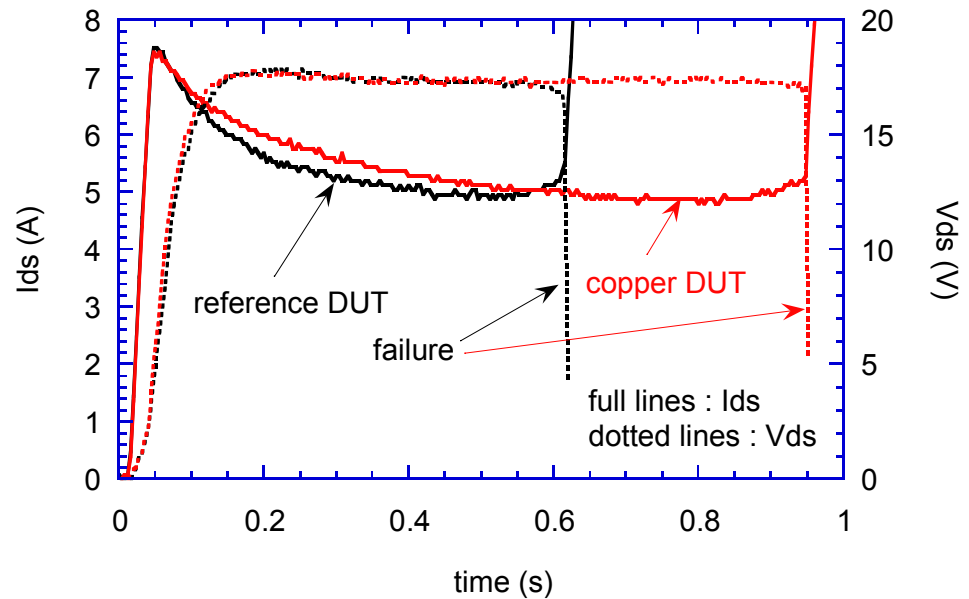
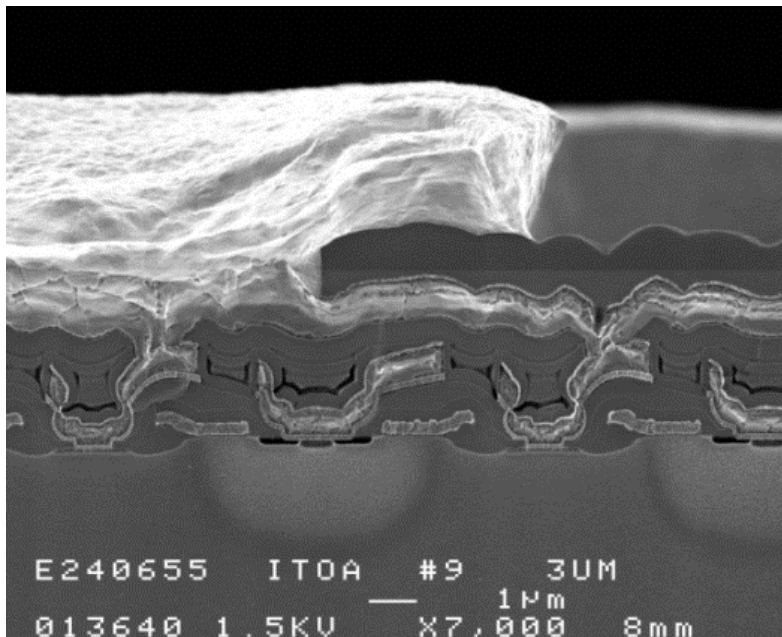
- **Electro-thermal interaction : de-activate parasitic NPN [40] to isolate pure thermal effects.**



- **Non silicided device yields  $T_{crit}=730 K$**
- **Silicided device :  $T_{crit}=630 K$ , due to electro-thermal coupling.**

# Thermal SOA (9)

- ***In smart power, an electroplated top Cu layer (3-25  $\mu\text{m}$ ) is often used :***
  - ◆ To reduce the metal resistance [32]
  - ◆ As a heat sink/spreader to reduce  $T_{\text{max}}$  [38,41]



# *Reliability Problems in Smart Power*

- *DC hot carrier*
- *AC hot carrier (switching of a load)*
- *Electrical SOA (parasitic bipolar turn-on)*
- *Thermal SOA (energy capability)*
- ***Simulation of thermal effects***

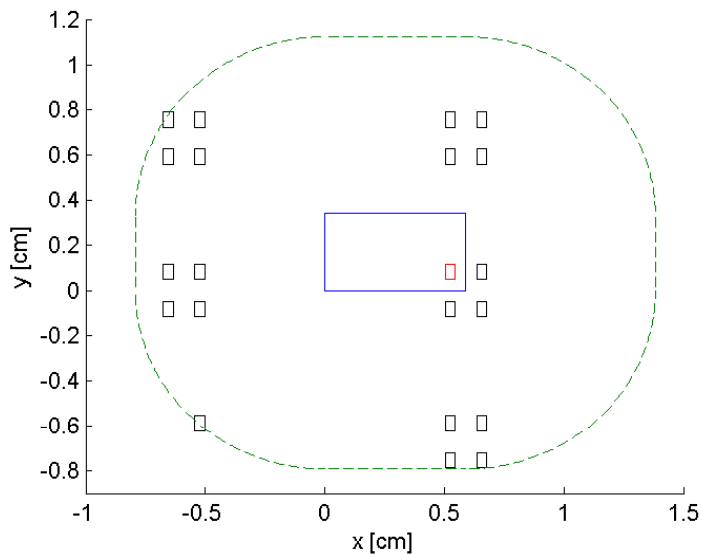
# *Thermal simulations (1)*

- ***Need for a fast thermal simulation tool as a function of pulse time, bias conditions, device area and chip size.***
  - ◆ Green's function solution of the heat diffusion equation [56].
  - ◆ Adiabatic boundary conditions using the method of images [57].
  - ◆ Thermal boundaries defined by a typical thermal diffusion length.
- ***Calculation of  $T(x,y,z,t)$  for any number of power sources and for any power pulse  $P(t)$***

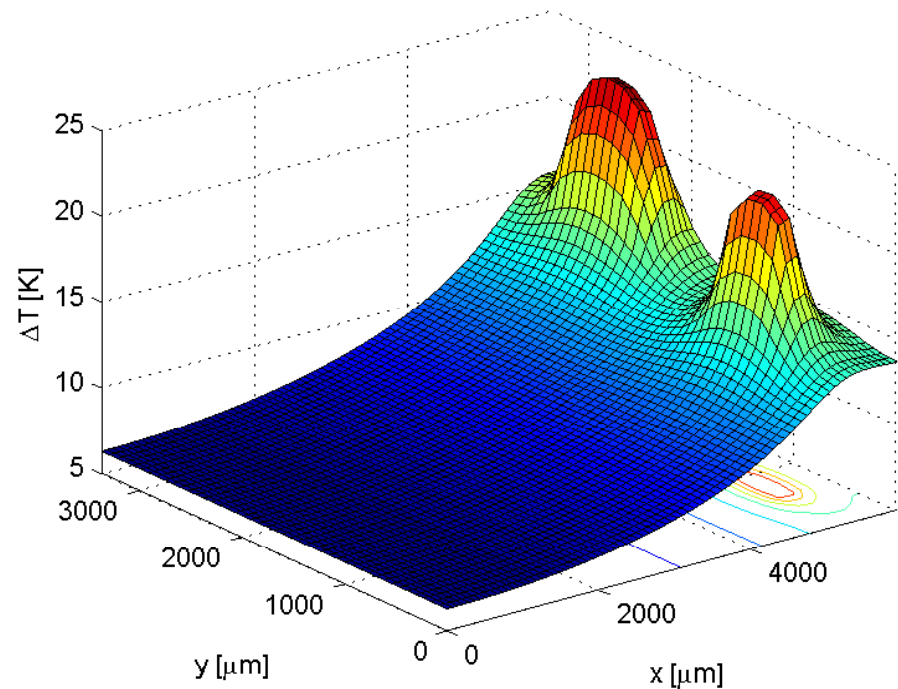
# Thermal simulations (2)

- **Examples of the model (1)**

Images and thermal boundary (100 ms)



VMCA temperature rise,  $P = 2.3 \text{ W}$ ,  $t = 1 \text{ s}$ ,  $\Delta T_{\text{sensor}} = 6.43 \text{ K}$

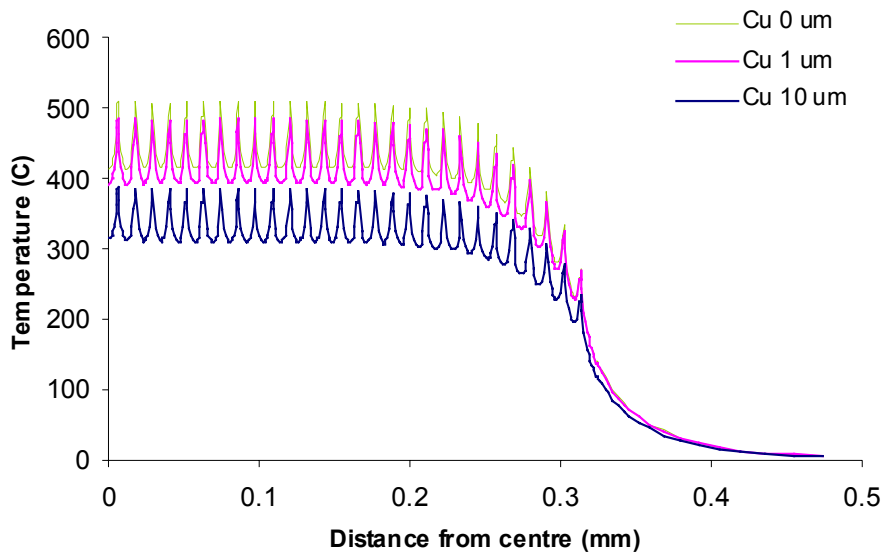


Temperature rise after 1s for 2 DMOS in a H-bridge, each dissipating 2.3 W

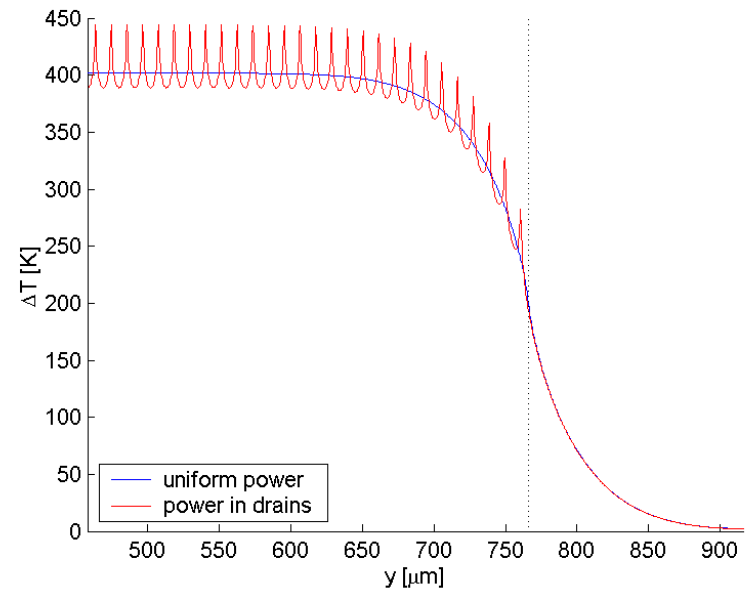
# Thermal simulations (3)

- Numerical simulations: **MSC.Marc** software
- Good **agreement** for small Cu thickness

commercial software



developed tool



# Thermal simulations (4)

- **Accuracy of the model : temperature rise in the center of a DMOS, for a 2 W power step function.**

- ◆ Flotherm : 24 hours

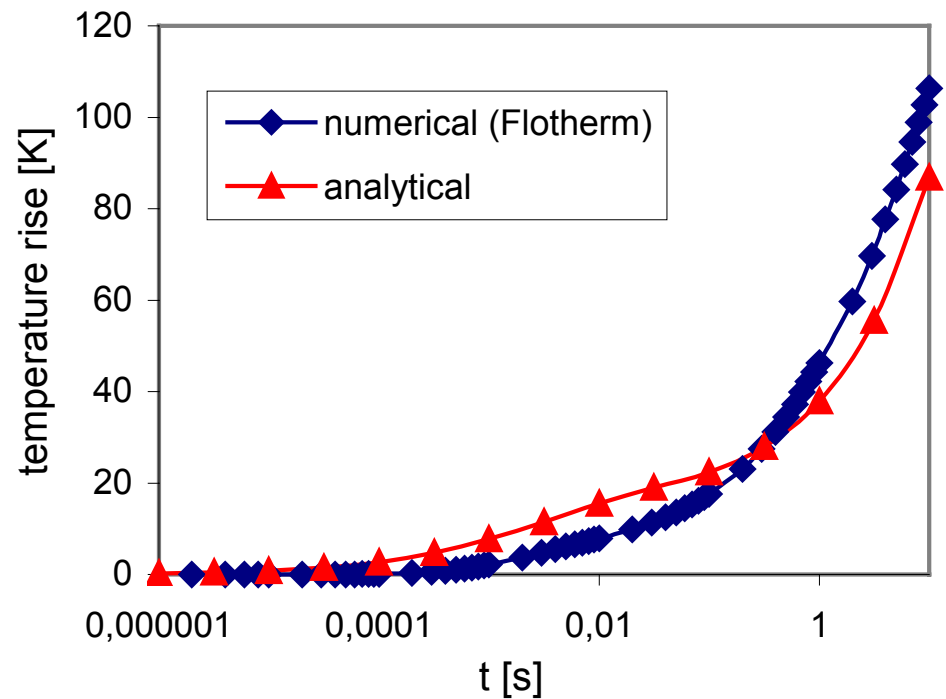
- ◆ Matlab : 20 min

- **Model is :**

- ◆ Fast

- ◆ Reliable

- ◆ Flexible to use

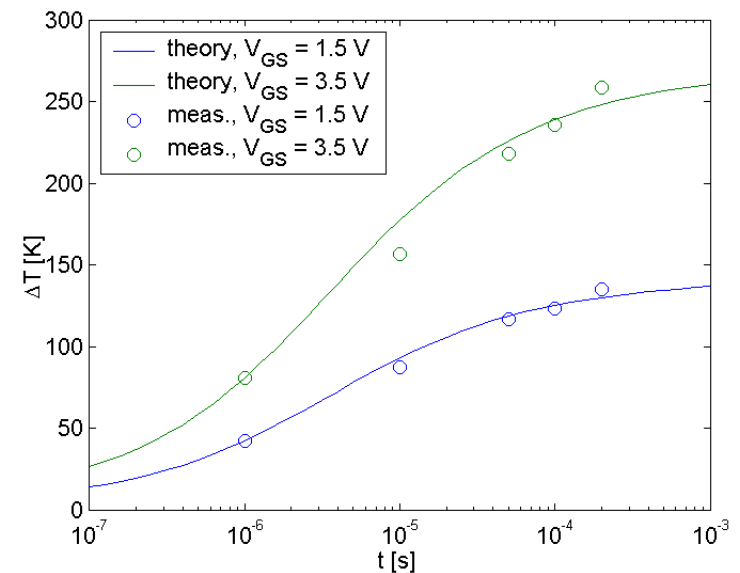
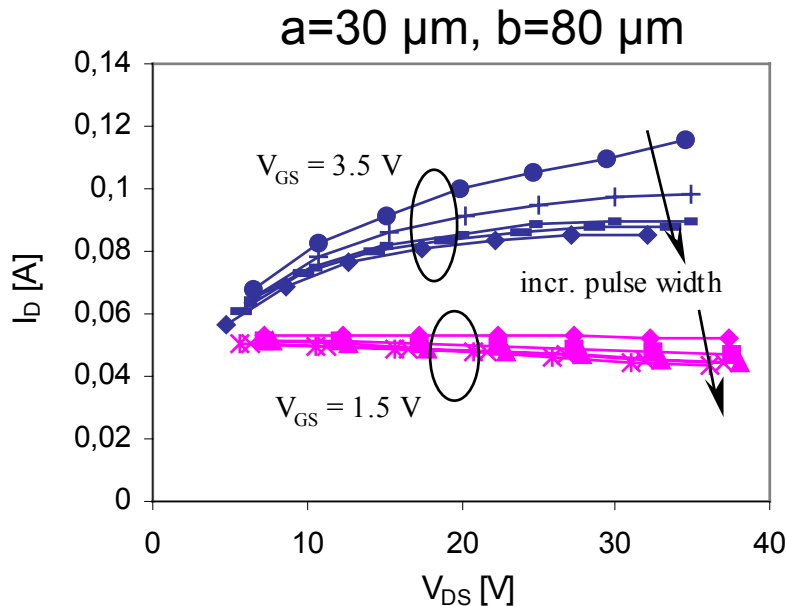


# Thermal simulations (5)

- **Experimental validation of the model :**

- ◆ Mobility reduction [58] :  $I_D / I_{D0} = (T / T_0)^{-k}$

- ◆ Integrated temperature sensors [59,39]

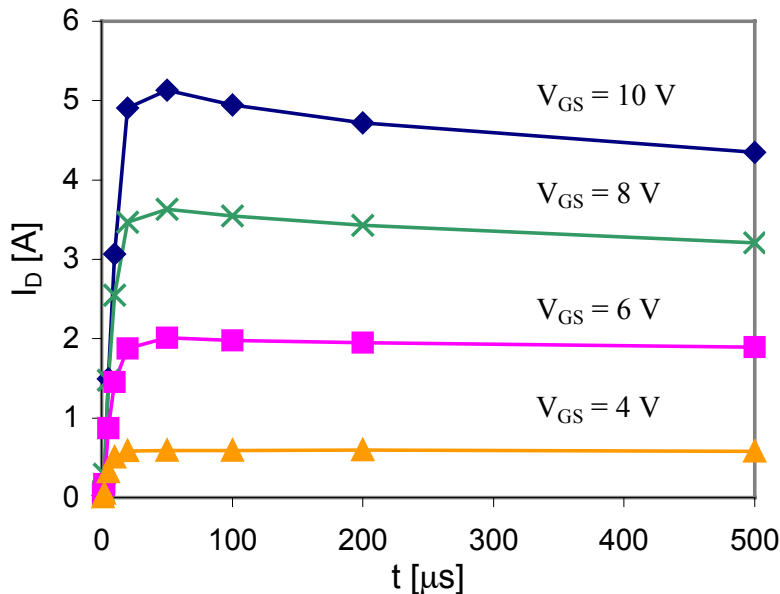


- Works reasonably well for relatively small devices.

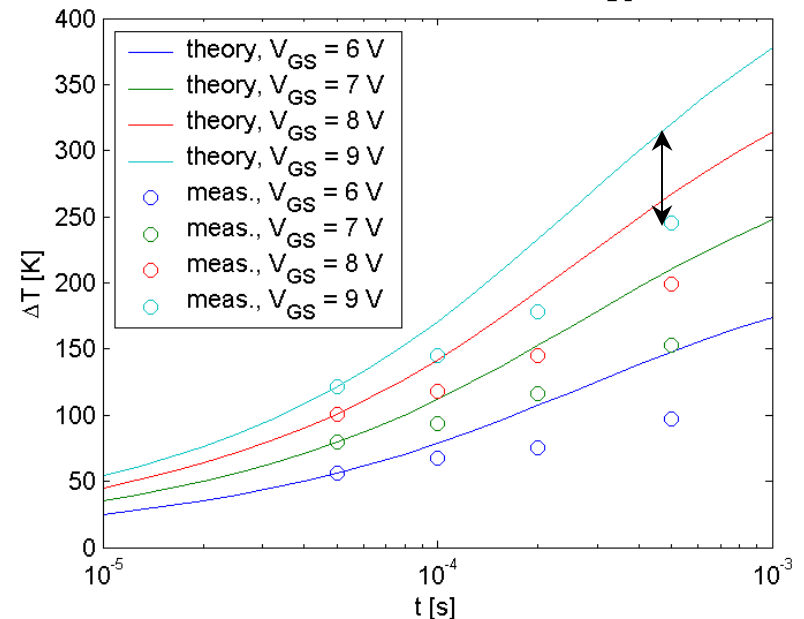
# Thermal simulations (6)

- **For large drivers : temperature seems to be overestimated.**
- **Need for electro-thermal coupling.**

$V_{ds}=10V$ ,  $a=b=450\ \mu m$



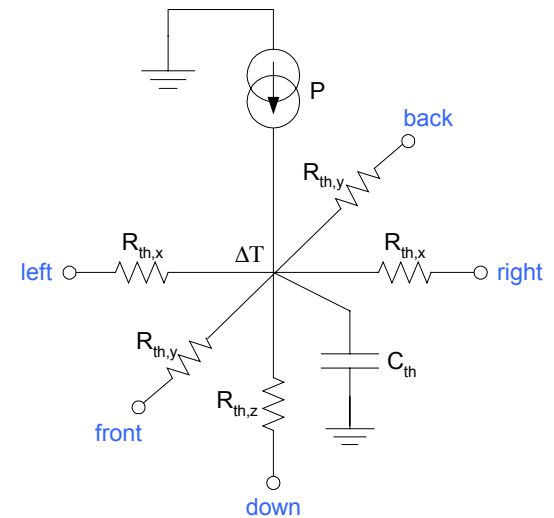
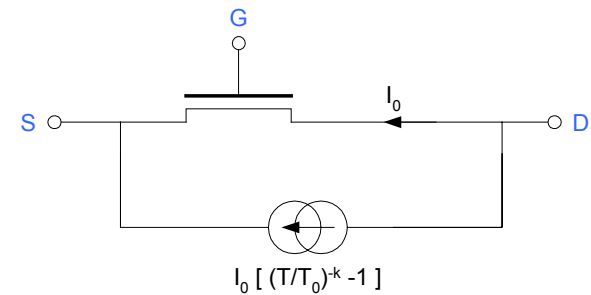
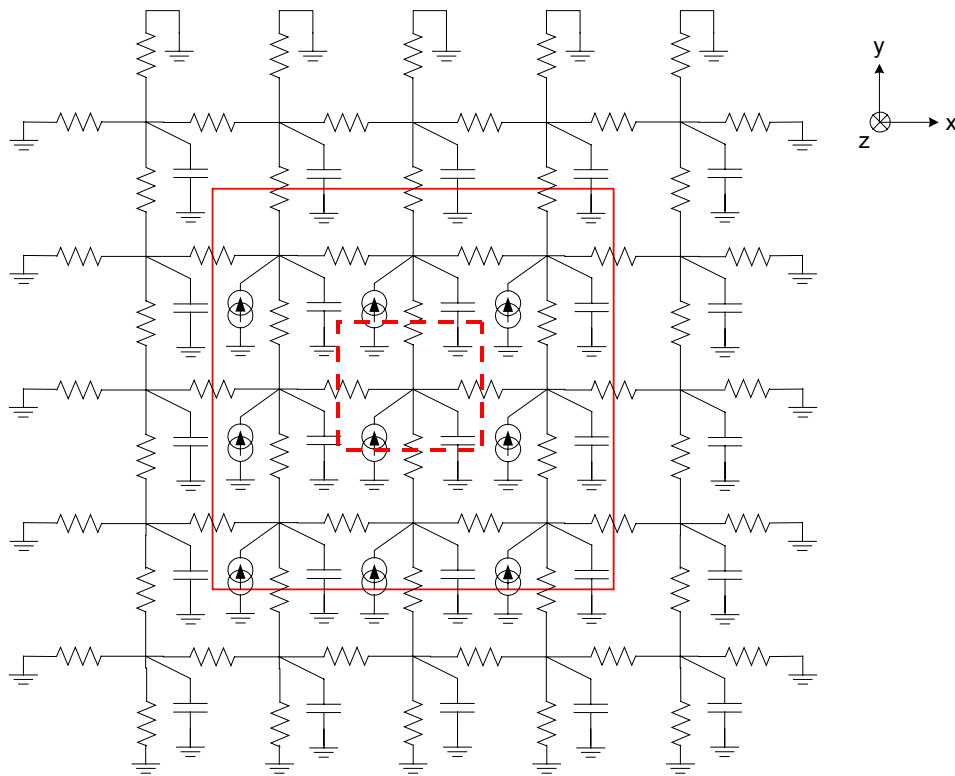
FND40MultRef large,  $W = L = 450\ \mu m$ ,  $V_{DS} = 10V$



# Thermal simulations (7)

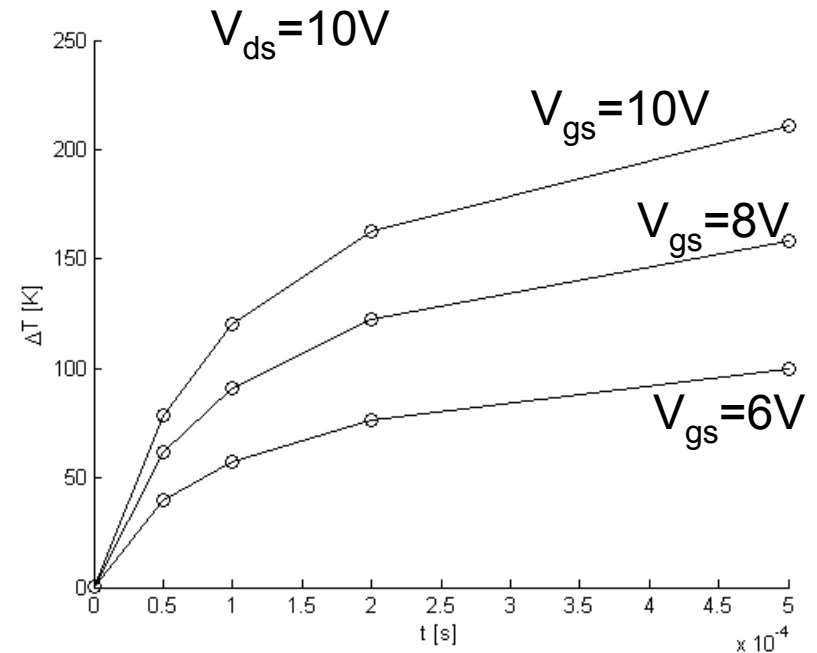
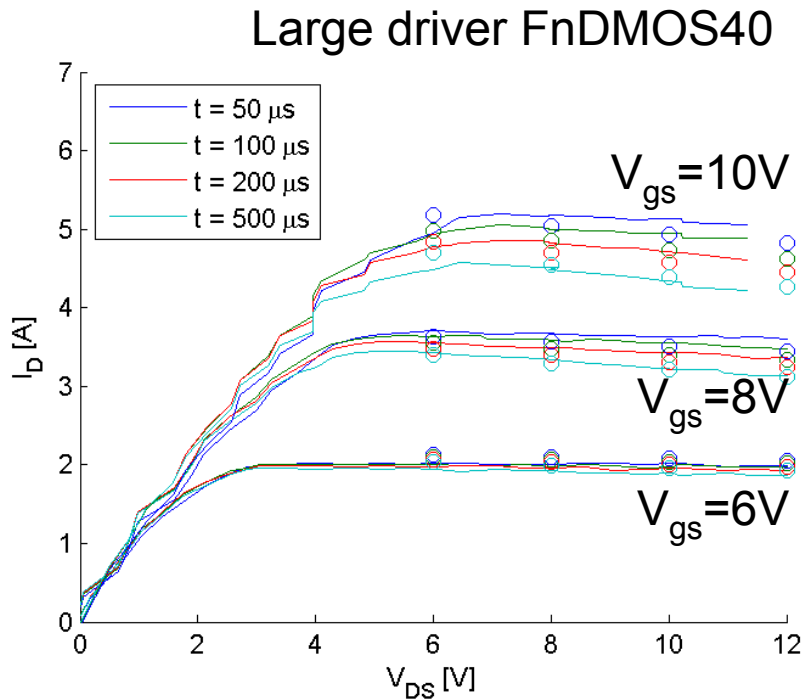
- **Electro-thermal coupling : discretization**

Topview thermal network



# Thermal simulations (8)

- **Current and temperature versus pulse time**

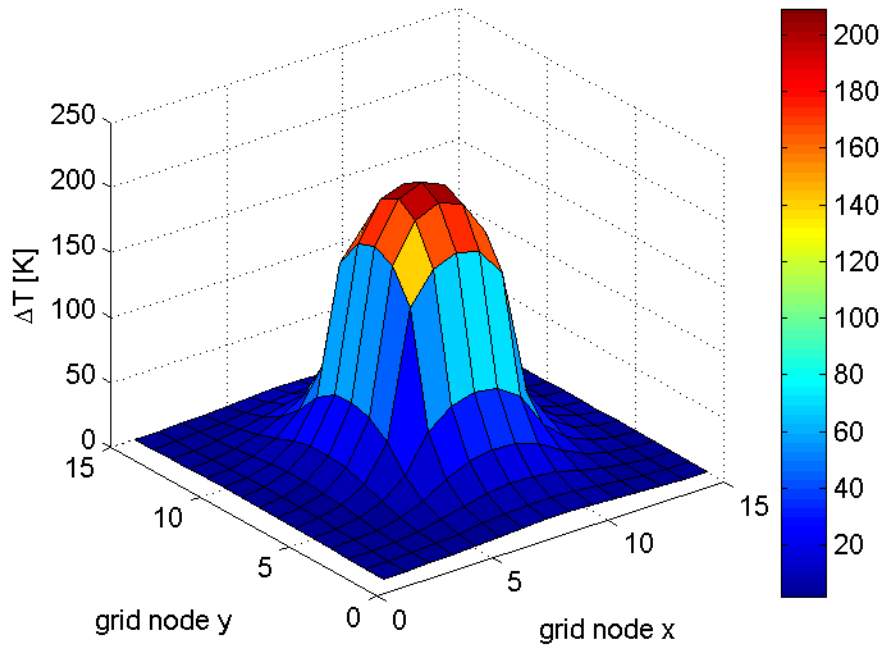


- **Electro-thermal model yields good agreement with measurements**

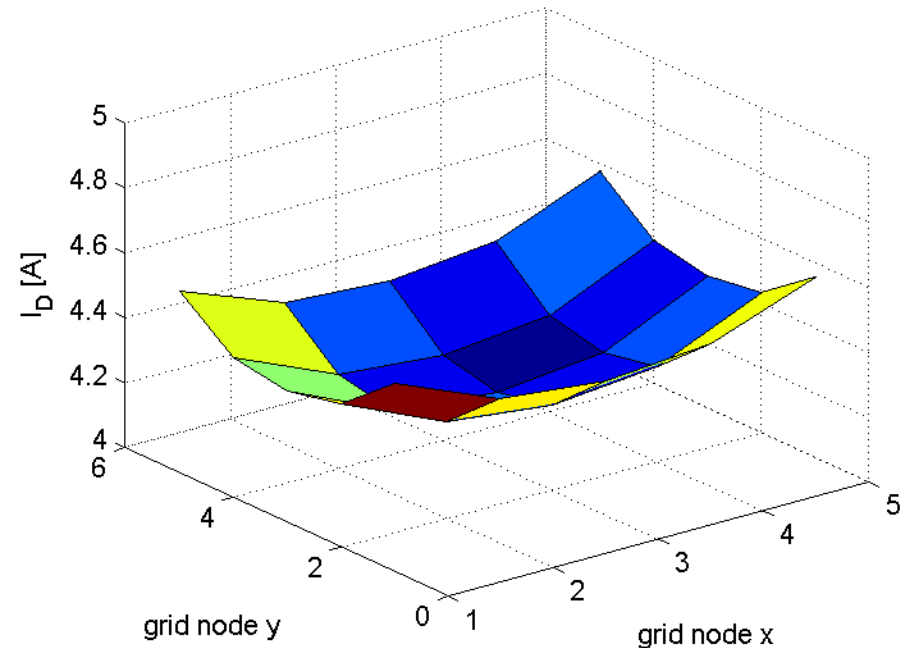
# Thermal simulations (9)

- 3D temperature and current distribution

FND40, large driver, temperature rise after 500  $\mu$ s



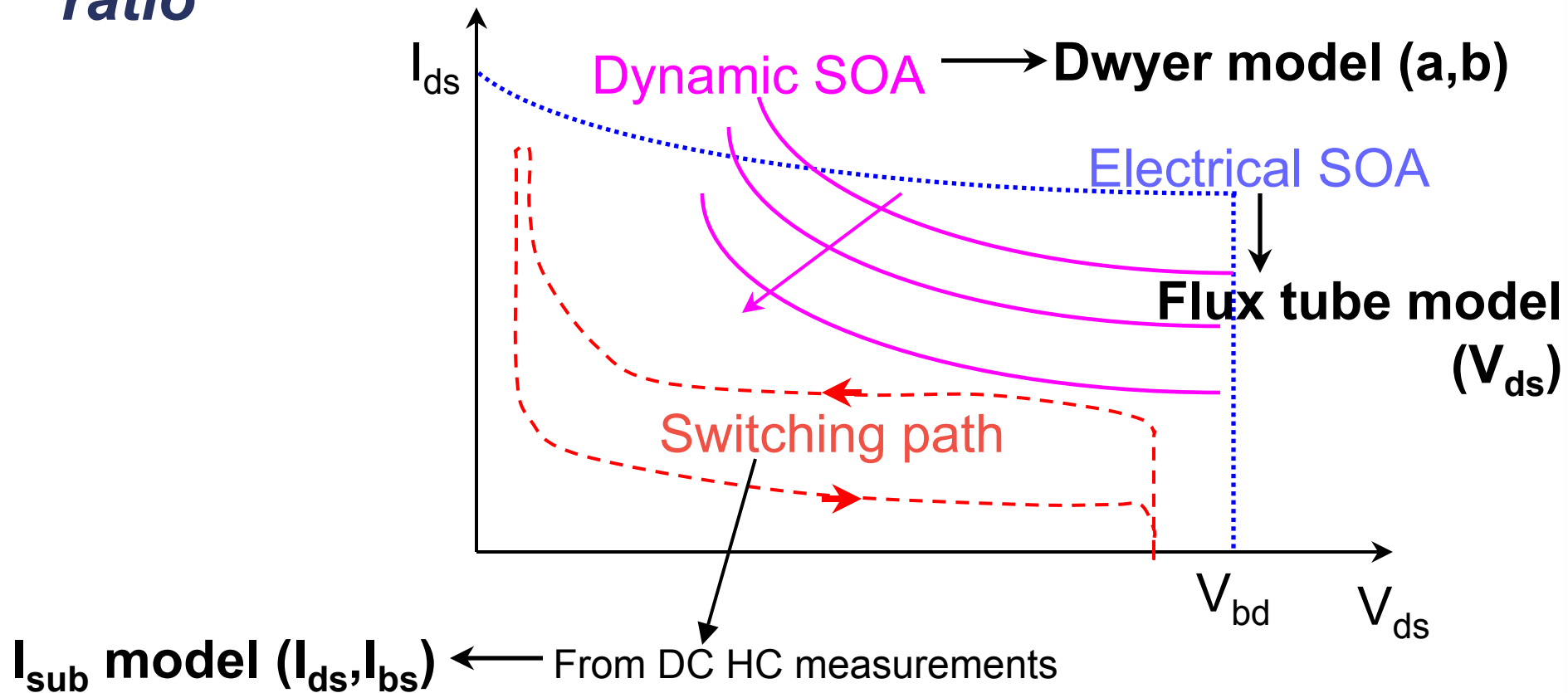
FND40, large driver, drain current after 500  $\mu$ s



- *Introduction*
- *Q&R Specifications*
- *Device Overview*
- *Reliability Measurements and Analysis Techniques*
- *Reliability Problems in Smart Power Technologies*
- **Conclusions**

# Conclusion : Total SOA (1)

- **Electrical and DC SOA are independent of  $W$**
- **Dynamic SOA is dependent on area and aspect ratio**



## Conclusion : Total SOA (2)

- **Calculate static and AC HC degradation for a given device. Device must always operate within SOA.**
- **Thermal SOA will determine  $\Delta T_{device}$ , to be calculated using electro-thermal model.  $\Delta T_{device}$  must be within technology specifications ( $T_{junction} = T_{ambient} + \Delta T_{device}$ ).  $\Delta T_{device}$  is largely dependent on the application.**

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