

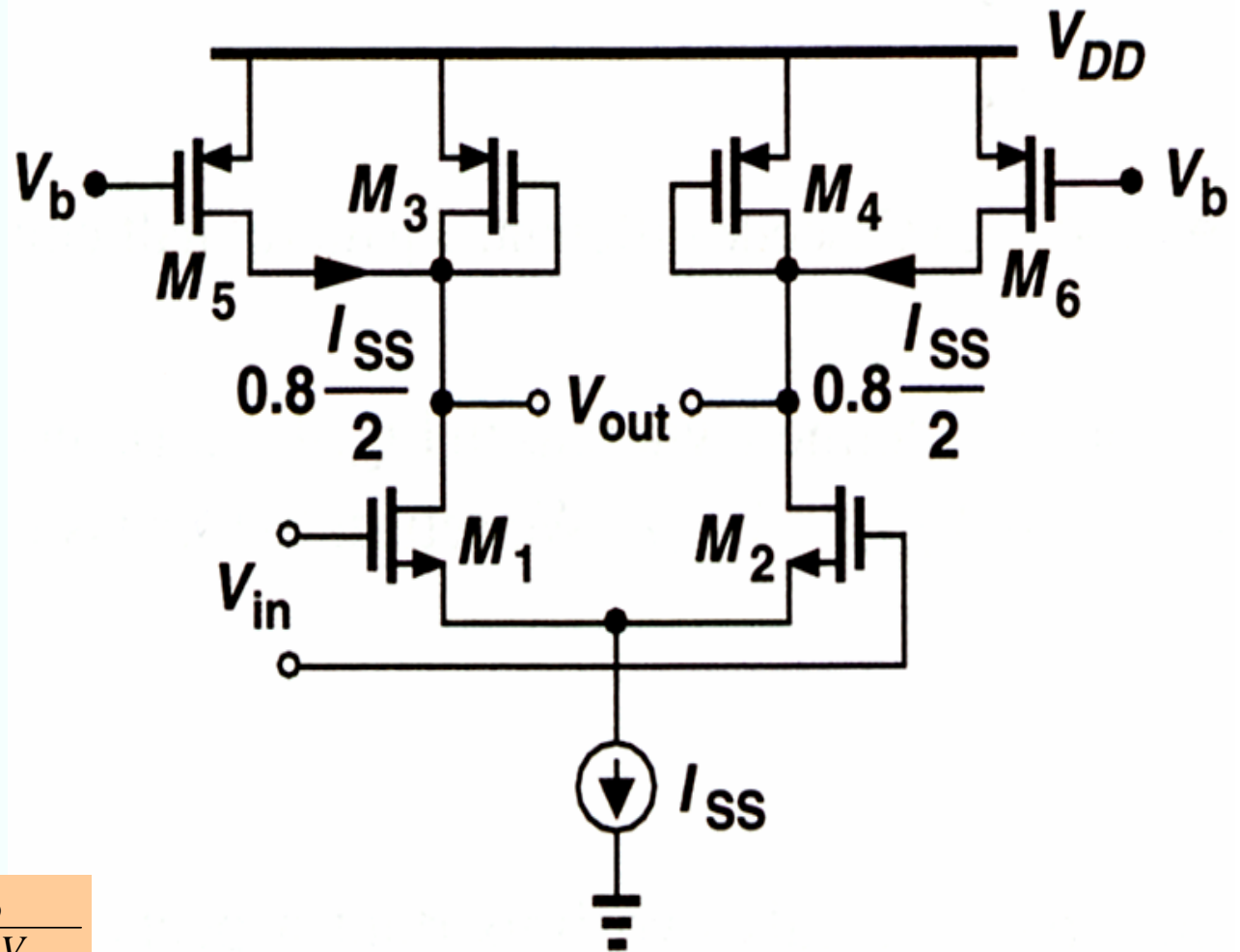
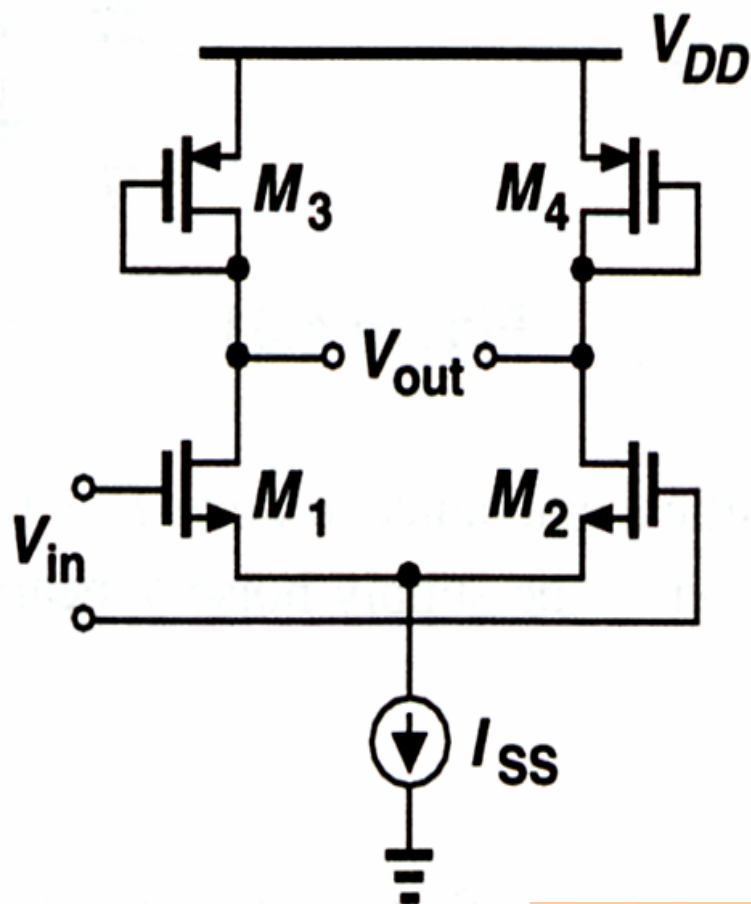
$$A_v = -g_{mN} (g_{mP}^{-1} \parallel r_{ON} \parallel r_{OP})$$

$$\approx -\frac{g_{mN}}{g_{mP}}$$

$$g_m = \sqrt{2\mu C_{ox} \cdot \frac{W}{L} \cdot I_D} = \frac{I_D}{\frac{V_{GS} - V_{TH}}{2}}$$

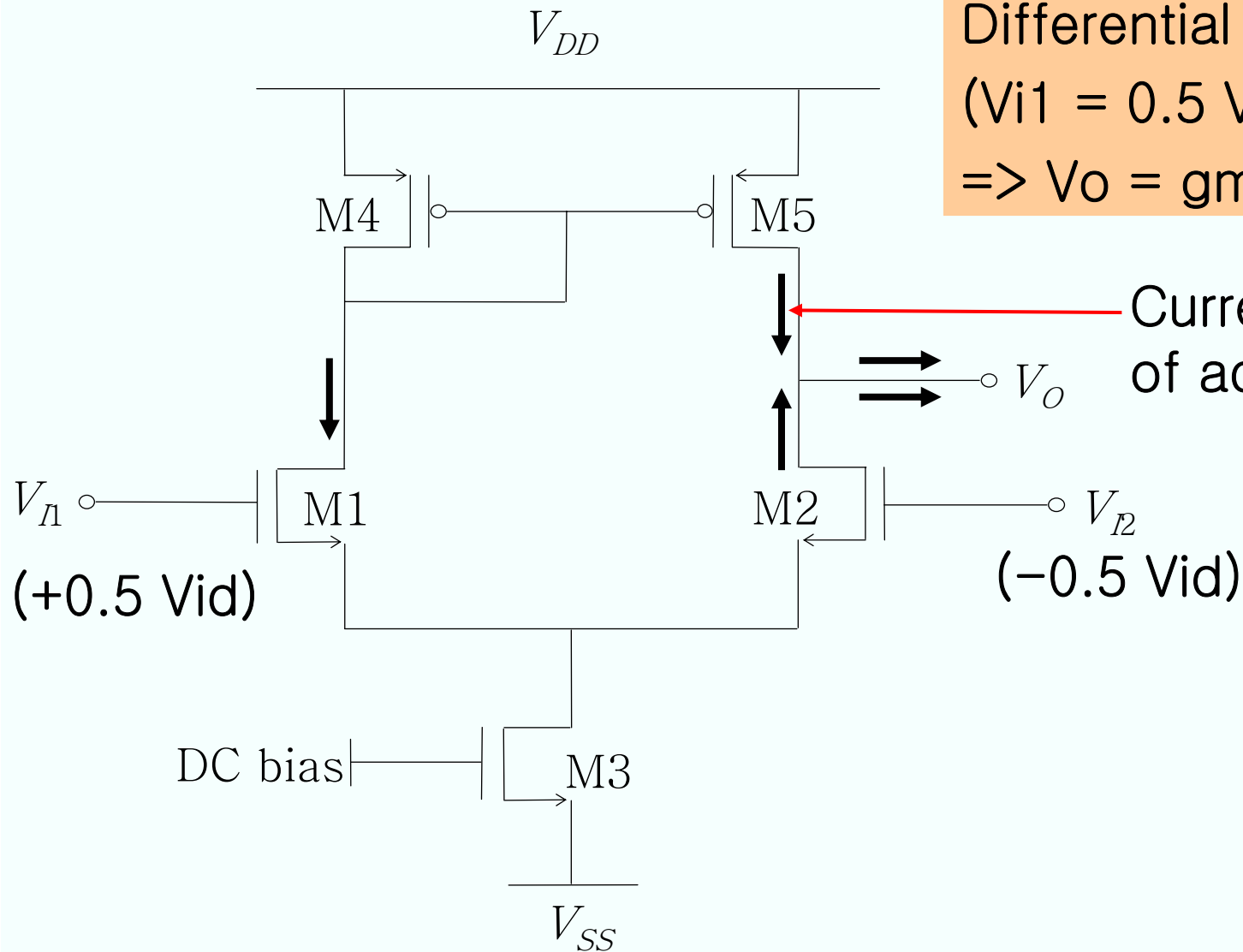
$$A_v \approx -\sqrt{\frac{\mu_N (W/L)_N}{\mu_P (W/L)_P}} = \frac{(V_{GS} - V_{TH})_P}{(V_{GS} - V_{TH})_N}$$

Trade-off between gain and Vout range



$$g_m = \sqrt{2 \mu C_{ox} \cdot \frac{W}{L} \cdot I_D} = \frac{I_D}{\frac{V_{GS} - V_{TH}}{2}}$$

5x reduction of I_D (M_3 , M_4) \Rightarrow g_{mP} reduced 5x \Rightarrow
 5x gain increased with the same V_{out} range ($V_{GS} - V_{TH}$ fixed)



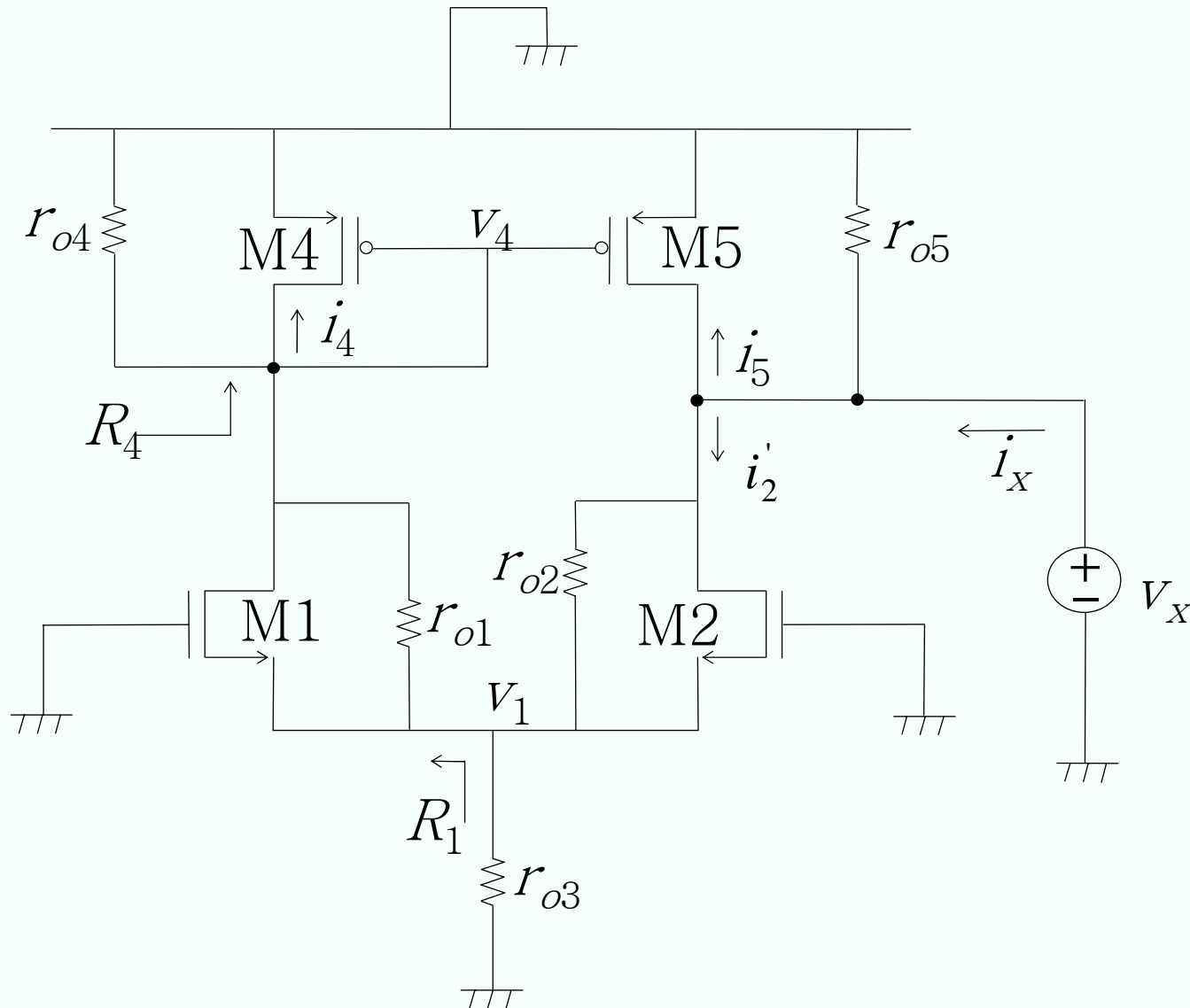
Differential mode input
 $(V_{i1} = 0.5 V_{id}, V_{i2} = -0.5 V_{id})$
 $\Rightarrow V_o = g_m \times R_o \times V_{id}$

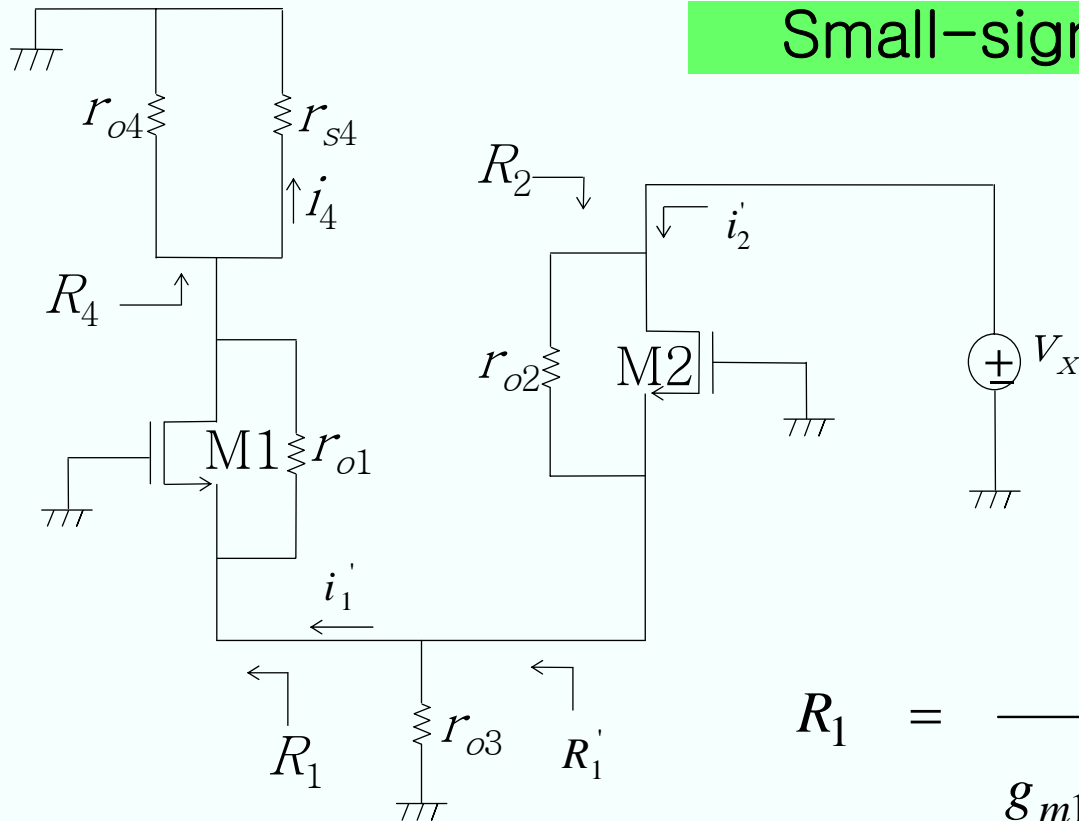
Current mirror operation
 of active load (M4, M5)

$$0.5 \times g_m \times V_{id}$$

Apply the Gm Ro method

Small-signal analysis (Ro)



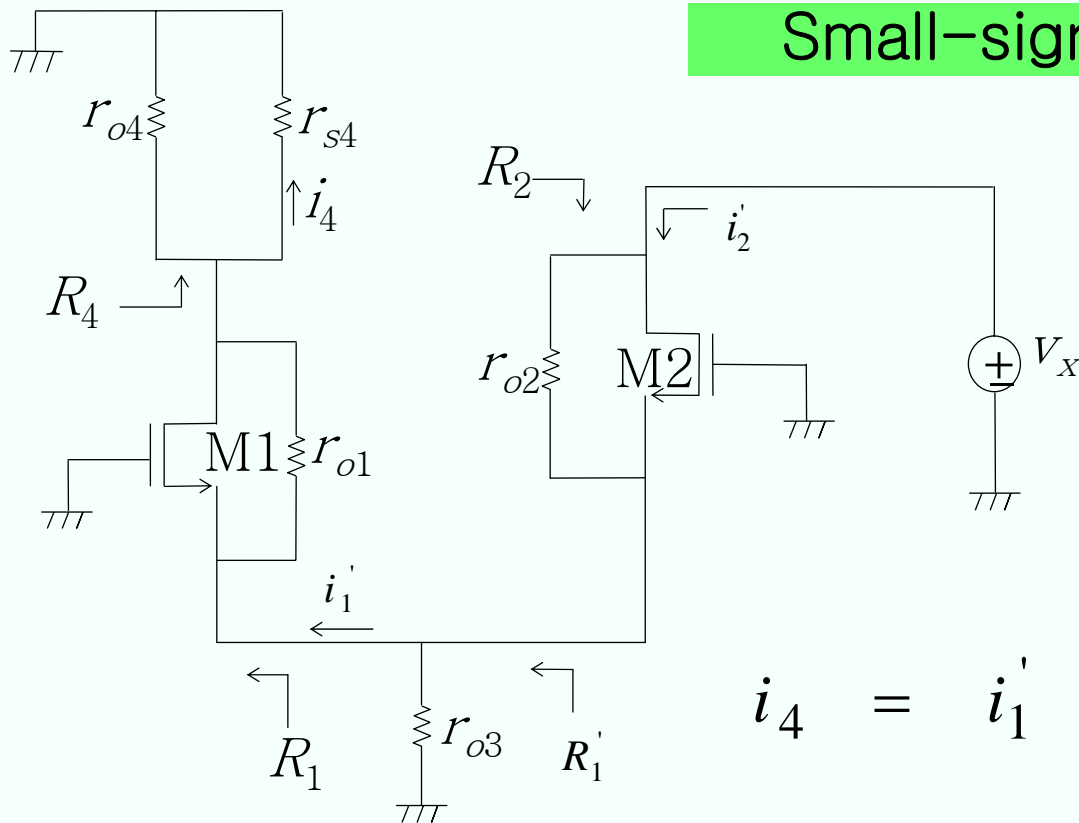
Small-signal analysis (R_o)

$$R_4 = r_{s4} \parallel r_{o4} \approx r_{s4} = \frac{1}{g_{m4}}$$

$$R_1 = \frac{1}{g_{m1} + \frac{1}{r_{o1}}} \cdot \left(1 + \frac{R_4}{r_{o1}} \right) \approx \frac{1}{g_{m1}} = r_{s1}$$

$$R_1' = r_{o3} \parallel R_1 = r_{o3} \parallel r_{s1} \approx r_{s1}$$

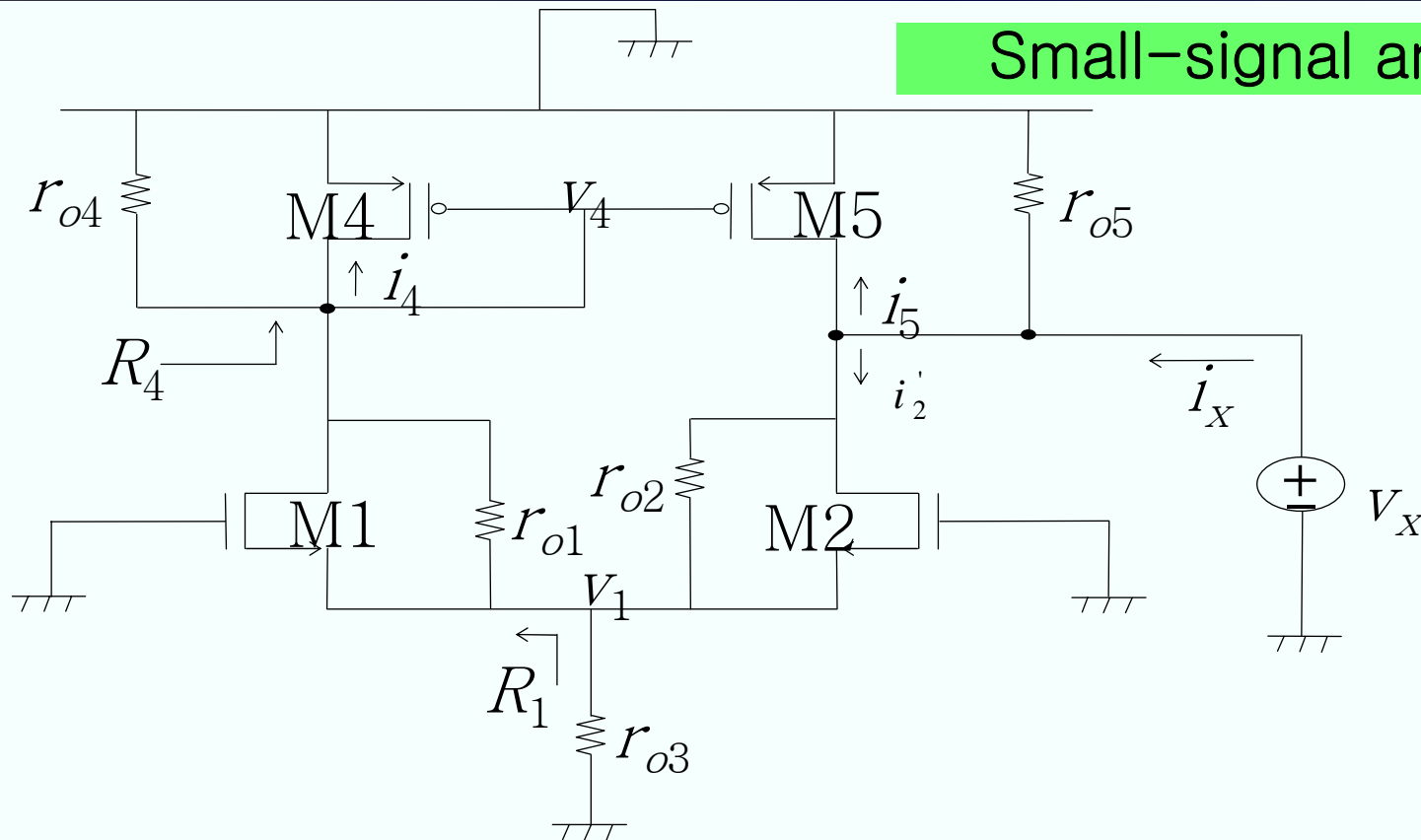
$$i_1' \approx i_2'$$

Small-signal analysis (R_o)

$$i_4 = i_1' \cdot \frac{r_{o4}}{r_{s4} + r_{o4}} \approx i_1' \approx i_2'$$

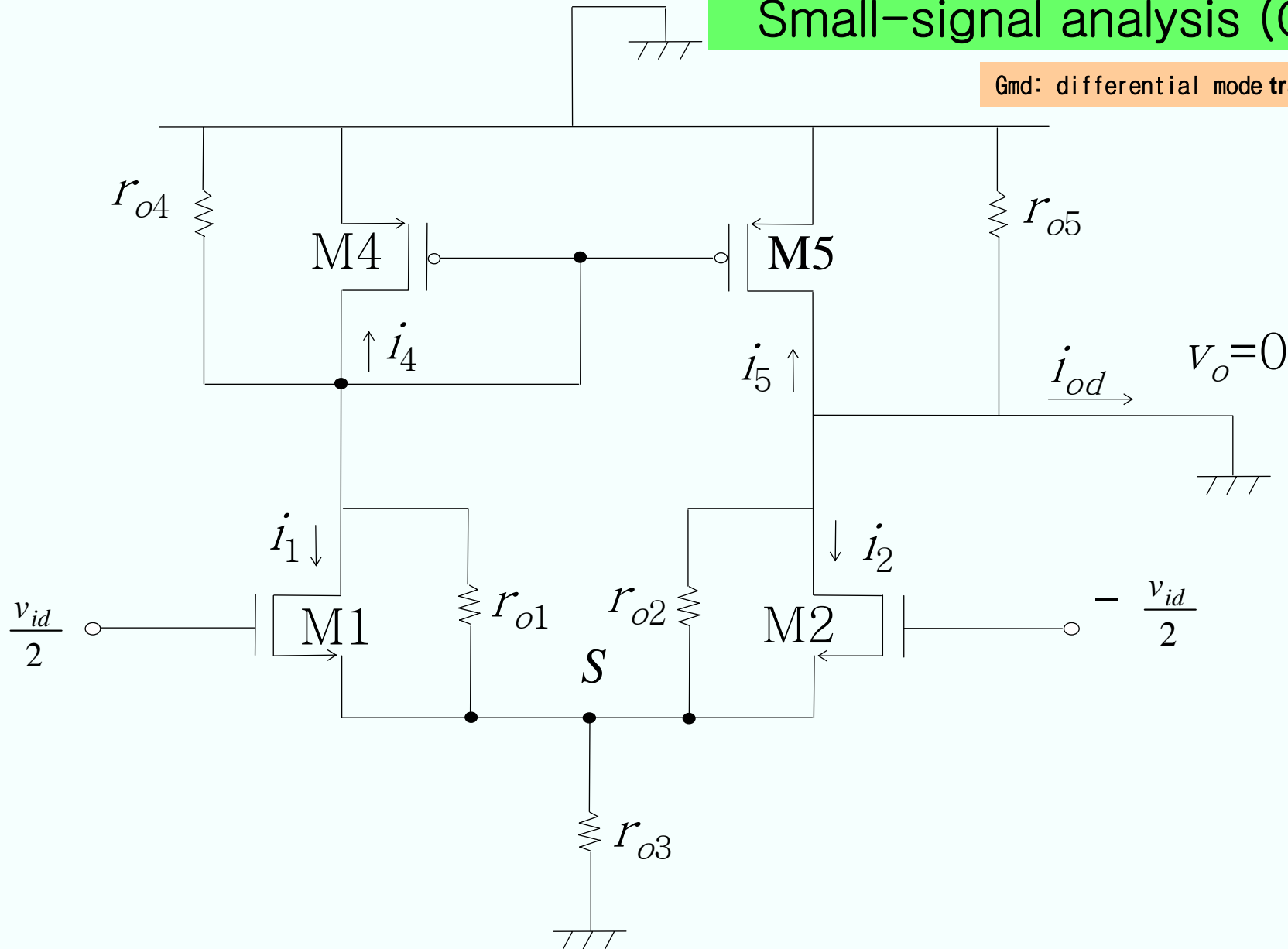
$$R_2 = g_{m2} r_{o2} \cdot R_1' + r_{o2} + R_1' = g_{m2} r_{o2} \cdot r_{s1} + r_{o2} + r_{s1} \approx 2r_{o2}$$

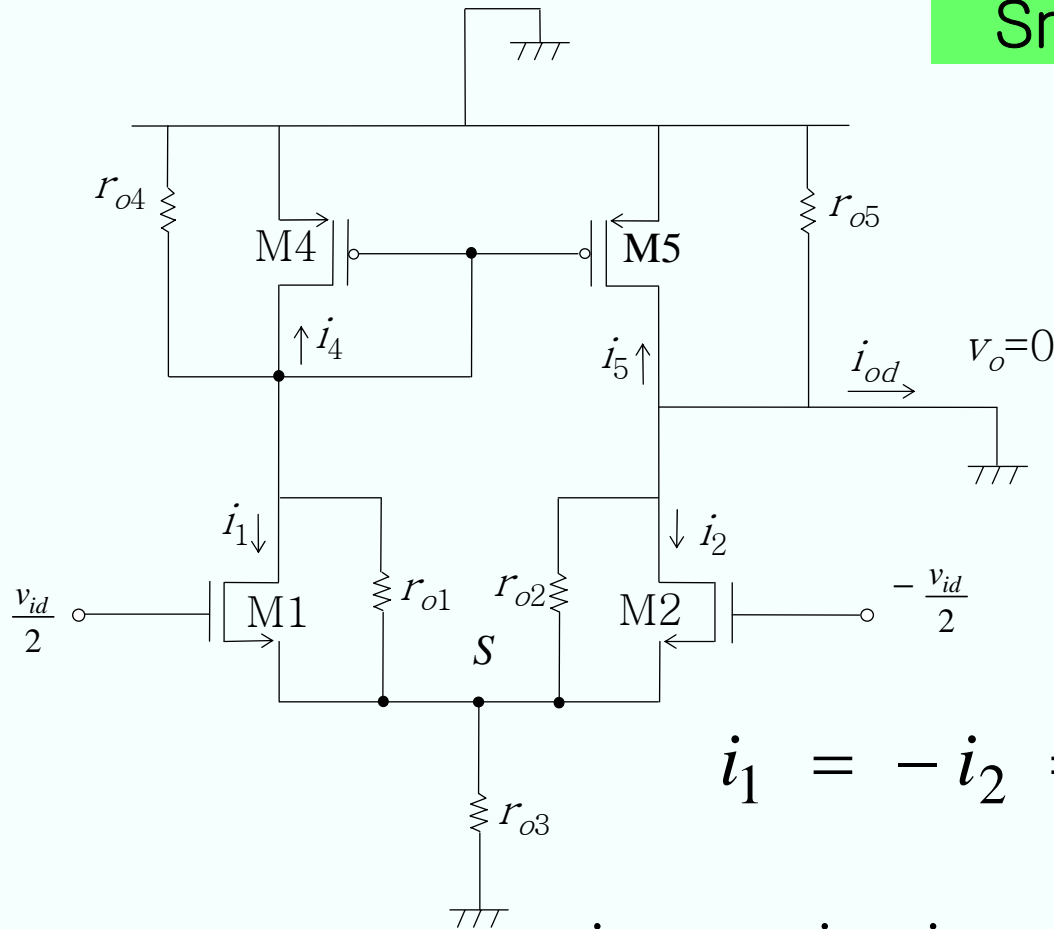
$$i_2' = \frac{v_x}{R_2} = \frac{v_x}{2r_{o2}}$$

Small-signal analysis (R_o)

$$i_x = i_5 + i_2' + \frac{v_x}{r_{o5}} = i_2' + i_2' + \frac{v_x}{r_{o5}} = v_x \cdot \left(\frac{1}{r_{o2}} + \frac{1}{r_{o5}} \right)$$

$$R_o = \frac{v_x}{i_x} = \left(\frac{1}{r_{o2}} + \frac{1}{r_{o5}} \right)^{-1} = r_{o2} \parallel r_{o5}$$

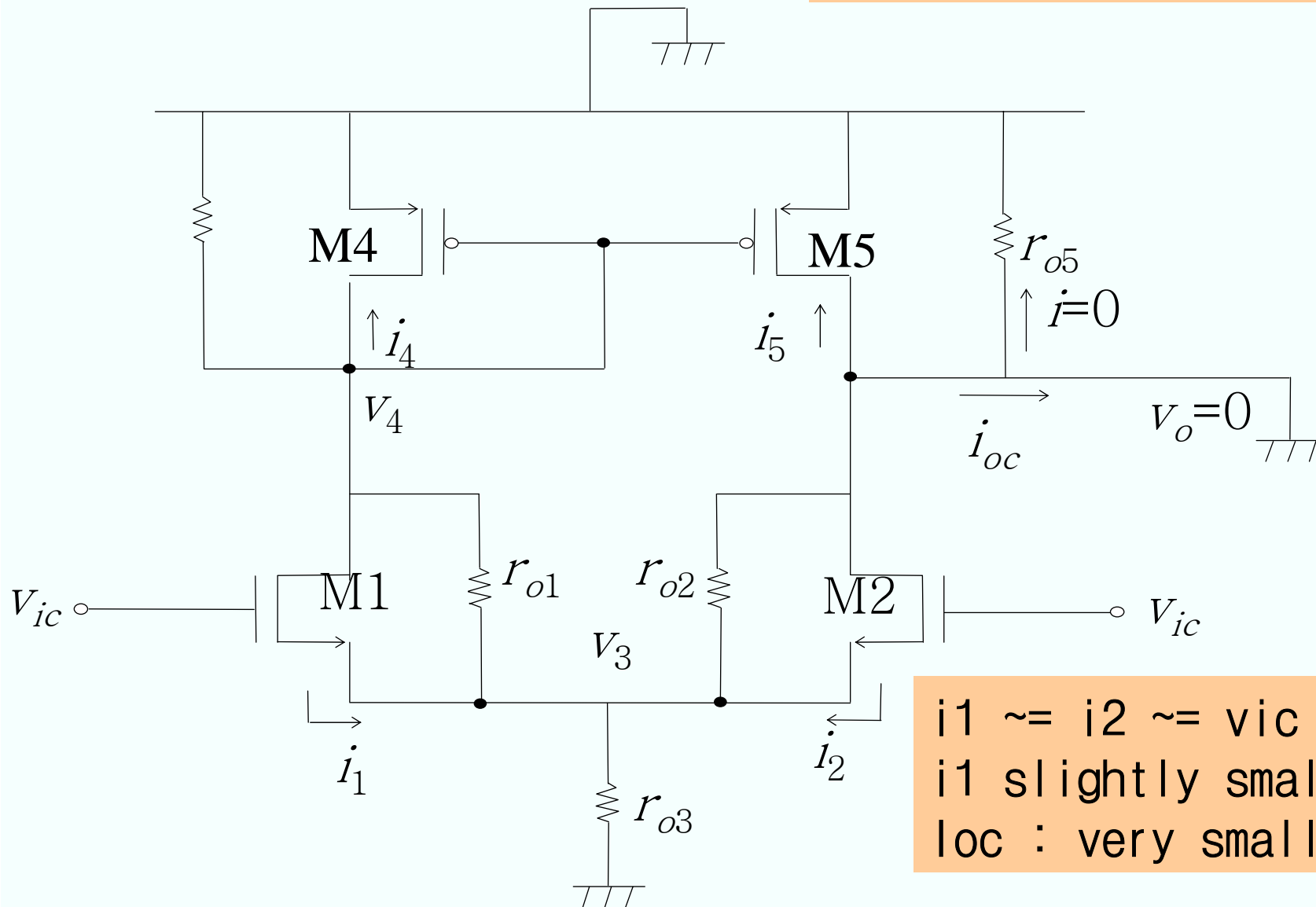
Small-signal analysis (G_{md})G_{md}: differential mode transconductance

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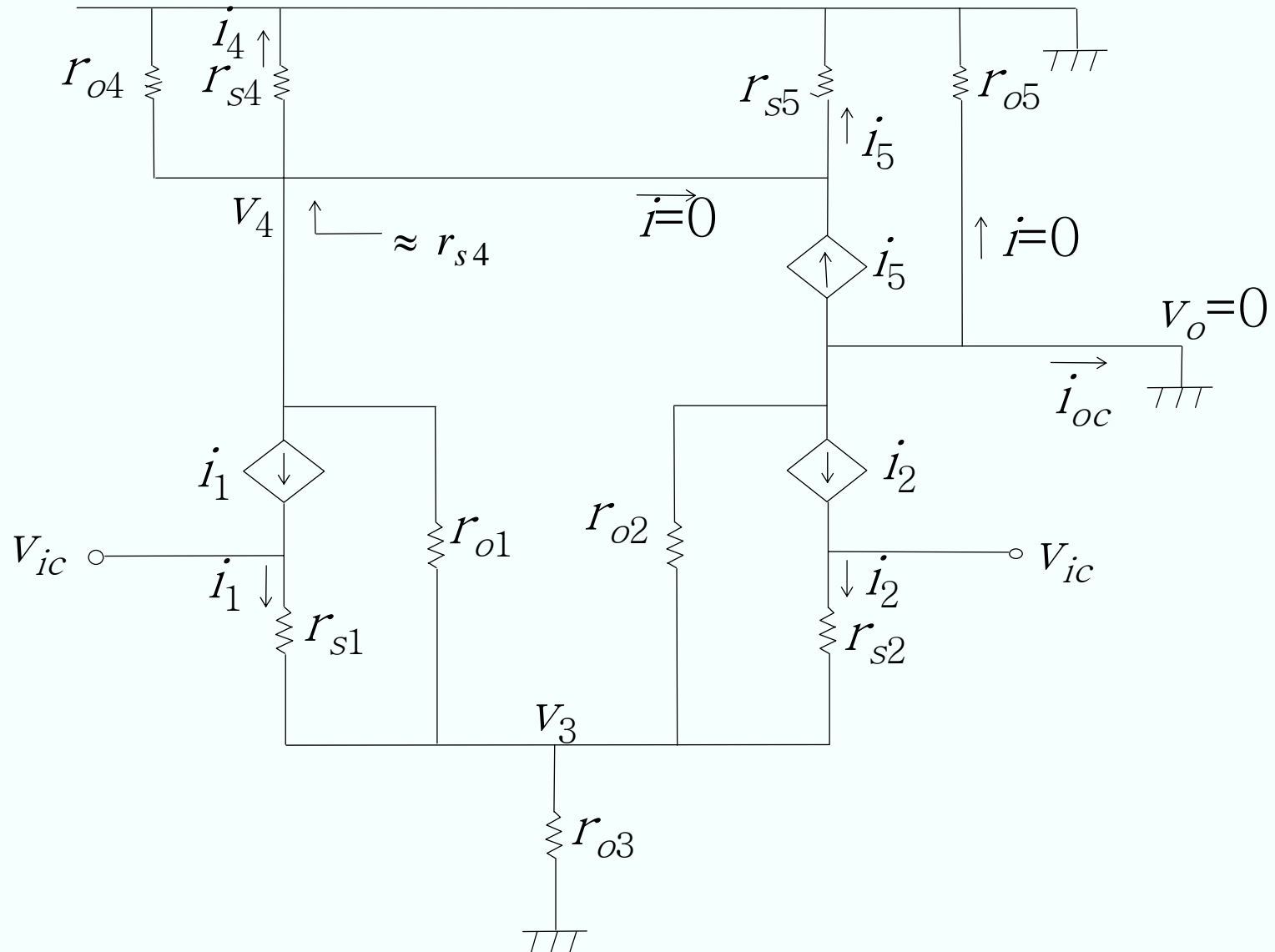
$$i_1 = -i_2 = v_{id} / (r_{s1} + r_{s2}) = v_{id} / (2r_{s1})$$

$$i_{od} = -i_5 - i_2 = -i_4 - i_2 = i_1 - i_2 = \frac{v_{id}}{r_{s1}} = g_{m1} v_{id}$$

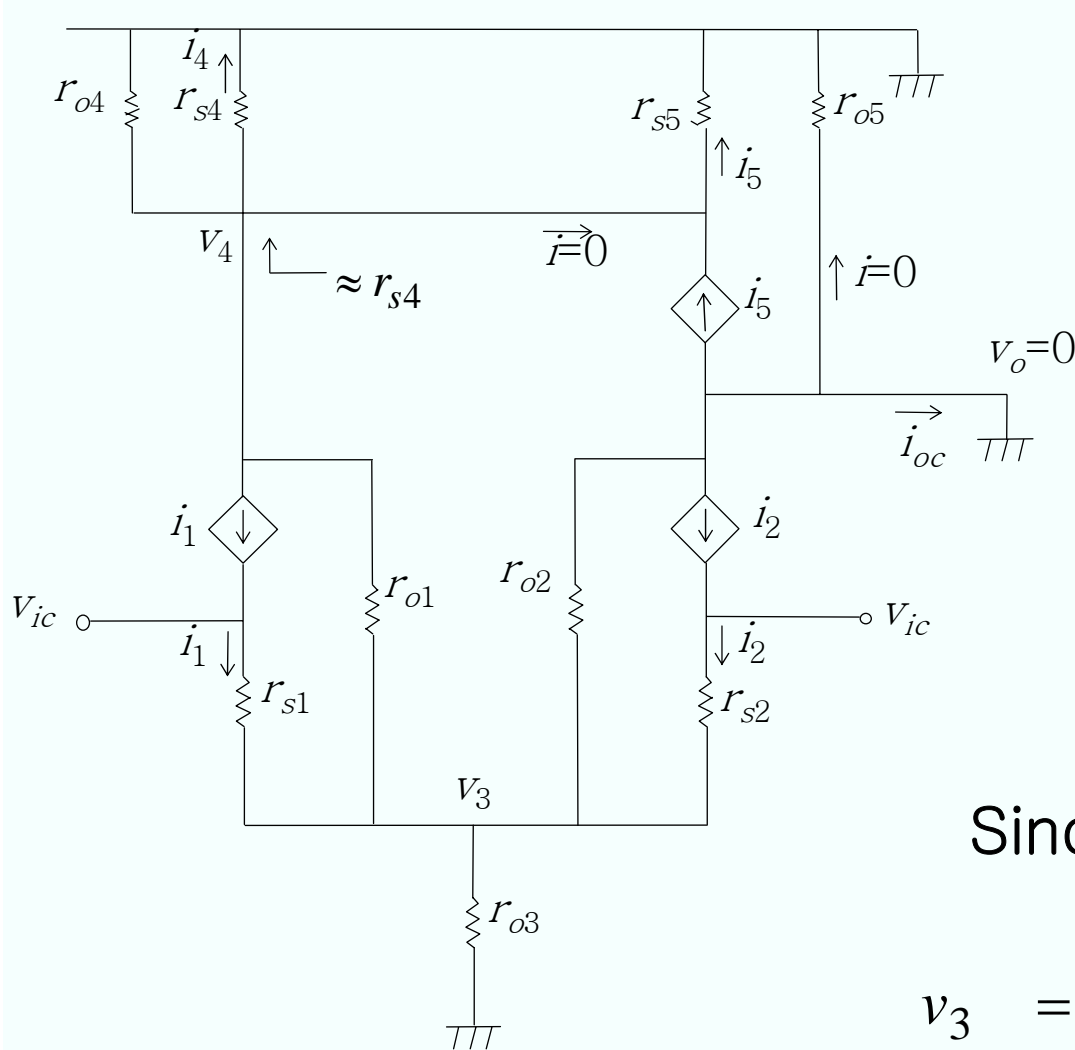
$$G_{md} = \frac{i_{od}}{v_{id}} = g_{m1}$$

Small-signal analysis (G_{mc})G_{mc} : common mode transconductance

$i_1 \approx i_2 \approx v_{ic} / (2 \times r_{o3})$
 i_1 slightly smaller than i_2
 i_{oc} : very small negative

Small-signal analysis (G_{mc})G_{mc}: common mode transconductance

Small-signal analysis (Gmc)

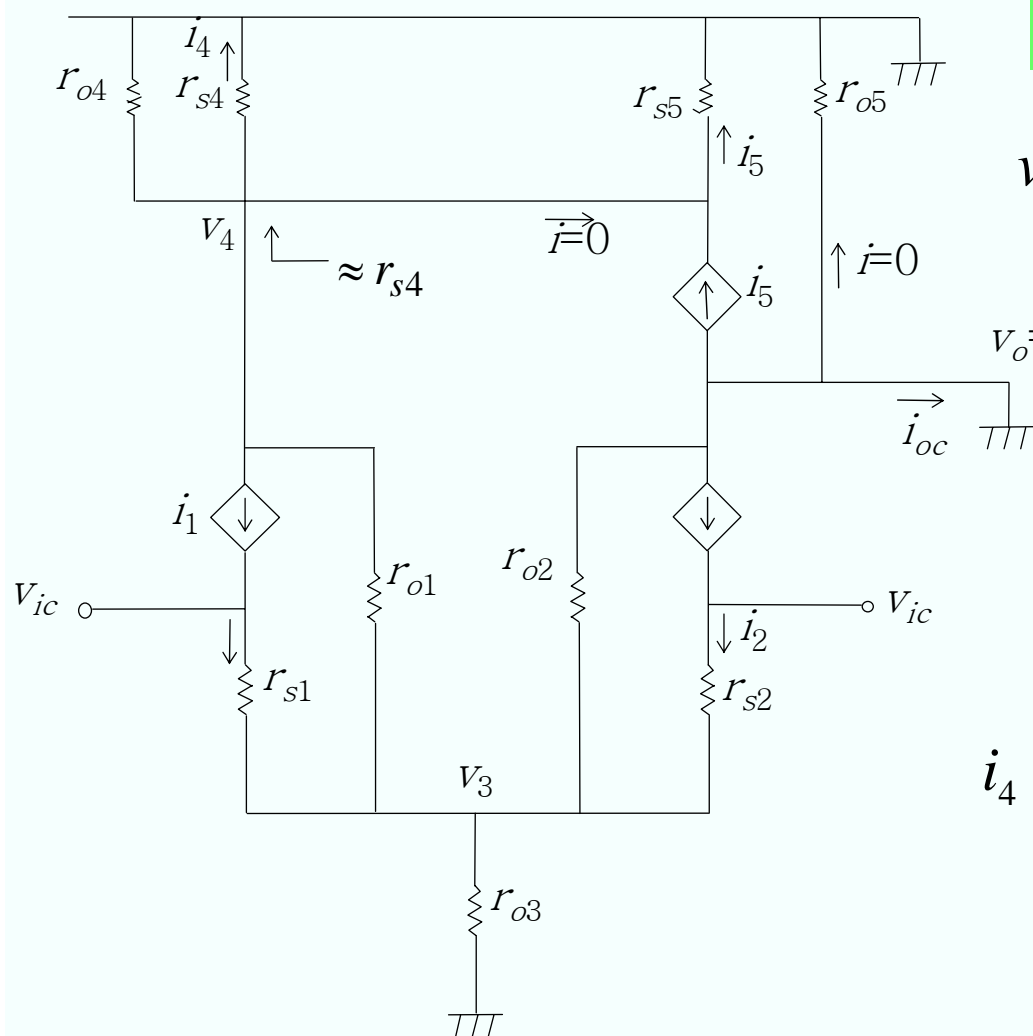


$$v_4 = \left(-i_1 + \frac{v_3 - v_4}{r_{o1}} \right) \cdot (r_{s4} \parallel r_{o4})$$

$$v_4 = \frac{-i_1 \cdot (r_{s4} \parallel r_{o4}) + \frac{r_{s4} \parallel r_{o4}}{r_{o1}} \cdot v_3}{1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}}}$$

Since $v_4 \ll v_3$, assume $v_4 \approx 0$

$$v_3 = \frac{2(r_{o1} \parallel r_{o2} \parallel r_{o3})}{r_{s1} + 2(r_{o1} \parallel r_{o2} \parallel r_{o3})} \cdot v_{ic} \approx v_{ic}$$



Intuitive way of deriving Gmc

$$v_3 = \frac{2(r_{o1} \parallel r_{o2} \parallel r_{o3})}{r_{s1} + 2(r_{o1} \parallel r_{o2} \parallel r_{o3})} \cdot v_{ic} \approx v_{ic}$$

$$i_3 = \frac{v_{ic}}{r_{o3}}$$

$$i_{M1} \approx 0.5 \cdot i_3 \cdot \left(1 - 0.5 \cdot \frac{r_{s4}}{r_{o1}}\right)$$

$$i_{M2} \approx 0.5 \cdot i_3 \cdot \left(1 + 0.5 \cdot \frac{r_{s4}}{r_{o1}}\right)$$

$$i_4 = i_5 = -0.5 \cdot i_3 \cdot \left(1 - 0.5 \cdot \frac{r_{s4}}{r_{o1}}\right) \cdot \left(1 - \frac{r_{s4}}{r_{o4}}\right)$$

$$\approx -0.5 \cdot i_3 \cdot \left(1 - 0.5 \cdot \frac{r_{s4}}{r_{o1}} - \frac{r_{s4}}{r_{o4}}\right)$$

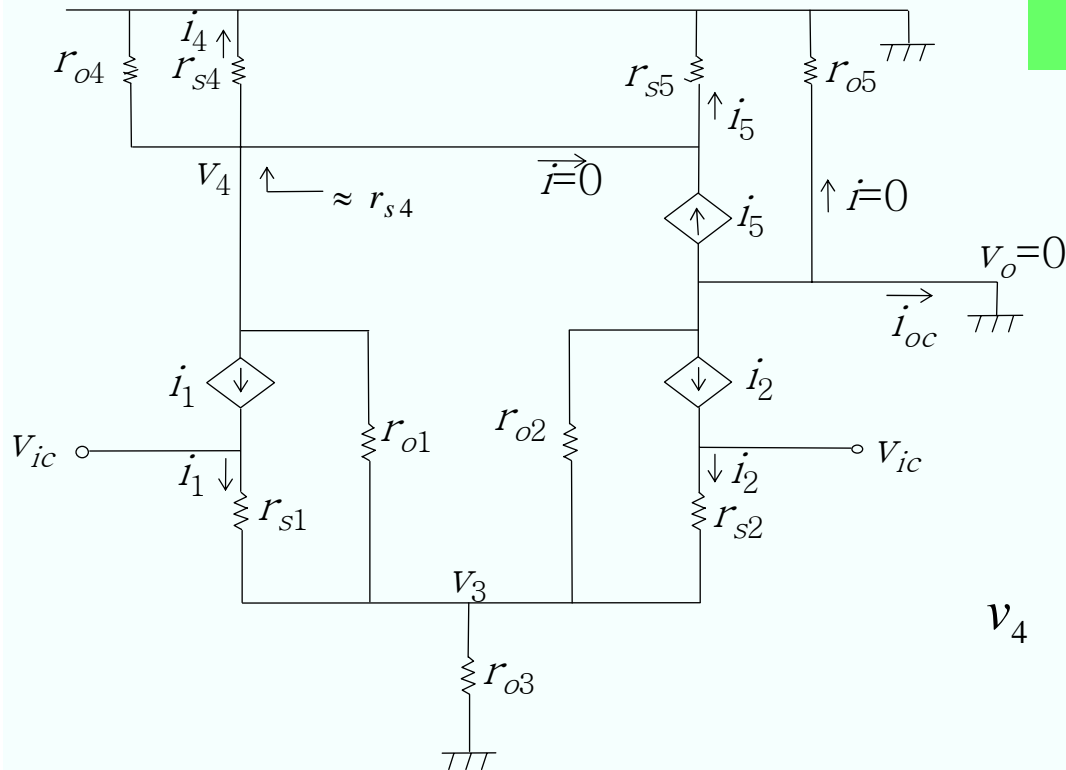
$$i_{oc} = -i_5 - i_{M2} = -0.5 i_3 \cdot r_{s4} \cdot \left(\frac{1}{r_{o1}} + \frac{1}{r_{o4}}\right) = \frac{-v_{ic}}{2 r_{o3} \cdot g_{m4} \cdot (r_{o1} \parallel r_{o4})}$$

Details of deriving Gmc

$$v_4 = \frac{r_{s4} \parallel r_{o4}}{1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}}} \cdot v_{ic} \cdot \left\{ \frac{1}{r_{o1}} - \frac{1}{2(r_{o1} \parallel r_{o2} \parallel r_{o3})} \right\}$$

$$v_4 = \frac{r_{s4} \parallel r_{o4}}{1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}}} \cdot v_{ic} \cdot \left\{ -\frac{1}{2r_{o3}} \right\} \quad \text{if } r_{o1} = r_{o2}$$

Details of deriving Gmc



$$v_4 = \frac{r_{s4} \parallel r_{o4}}{1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}}} \cdot v_{ic} \cdot \left\{ -\frac{1}{2r_{o3}} \right\} \quad \text{if } r_{o1} = r_{o2}$$

$$i_4 = \frac{v_4}{r_{s4}} = \frac{v_{ic}}{1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}}} \cdot \left(-\frac{1}{2r_{o3}} \right) \cdot \frac{r_{s4} \parallel r_{o4}}{r_{s4}} = \frac{v_{ic}}{1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}}} \cdot \left(-\frac{1}{2r_{o3}} \right) \cdot \frac{1}{1 + \frac{r_{s4}}{r_{o4}}}$$

$$\approx \left\{ -\frac{v_{ic}}{2r_{o3}} \right\} \cdot \left(1 - \frac{r_{s4} \parallel r_{o4}}{r_{o1}} - \frac{r_{s4}}{r_{o4}} \right) \approx \left\{ -\frac{v_{ic}}{2r_{o3}} \right\} \cdot \left(1 - \frac{r_{s4}}{r_{o1}} - \frac{r_{s4}}{r_{o4}} \right) = \left\{ -\frac{v_{ic}}{2r_{o3}} \right\} \cdot \left\{ 1 - \frac{1}{g_{m4}(r_{o1} \parallel r_{o4})} \right\}$$

Details of deriving Gmc

$$\begin{aligned}
 i_{oc} &= -i_5 - i_2 + \frac{v_3}{r_{o2}} = -i_4 - i_1 + \frac{v_{ic}}{r_{o2}} \\
 &= v_{ic} \cdot \left(\frac{1}{2r_{o3}} \cdot \left\{ 1 - \frac{1}{g_{m4}(r_{o1} \parallel r_{o4})} \right\} - \frac{1}{2(r_{o1} \parallel r_{o2} \parallel r_{o3})} + \frac{1}{r_{o2}} \right) \\
 &\approx v_{ic} \cdot \left\{ \frac{1}{2r_{o3}} \cdot \left(1 - \frac{1}{g_{m4} \cdot (r_{o1} \parallel r_{o4})} \right) - \frac{1}{2} \left(\frac{1}{r_{o1}} + \frac{1}{r_{o2}} + \frac{1}{r_{o3}} \right) + \frac{1}{r_{o2}} \right\} \\
 &\approx v_{ic} \cdot \left\{ \frac{1}{2r_{o3}} \cdot \left(1 - \frac{1}{g_{m4} \cdot (r_{o1} \parallel r_{o4})} \right) - \frac{1}{2r_{o3}} \right\} \quad \text{if } r_{o1} = r_{o2}
 \end{aligned}$$

$$= - \frac{v_{ic}}{2r_{o3} \cdot \{g_{m4}(r_{o1} \parallel r_{o4})\}}$$

Small-signal analysis (G_{mc})

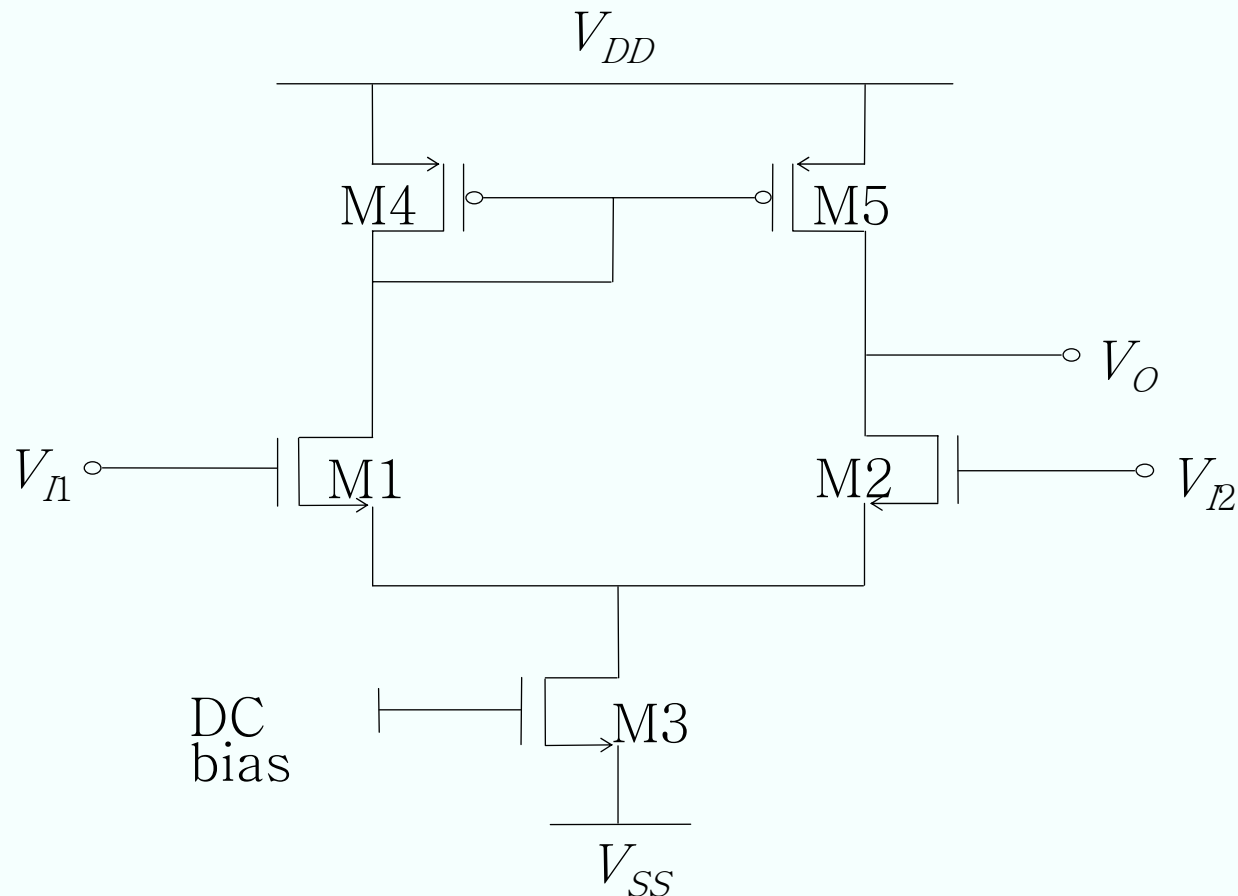
$$i_{oc} = - \frac{v_{ic}}{2r_{o3} \cdot \{g_{m4}(r_{o1} \parallel r_{o4})\}}$$

$$G_{mc} = \frac{i_{oc}}{v_{ic}} = - \frac{1}{2r_{o3} \cdot g_{m4}(r_{o1} \parallel r_{o4})}$$

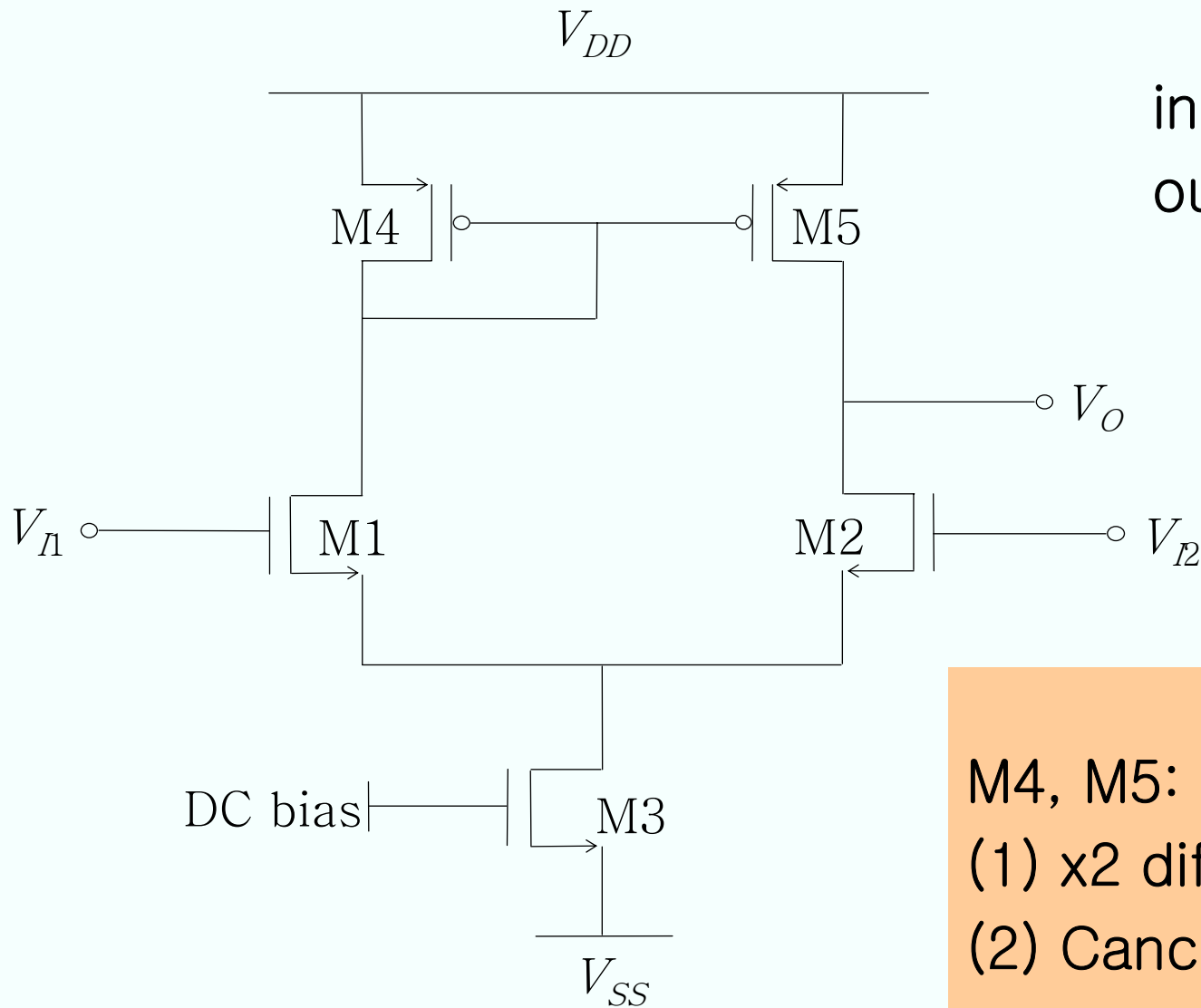
$$CMRR = \left| \frac{A_{vd}}{A_{vc}} \right| = \left| \frac{G_{md}R_o}{G_{mc}R_o} \right| = \left| \frac{G_{md}}{G_{mc}} \right| = (2g_{m1}r_{o3}) \cdot \{g_{m4}(r_{o1} \parallel r_{o4})\}$$

$$CMRR(\text{active load}) = CMRR(\text{R load}) \times g_{m4}(r_{o1} \parallel r_{o4})$$

Active input common mode voltage range



$$V_{SS} + V_{DSAT3} + V_{DSAT1} + V_{TH.1} \leq V_{IC} \leq V_{DD} - |V_{DSAT4}| - |V_{TH.4}| + V_{TH.1}$$



input: differential signal
output: single-ended

M4, M5: active load
(1) x2 differential mode input
(2) Cancel common mode input

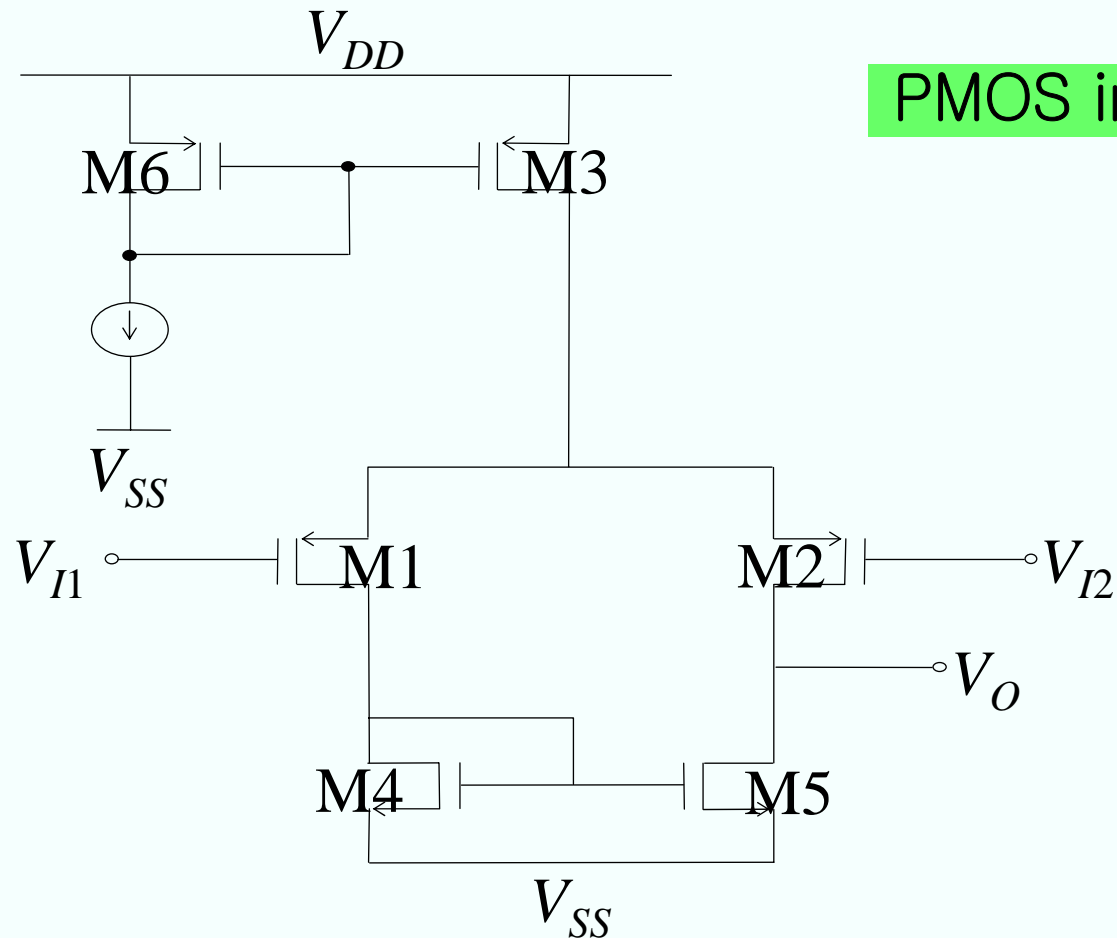
$$G_{md} = \frac{i_{od}}{v_{id}} = g_{m1}$$

$$G_{mc} = \frac{i_{oc}}{v_{ic}} = - \frac{1}{2r_{o3} \cdot g_{m4} (r_{o1} \parallel r_{o4})}$$

$$CMRR = \left| \frac{A_{vd}}{A_{vc}} \right| = \left| \frac{G_{md}R_o}{G_{mc}R_o} \right| = \left| \frac{G_{md}}{G_{mc}} \right| = (2g_{m1}r_{o3}) \cdot \{g_{m4}(r_{o1} \parallel r_{o4})\}$$

$$R_o = \frac{v_x}{i_x} = \left(\frac{1}{r_{o2}} + \frac{1}{r_{o5}} \right)^{-1} = r_{o2} \parallel r_{o5}$$

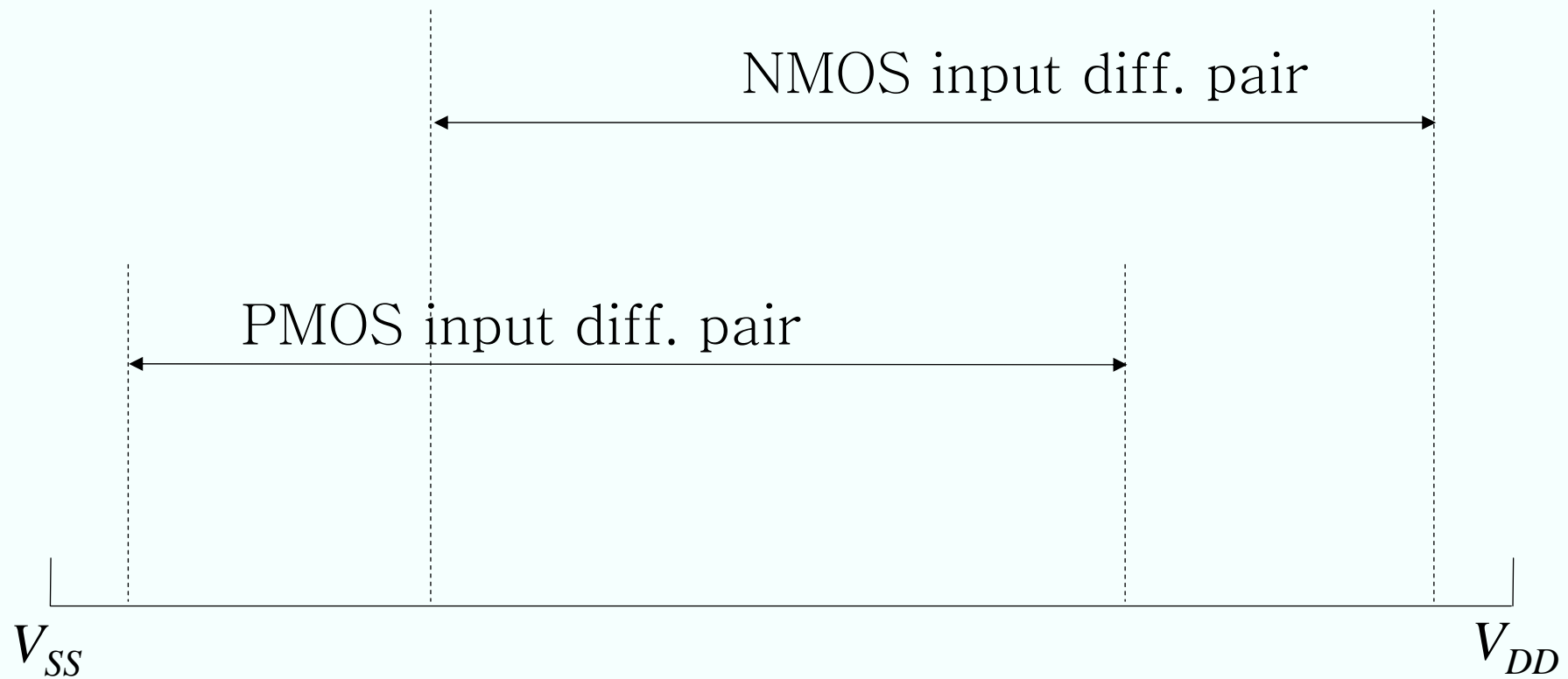
Active input common mode voltage range



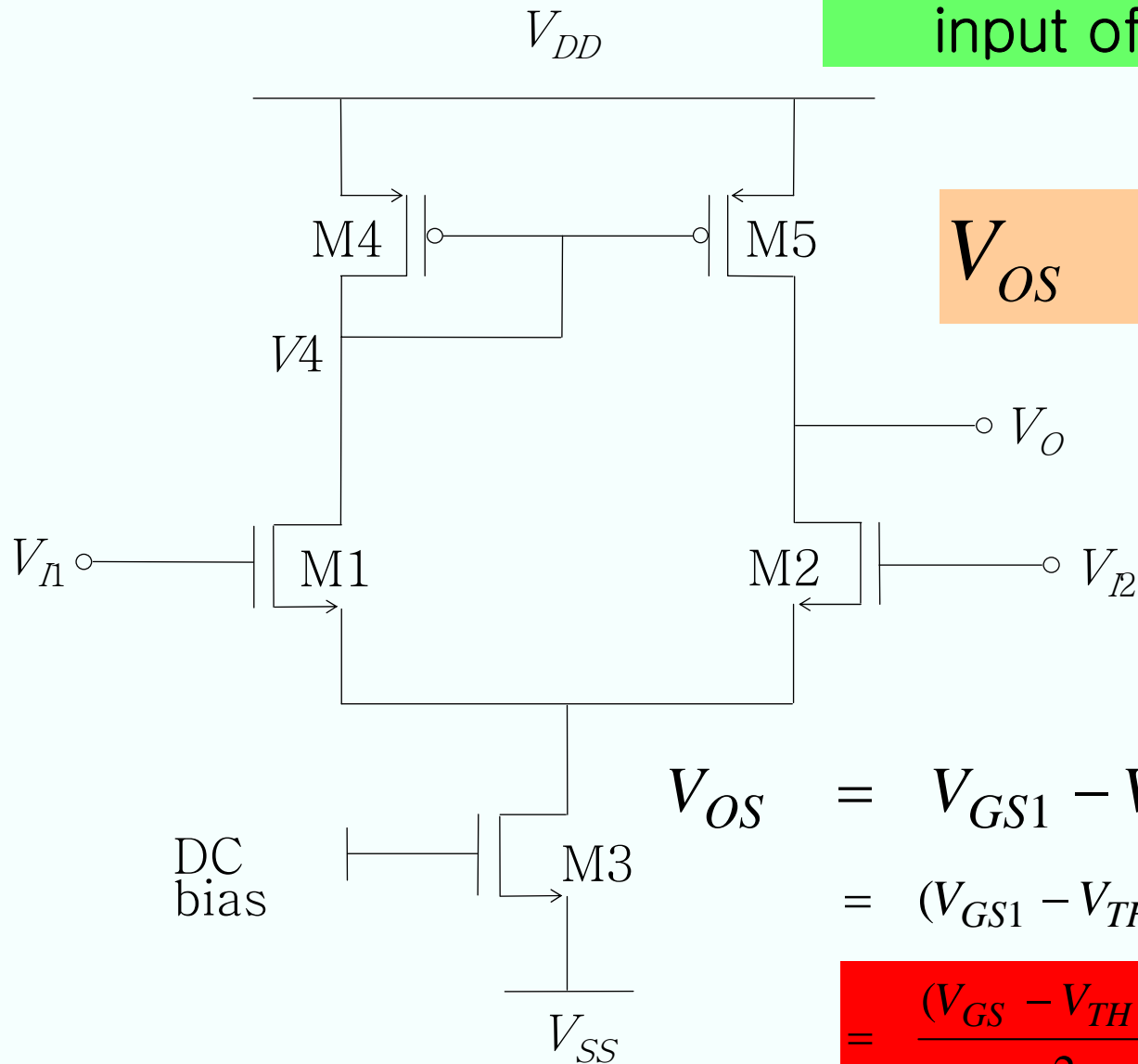
PMOS input diff pair

$$V_{SS} + V_{DSAT.4} + V_{TH.4} - |V_{TH.1}| \leq V_{IC} \leq V_{DD} - |V_{DSAT.3}| - |V_{DSAT.1}| - |V_{TH.1}|$$

Active input common mode voltage range



input offset voltage



$$V_{OS} = V_{I1} - V_{I2} \quad (V_O = V_4)$$

$$\begin{aligned} V_{OS} &= V_{GS1} - V_{GS2} \\ &= (V_{GS1} - V_{TH1}) - (V_{GS2} - V_{TH2}) + (V_{TH1} - V_{TH2}) \end{aligned}$$

$$= \frac{(V_{GS} - V_{TH})_{1,2}}{2} \cdot \left\{ \frac{\Delta I_{D1,2}}{I_{D1,2}} - \frac{\Delta(W/L)_{1,2}}{(W/L)_{1,2}} \right\} + \Delta V_{TH1,2}$$

input offset voltage

$$V_{OS} = \frac{(V_{GS} - V_{TH})_{1,2}}{2} \cdot \left\{ \frac{\Delta I_{D1,2}}{I_{D1,2}} - \frac{\Delta(W/L)_{1,2}}{(W/L)_{1,2}} \right\} + \Delta V_{TH1,2}$$

$$I_D = \mu_p C_{ox} \cdot (W/L) \cdot (V_{GS} - V_{TH})^2 / 2$$

$$\ln I_D = \ln \left(\frac{1}{2} \mu_p C_{ox} \right) + \ln(W/L) + 2 \ln(V_{GS} - V_{TH})$$

$$\frac{\Delta I_{D4,5}}{I_{D4,5}} = \frac{\Delta(W/L)_{4,5}}{(W/L)_{4,5}} + \left\{ \frac{\Delta(V_{GS} - V_{TH})_{4,5}}{(V_{GS} - V_{TH})_{4,5}} \right\} / 2$$

input offset voltage

$$V_{OS} = \Delta V_{TH1,2} - \frac{(V_{GS} - V_{TH})_{1,2}}{(V_{GS} - V_{TH})_{4,5}} \cdot \Delta V_{TH4,5} + \frac{(V_{GS} - V_{TH})_{1,2}}{2} \cdot \left\{ -\frac{\Delta(W/L)_{1,2}}{(W/L)_{1,2}} + \frac{\Delta(W/L)_{4,5}}{(W/L)_{4,5}} \right\}$$

$$(V_{GS} - V_{TH})_{1,2} / (V_{GS} - V_{TH})_{4,5} = g_{m4,5} / g_{m1,2} = \sqrt{\mu_p (W/L)_{4,5} / \{\mu_n (W/L)_{1,2}\}}$$

$$= \Delta V_{TH1,2} + \Delta V_{TH4,5} \cdot \sqrt{\frac{\mu_p \cdot (W/L)_{4,5}}{\mu_n \cdot (W/L)_{1,2}}} + \sqrt{\frac{I_{D1,2}}{2\mu_n C_{ox} (W/L)_{1,2}}} \cdot \left\{ -\frac{\Delta(W/L)_{1,2}}{(W/L)_{1,2}} + \frac{\Delta(W/L)_{4,5}}{(W/L)_{4,5}} \right\}$$

For small offset

$$(W/L)_{4,5} \ll (W/L)_{1,2}$$

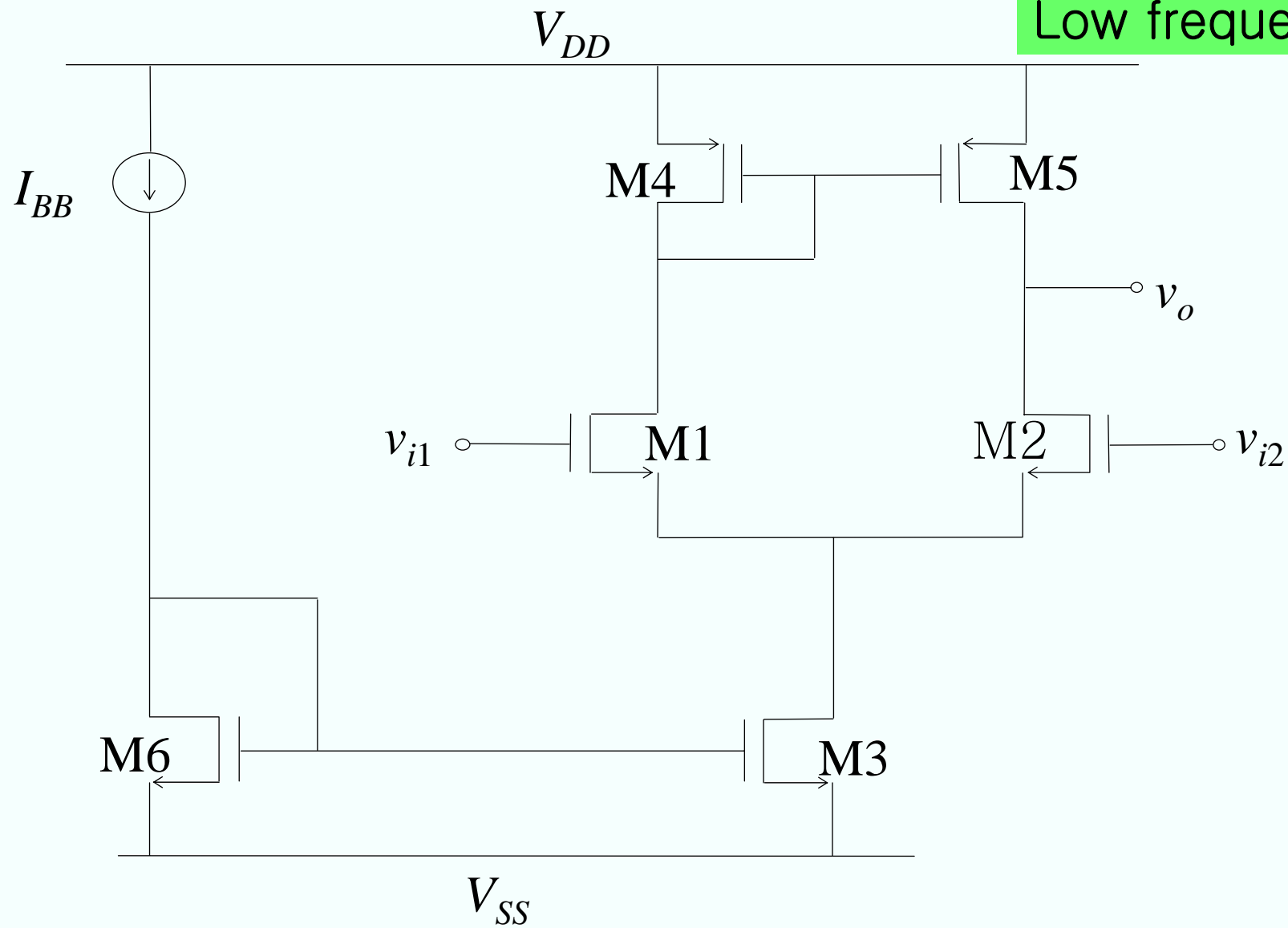
5.2.6 Low frequency PSRR of CMOS active load diff. pair (*)

PSRR : power supply rejection ratio

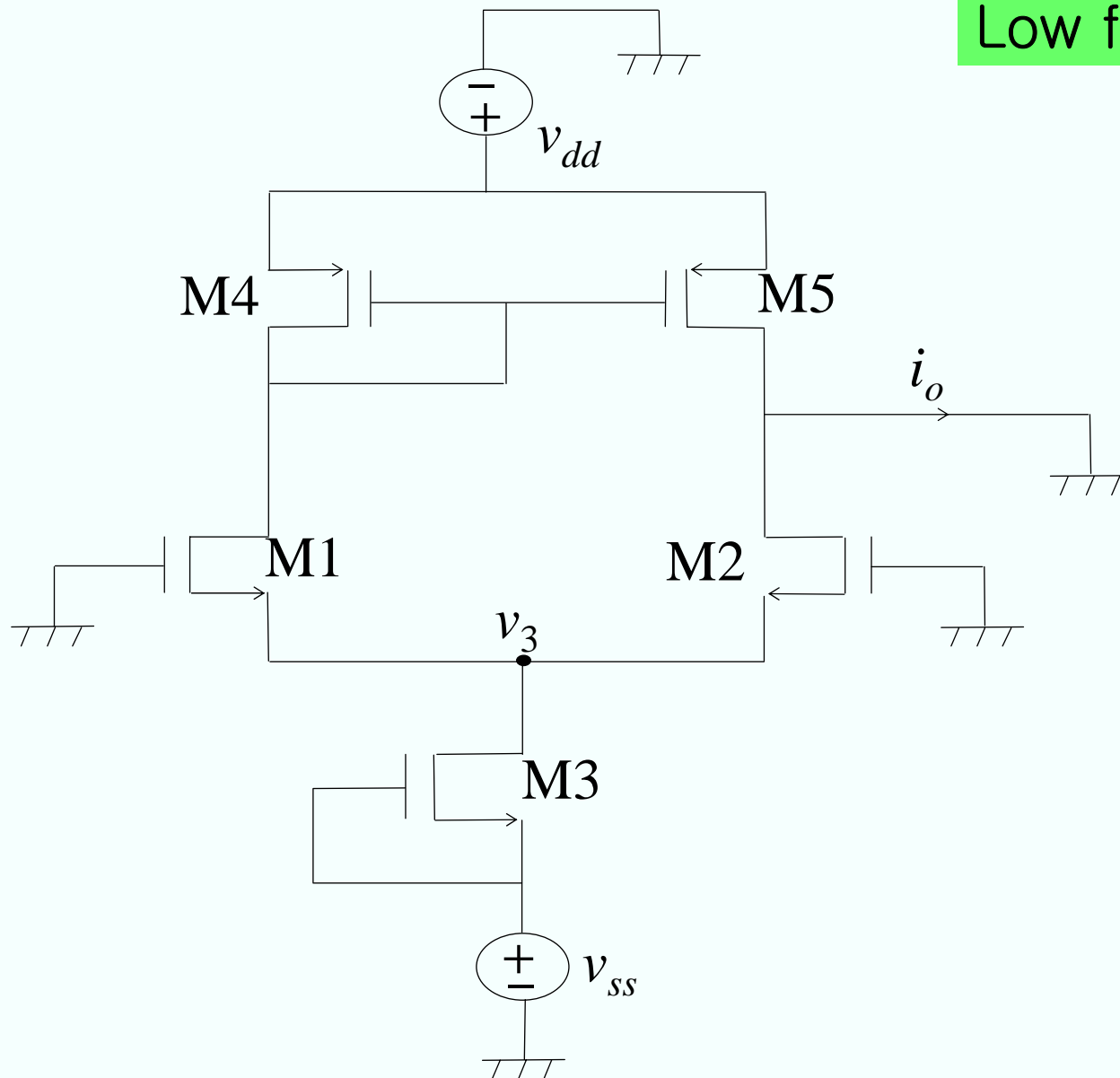
$$PSRR^+ \triangleq \left| \frac{A_{vd}}{v_o / v_{dd}} \right|$$

$$PSRR^- \triangleq \left| \frac{A_{vd}}{v_o / v_{ss}} \right|$$

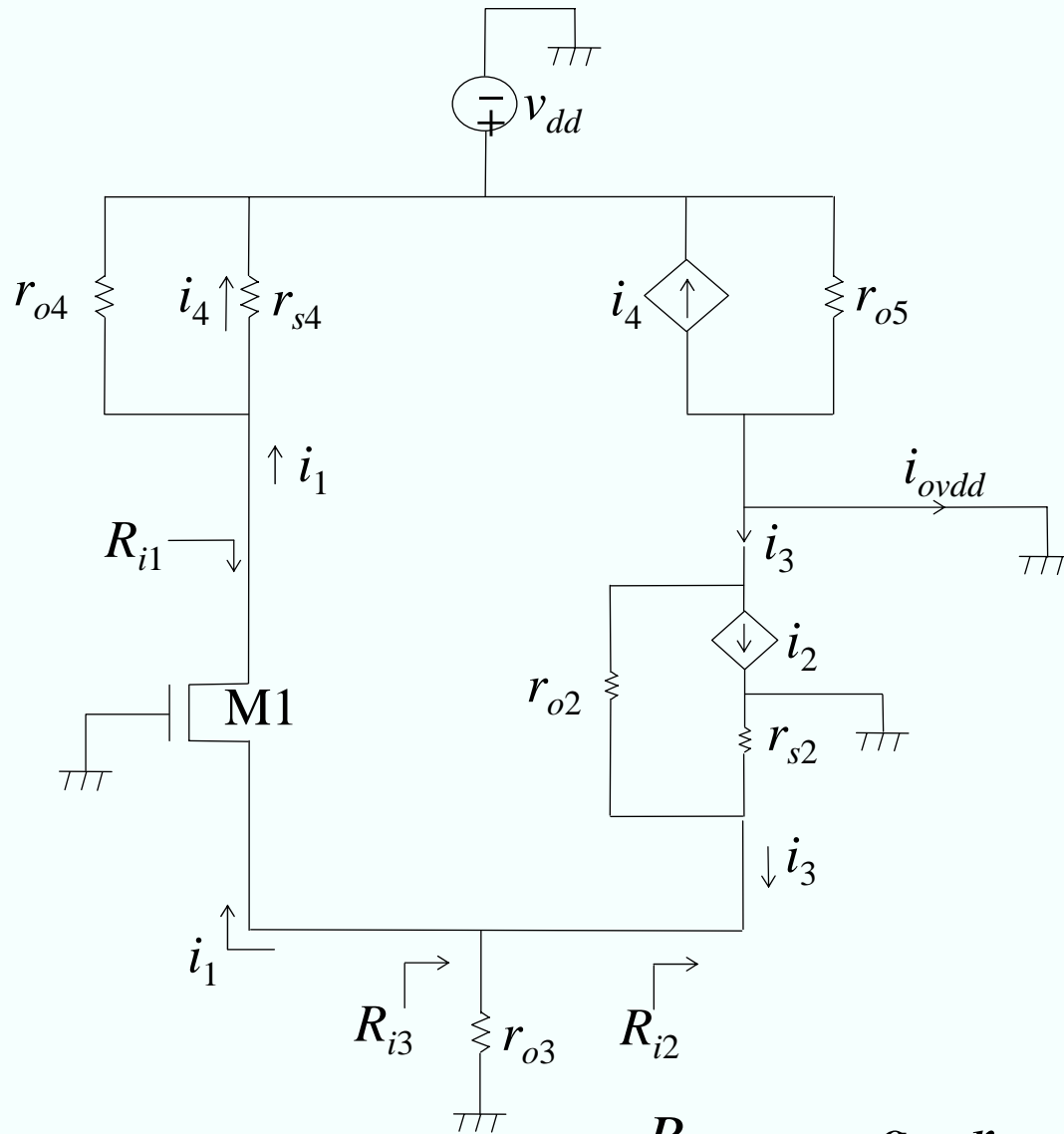
Low frequency PSRR



Low frequency PSRR +



Low frequency PSRR +

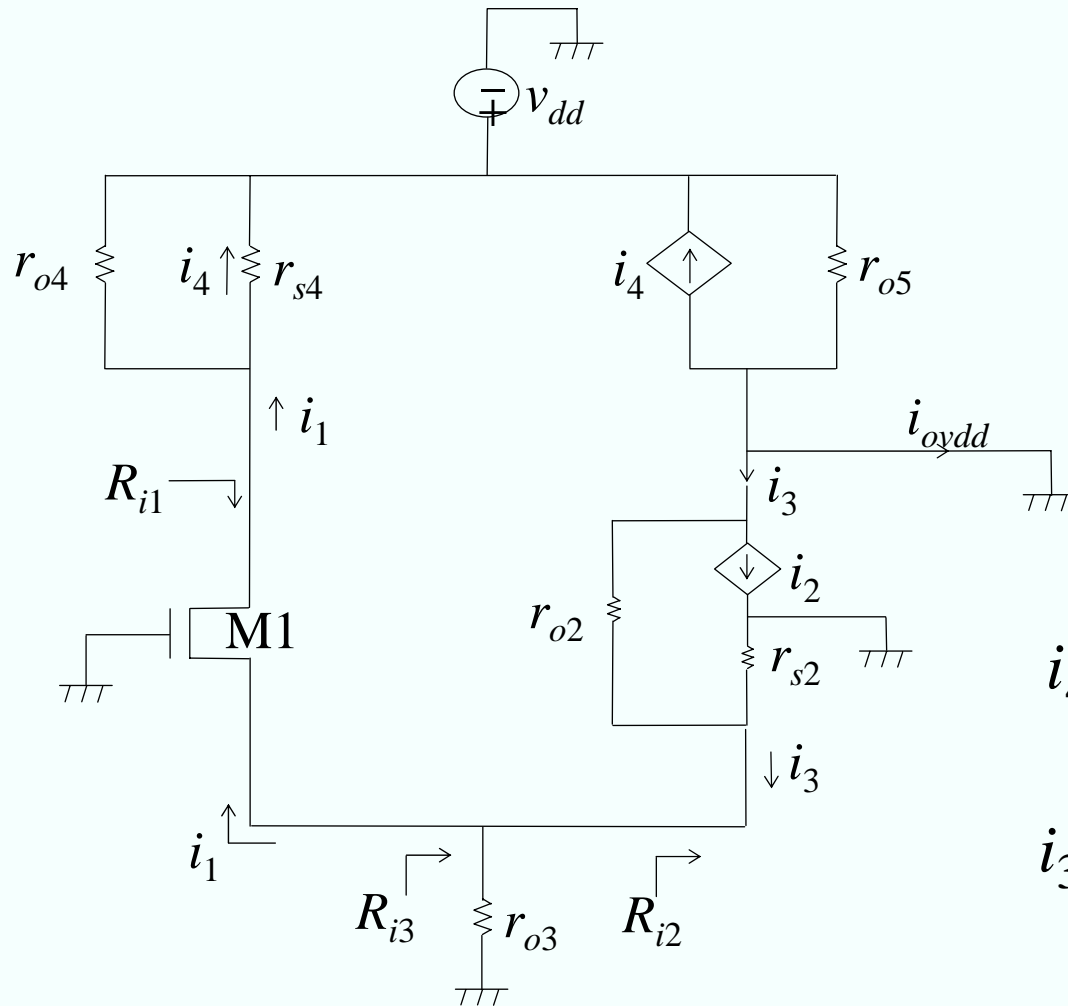


$$R_{i2} = r_{s2} \parallel r_{o2} \approx r_{s2}$$

$$R_{i3} = r_{o3} \parallel R_{i2} \approx r_{s2}$$

$$R_{i1} = g_{m1} r_{o1} \cdot R_{i3} + r_{o1} + R_{i3} = 2r_{o1} + r_{s2} \approx 2r_{o1}$$

Low frequency PSRR +



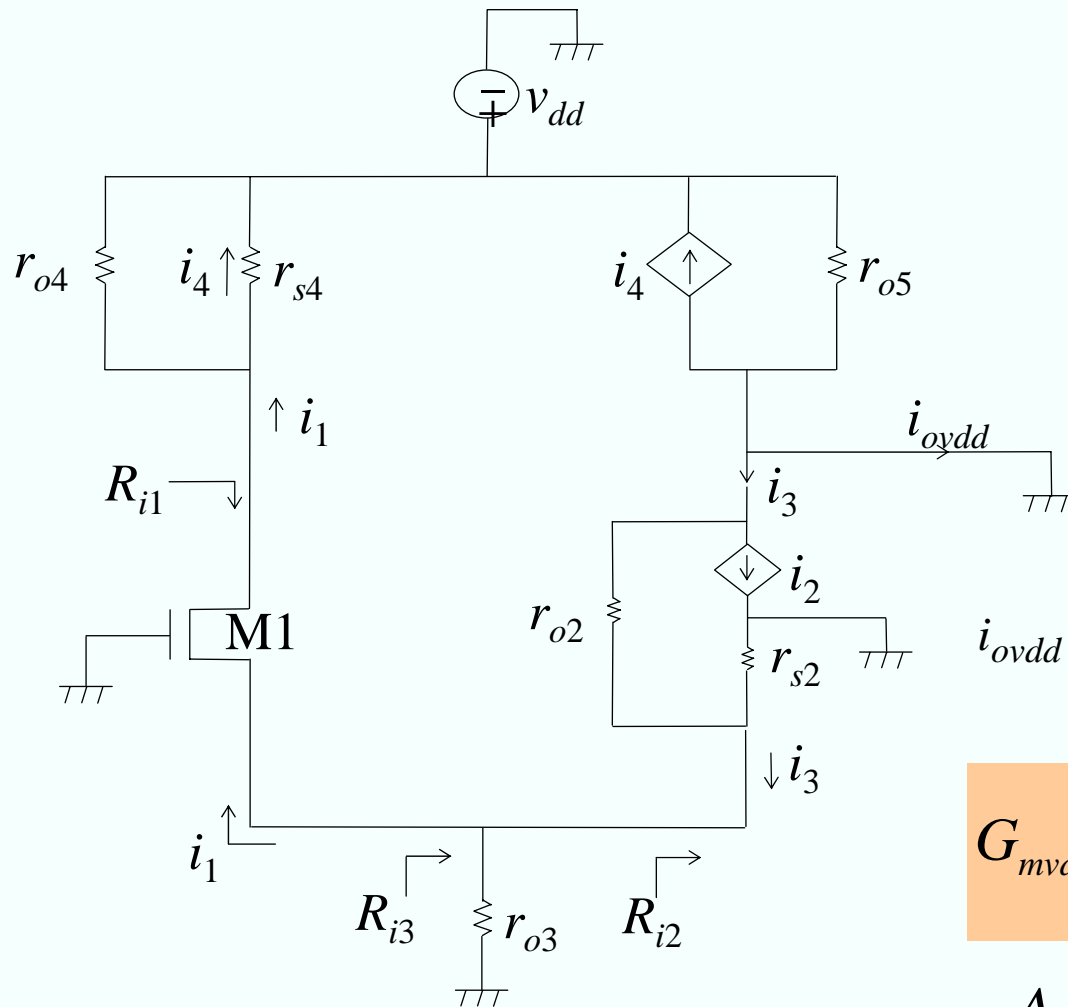
$$\begin{aligned} i_1 &= -\frac{v_{dd}}{(r_{s4} \parallel r_{o4}) + R_{i1}} = -\frac{v_{dd}}{(r_{s4} \parallel r_{o4}) + 2r_{o1}} \\ &= -\frac{v_{dd}}{r_{s4} + 2r_{o1}} = -\frac{v_{dd}}{2r_{o1}} \end{aligned}$$

$$i_4 = i_1 \cdot \frac{r_{o4}}{r_{s4} + r_{o4}} \approx i_1$$

$$i_3 = i_1 \cdot \frac{r_{o3}}{(r_{s2} \parallel r_{o2}) + r_{o3}} \approx i_1$$

$$i_{ovdd} = \frac{v_{dd}}{r_{o5}} - i_4 - i_3 = \frac{v_{dd}}{r_{o5}} - 2i_1 = v_{dd} \cdot \left(\frac{1}{r_{o5}} + \frac{1}{r_{o1}} \right) = \frac{v_{dd}}{r_{o5} \parallel r_{o1}}$$

Low frequency PSRR +



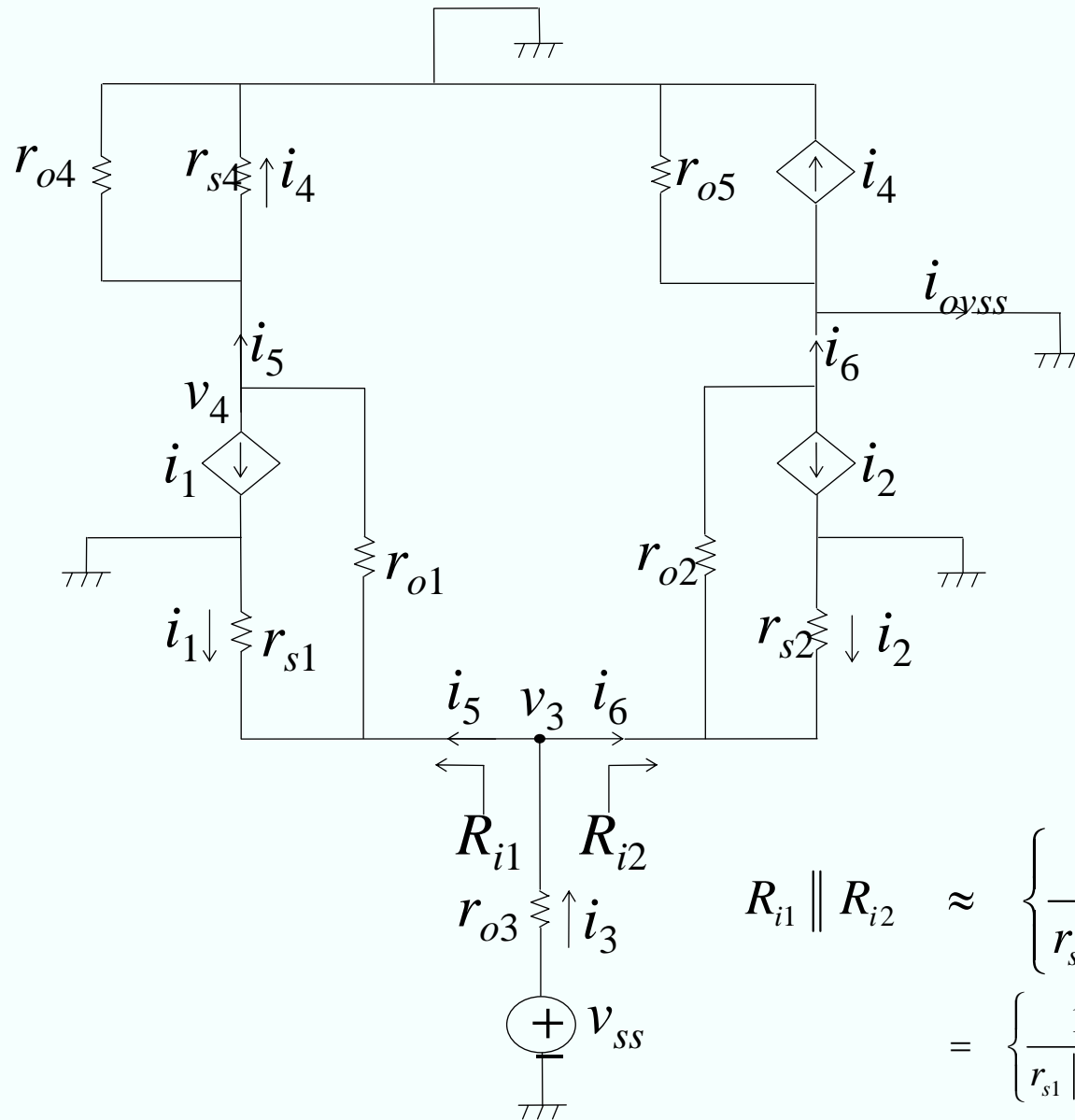
$$i_{ovdd} \approx \frac{v_{dd}}{r_{o5}} + \frac{v_{dd}}{r_{o1}} = \frac{v_{dd}}{r_{o1} \parallel r_{o5}} = \frac{v_{dd}}{r_{o2} \parallel r_{o5}}$$

$$G_{mvdd} \triangleq \frac{i_{ovdd}}{v_{dd}} = \frac{1}{r_{o2} \parallel r_{o5}} = \frac{1}{R_o}$$

$$A_{vdd} = G_{mvdd} \cdot R_o = +1$$

$$PSRR^+ \triangleq A_{vd} / A_{vdd} = A_{vd}$$

Low frequency PSRR –



$$R_{i1} = \frac{1}{g_{m1} + \frac{1}{r_{o1}}} \cdot \left(1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}} \right)$$

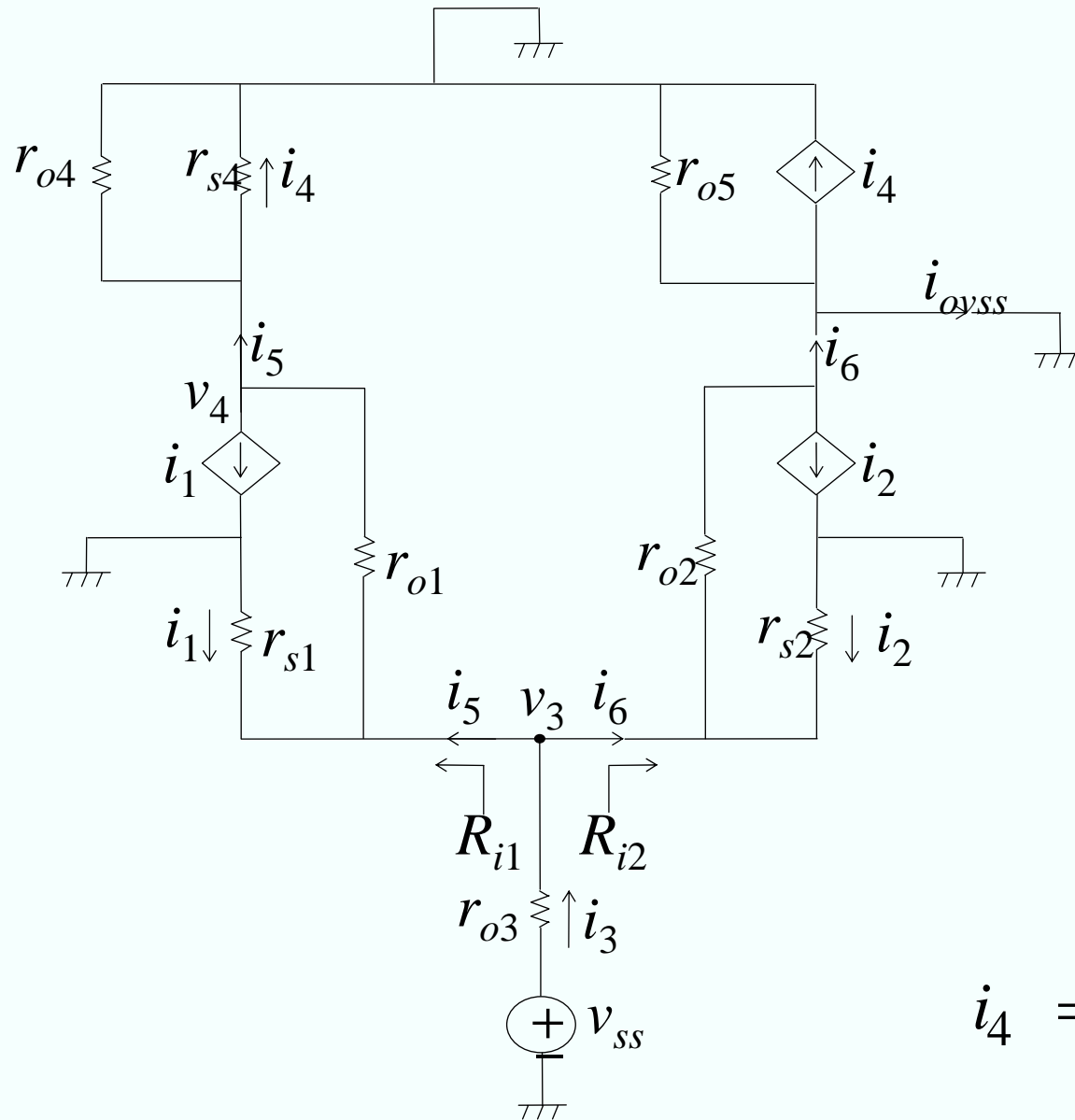
$$= (r_{s1} \parallel r_{o1}) \cdot \left(1 + \frac{r_{s4} \parallel r_{o4}}{r_{o1}} \right)$$

$$\approx (r_{s1} \parallel r_{o1}) \cdot \left(1 + \frac{r_{s4}}{r_{o1}} \right)$$

$$R_{i2} = r_{s2} \parallel r_{o2}$$

$$R_{i1} \parallel R_{i2} \approx \left\{ \frac{1}{r_{s1} \parallel r_{o1}} \cdot \left(1 - \frac{r_{s4}}{r_{o1}} \right) + \frac{1}{r_{s2} \parallel r_{o2}} \right\}^{-1}$$

$$= \left\{ \frac{1}{r_{s1} \parallel r_{o1}} \cdot \left(2 - \frac{r_{s4}}{r_{o1}} \right) \right\}^{-1} \approx \frac{r_{s1} \parallel r_{o1}}{2} \approx \frac{r_{s1}}{2}$$



Low frequency PSRR –

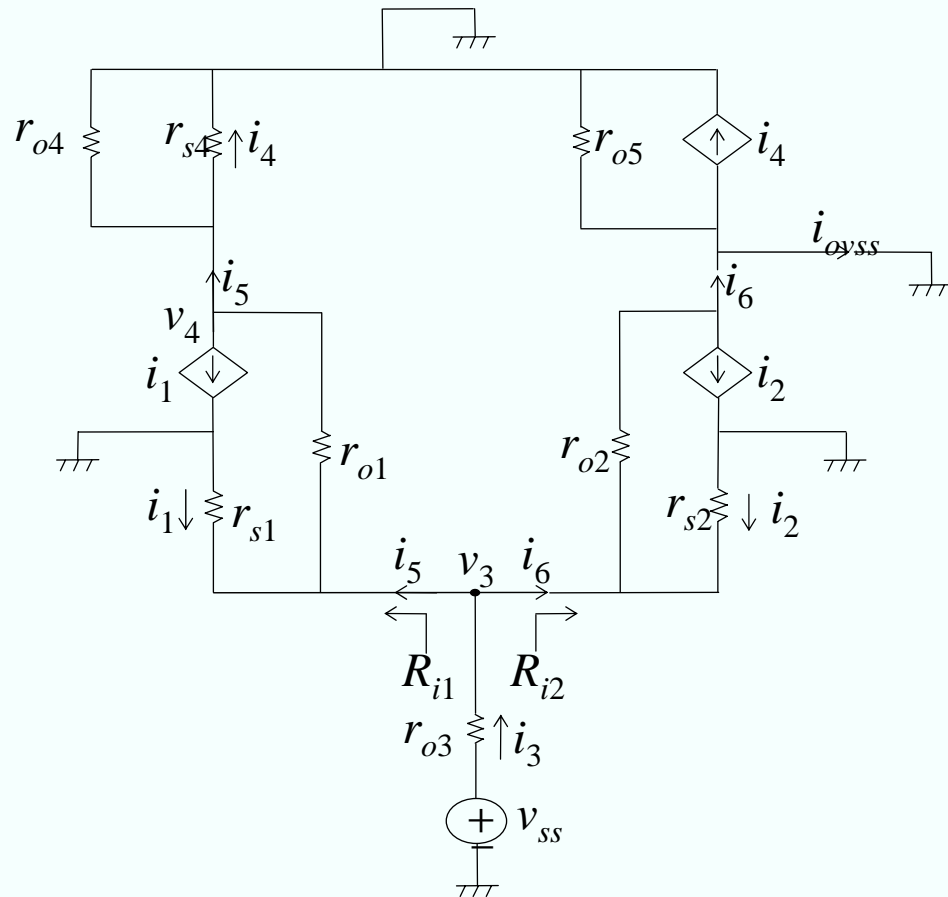
$$i_3 = \frac{v_{ss}}{r_{o3} + (R_{i1} \parallel R_{i2})}$$

$$= \frac{v_{ss}}{r_{o3} + (r_{s1}/2)} \approx \frac{v_{ss}}{r_{o3}}$$

$$i_6 = i_3 \cdot \frac{R_{i1}}{R_{i1} + R_{i2}}$$

$$i_5 = i_3 \cdot \frac{R_{i2}}{R_{i1} + R_{i2}}$$

$$i_4 = i_5 \cdot \frac{r_{o4}}{r_{s4} + r_{o4}} \approx i_5 \cdot \left(1 - \frac{r_{s4}}{r_{o4}}\right)$$



Low frequency PSRR –

$$i_4 = i_5 \cdot \frac{r_{o4}}{r_{s4} + r_{o4}} \approx i_5 \cdot \left(1 - \frac{r_{s4}}{r_{o4}}\right)$$

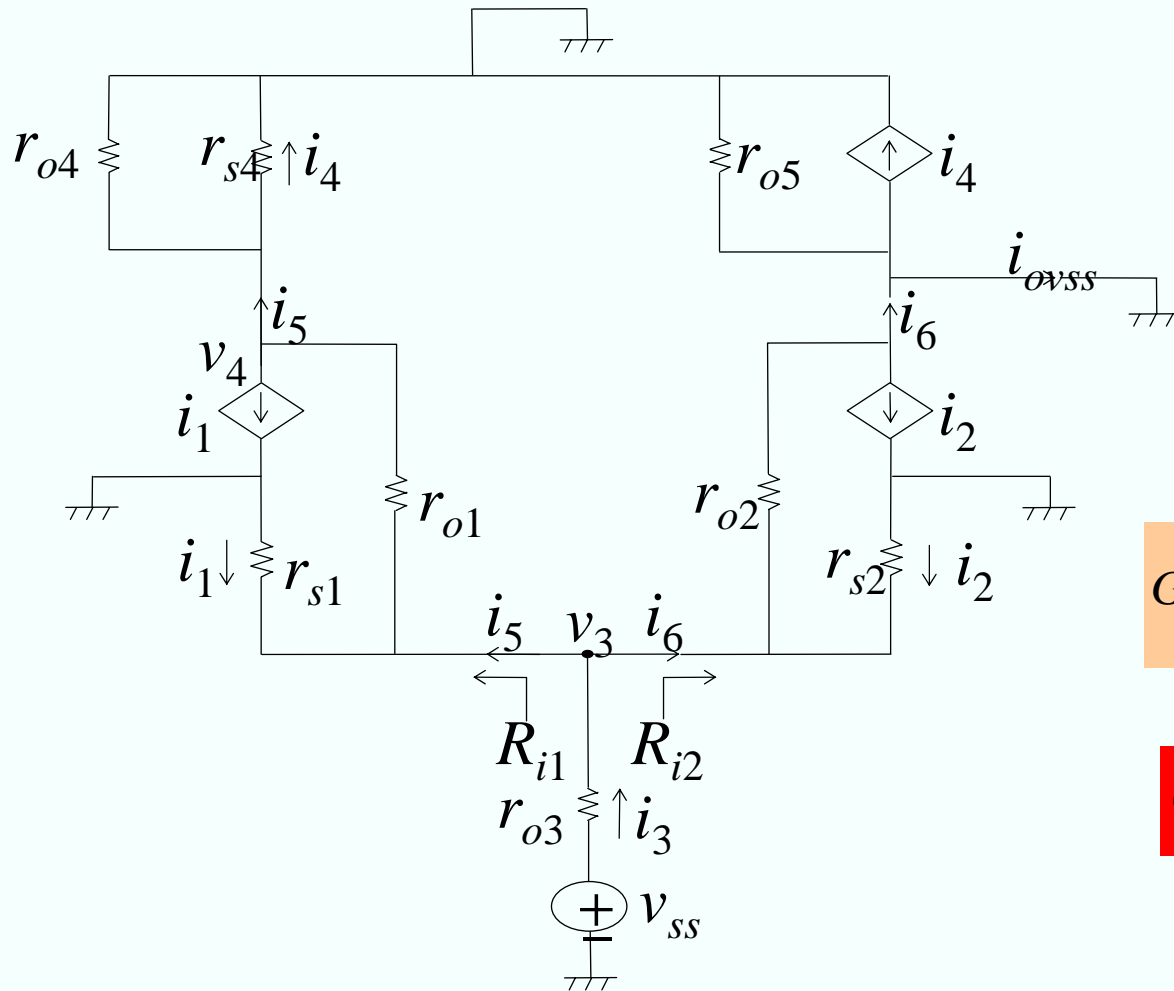
$$i_{ovss} = -i_4 + i_6$$

$$= \frac{i_3}{R_{i1} + R_{i2}} \cdot \left\{ R_{i1} - R_{i2} \cdot \left(1 - \frac{r_{s4}}{r_{o4}}\right) \right\}$$

$$= \frac{i_3}{R_{i1} + R_{i2}} \cdot (r_{s1} \parallel r_{o1}) \cdot \left(\frac{r_{s4}}{r_{o1}} + \frac{r_{s4}}{r_{o4}} \right)$$

$$\approx \frac{i_3}{2 \cdot (r_{s1} \parallel r_{o1})} \cdot (r_{s1} \parallel r_{o1}) \cdot \left(\frac{r_{s4}}{r_{o1}} + \frac{r_{s4}}{r_{o4}} \right)$$

$$\approx \frac{v_{ss}}{2 r_{o3}} \cdot \frac{r_{s4}}{(r_{o1} \parallel r_{o4})} \approx \frac{v_{ss}}{2 r_{o3} \cdot \{g_{m4}(r_{o1} \parallel r_{o4})\}}$$



Low frequency PSRR –

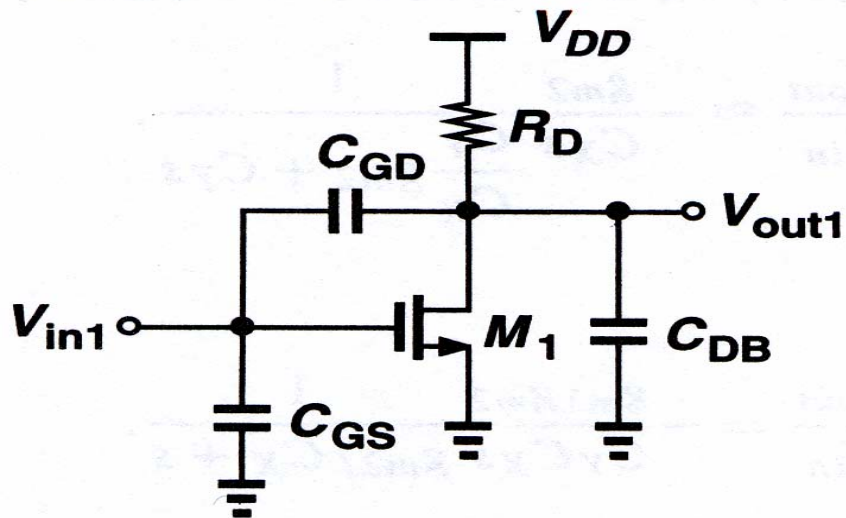
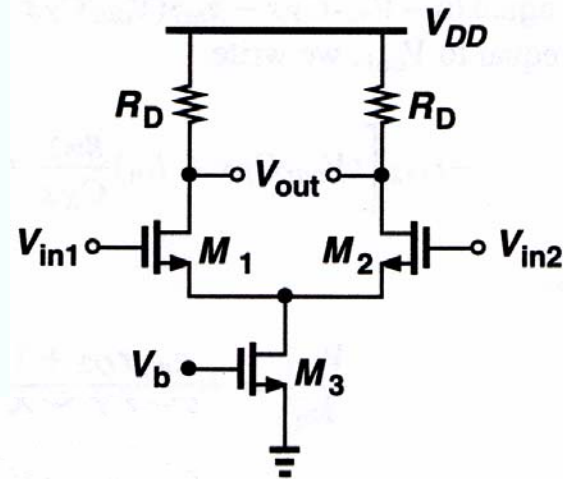
$$i_{ovss} \approx \frac{v_{ss}}{2 r_{o3} \cdot \{g_{m4} (r_{o1} \parallel r_{o4})\}}$$

$$G_{mvss} \triangleq \frac{i_{ovss}}{v_{ss}} = \frac{1}{2 r_{o3} \cdot \{g_{m4} (r_{o1} \parallel r_{o4})\}}$$

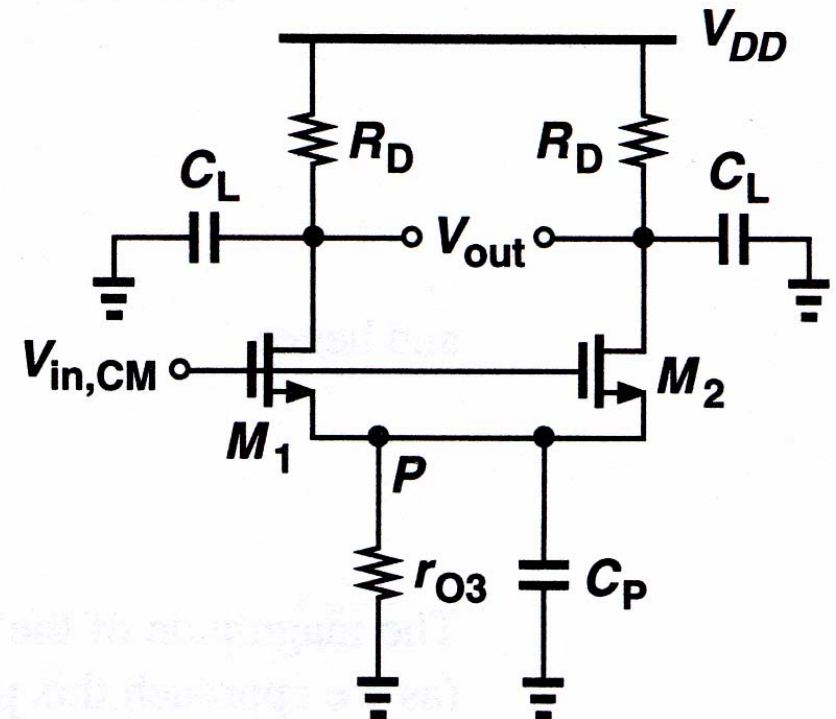
$$G_{mvss} = -G_{mc}$$

$$A_{vss} = G_{mvss} \cdot R_o = -A_{vc}$$

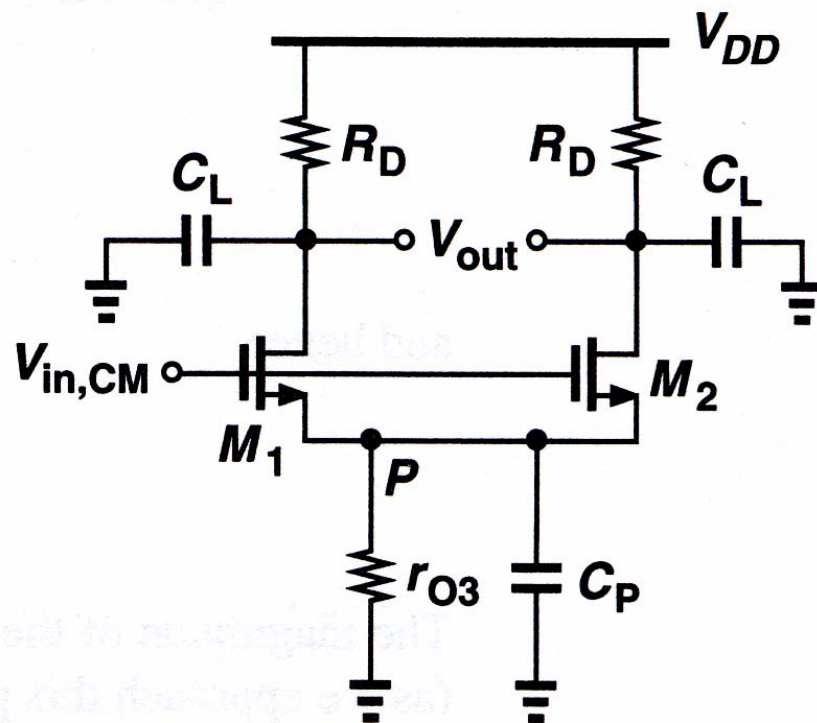
$$PSRR^- \triangleq \left| \frac{A_{vd}}{A_{vss}} \right| = \left| \frac{A_{vd}}{A_{vc}} \right| = (2g_{m1}r_{o3}) \cdot \{g_{m4} (r_{o1} \parallel r_{o4})\} = CMRR$$



Differential mode
Same as CS amp



Common mode
Pole at P decides HF response.
If this pole \ll output pole
→ Degrades HF CMRR too much



Common mode

Zero at node P : $1 / (r_{o3} * C_p)$

Pole at P decides HF response.

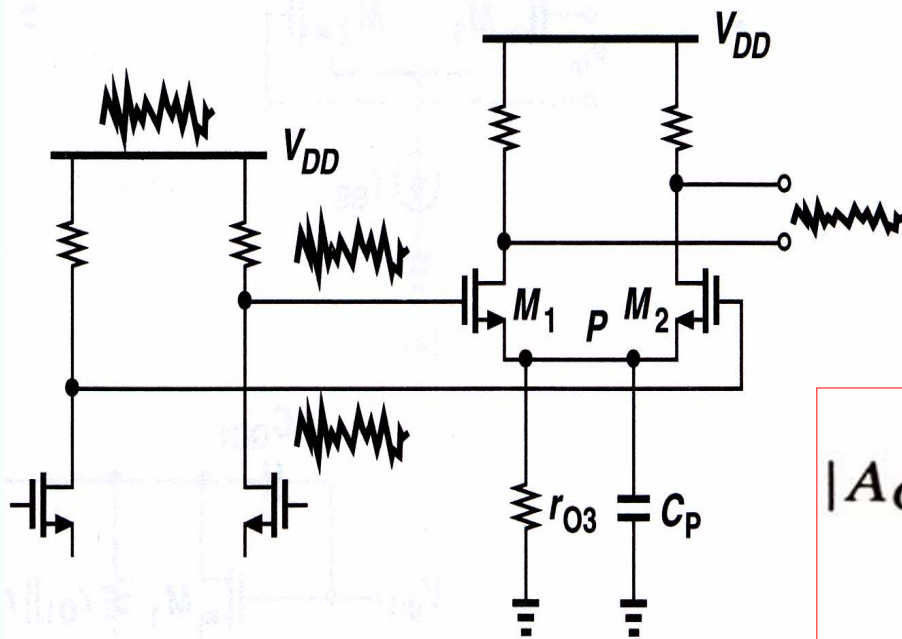
If this pole \ll output pole

→ Degrades HF CMRR too much

C_p : Reduction of HF CMRR → Mismatch causes
HF common mode noise → diff mode

Reduction of HF CMRR

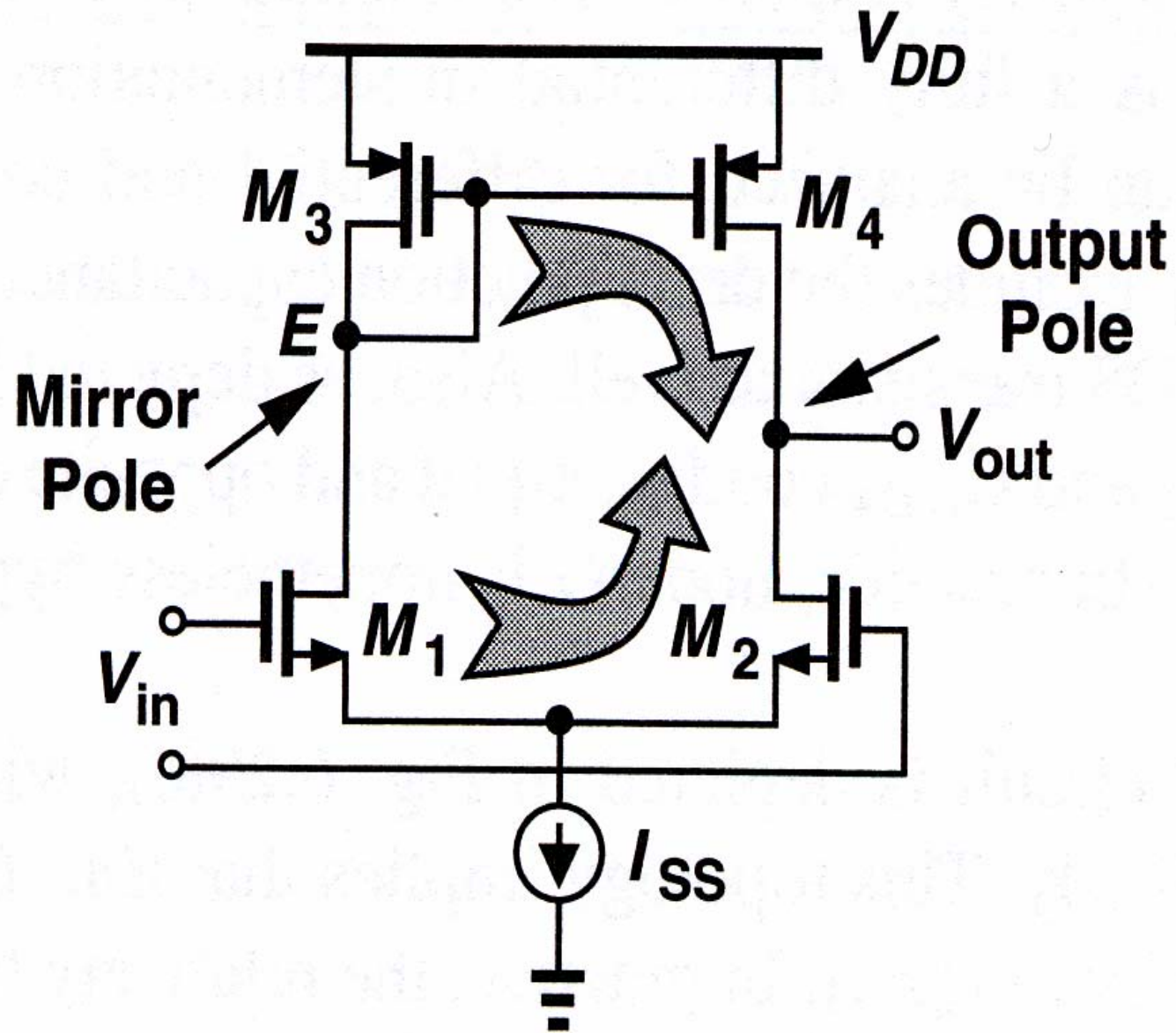
→ HF common mode noise → diff mode



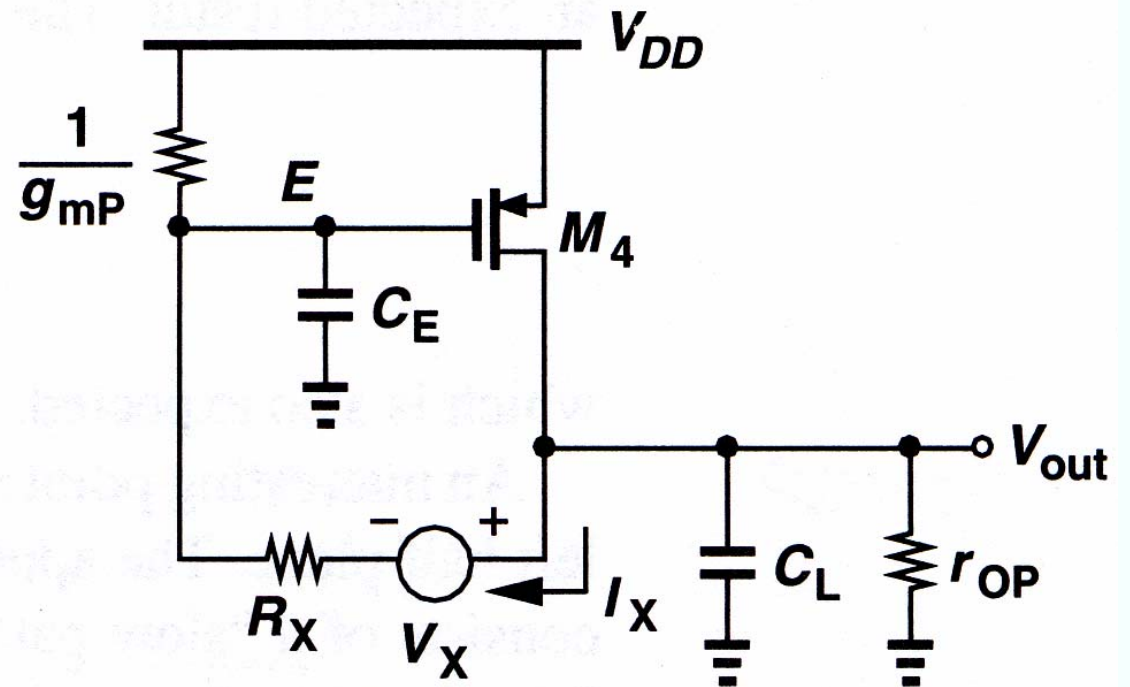
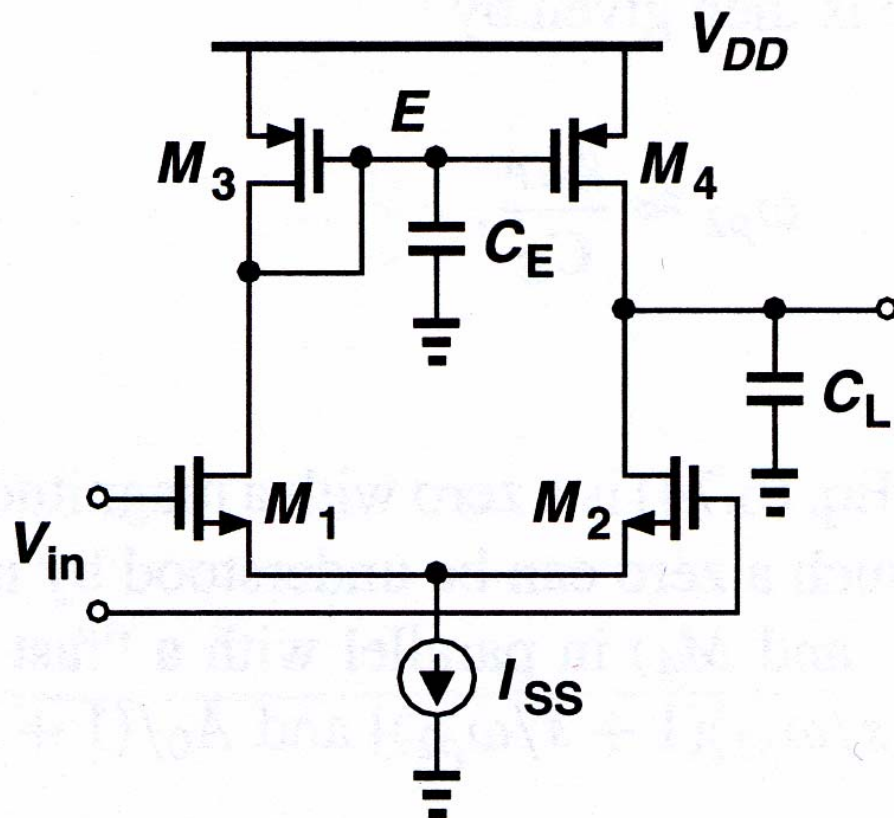
$$|A_{CM-DM}| = \frac{\Delta g_m R_D}{\sqrt{1 + (g_{m3} + g_{m4})^2 \left| \frac{1}{C_P \omega} \right|^2}}$$

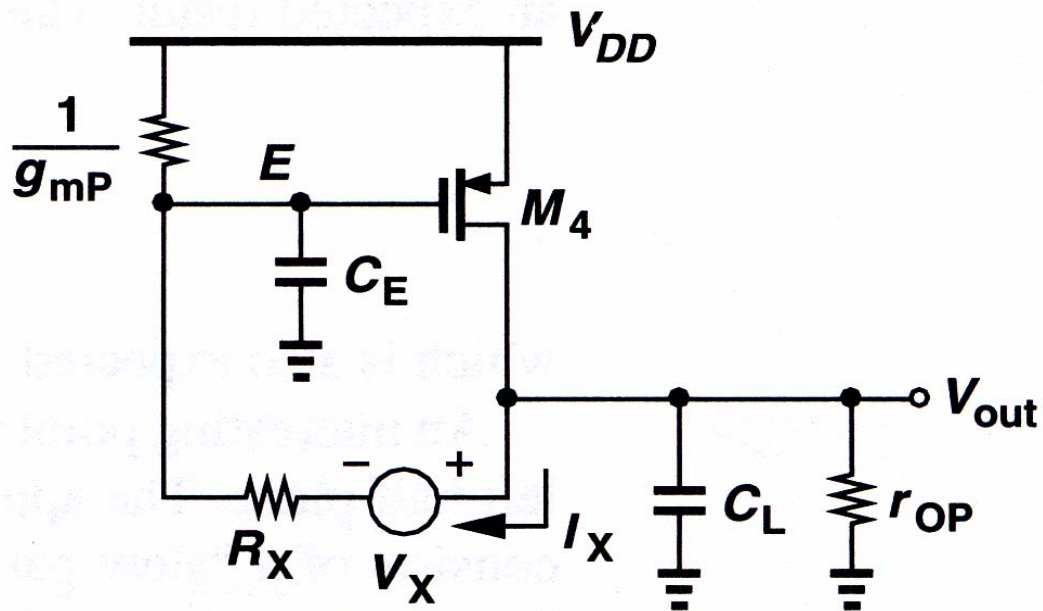
$R_{ss} \Rightarrow 1/(s C_p)$

More significant at high frequency



Mirror pole (at node E)





Mirror pole (at node E)

Thevenin equivalent ckt

$$V_X = g_{mN} r_{oN} V_{in}$$

$$R_X = 2 r_{oN}$$

$$I_x = \frac{v_E}{\frac{1}{g_{mP} + s C_E}}$$

$$V_E = (V_{out} - V_X) \frac{\frac{1}{C_E s + g_{mP}}}{\frac{1}{C_E s + g_{mP}} + R_X}$$

$$-g_{m4} v_E - I_x = v_{out} \cdot \left(s C_L + \frac{1}{r_{oP}} \right)$$

$$\frac{V_{out}}{V_{in}} = \frac{g_{mN} r_{ON} (2g_{mP} + C_{ES})}{2r_{OP} r_{ON} C_E C_L s^2 + [(2r_{ON} + r_{OP}) C_E + r_{OP} (1 + 2g_{mP} r_{ON}) C_L] s + 2g_{mP} (r_{ON} + r_{OP})}$$

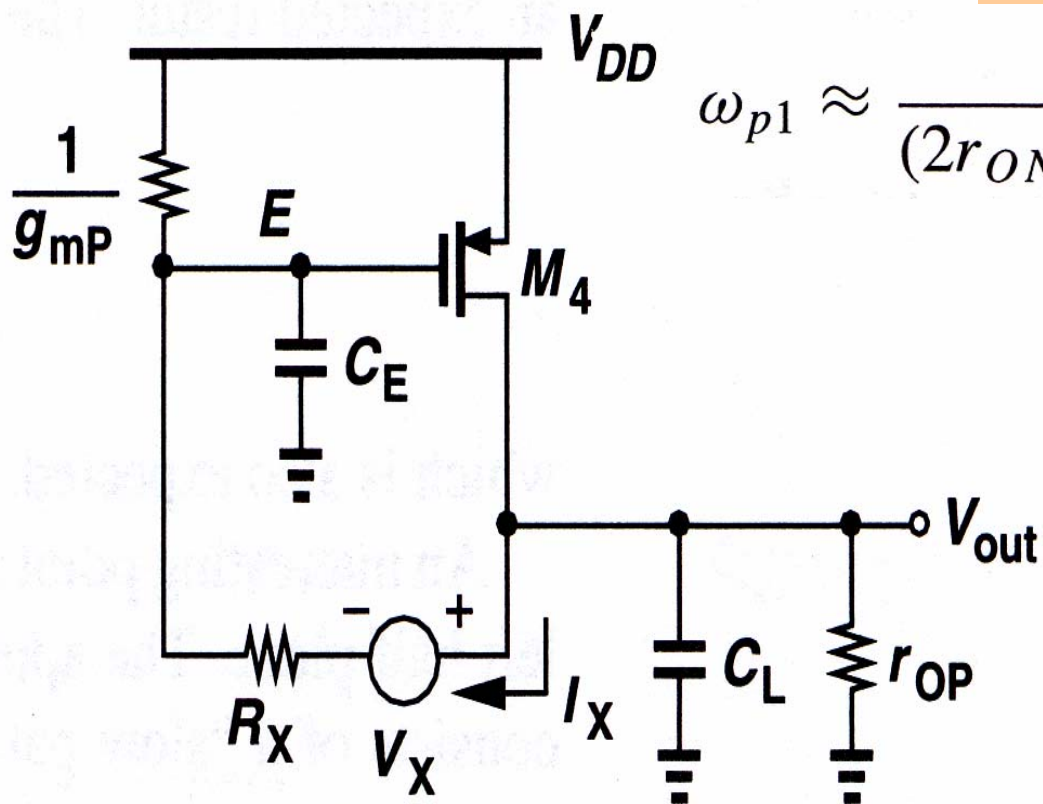
Dominant pole

$$\omega_{p1} \approx \frac{2g_{mP} (r_{ON} + r_{OP})}{(2r_{ON} + r_{OP}) C_E + r_{OP} (1 + 2g_{mP} r_{ON}) C_L}$$

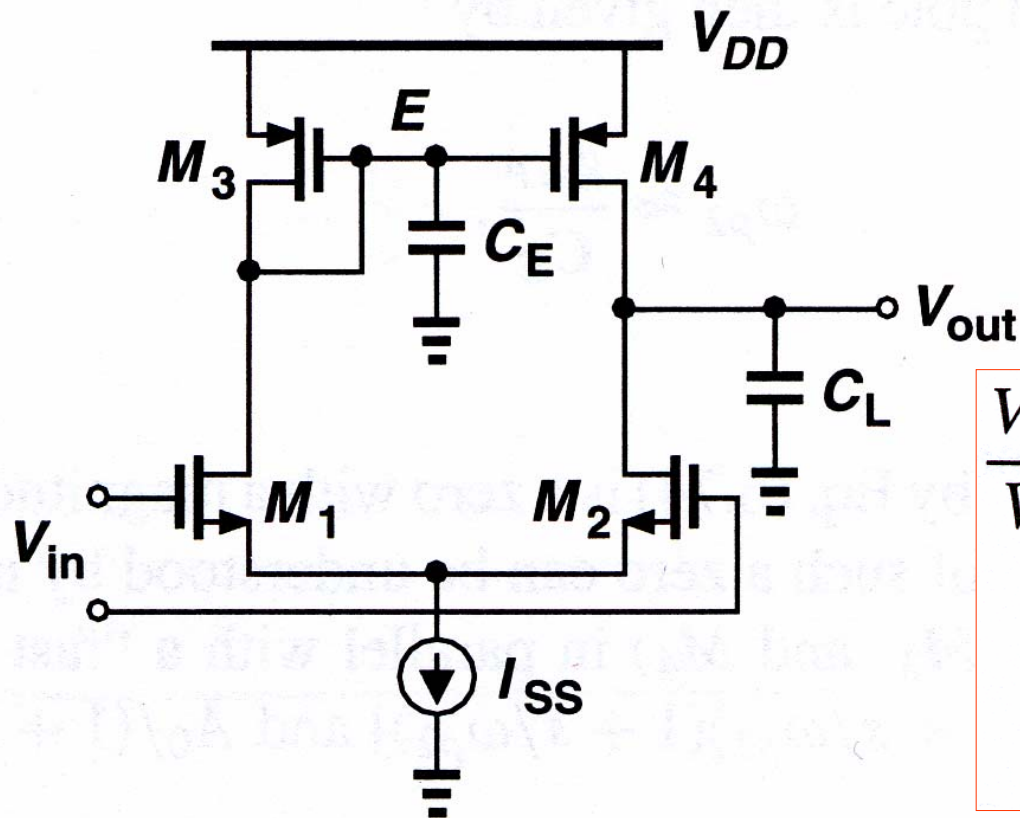
$$\omega_{p1} \approx \frac{1}{(r_{ON} \parallel r_{OP}) C_L}$$

Mirror pole (at node E)

$$\omega_{p2} \approx \frac{g_{mP}}{C_E}$$



$$\frac{V_{out}}{V_{in}} = \frac{g_{mN} r_{OP} (2g_{mP} + C_E s)}{2r_{OP} r_{ON} C_E C_L s^2 + [(2r_{ON} + r_{OP}) C_E + r_{OP} (1 + 2g_{mP} r_{ON}) C_L] s + 2g_{mP} (r_{ON} + r_{OP})}$$



Negative real zero
 Due to fast path (M1, M2) +
 slow path (M1, M3, M4)
 $A_v(s) = A_{slow}(s) + A_{fast}(s)$

$$\frac{V_{out}}{V_{in}} = \frac{A_0}{1 + s/\omega_{p1}} \left(\frac{1}{1 + s/\omega_{p2}} + 1 \right)$$

$$= \frac{A_0(2 + s/\omega_{p2})}{(1 + s/\omega_{p1})(1 + s/\omega_{p2})}$$

Differential pair with R load

- (1) Large signal transfer curve
- (2) Small signal characteristic (G_{md} , G_{mc} , R_o)
- (3) Active input common mode range
- (4) Input offset voltage
- (5) Common mode to differential conversion due to mismatches

Differential pair with diode load

Differential pair with active load

Differential pair: frequency response