

(a)

(b)

$$\omega_{p1} = \frac{1}{(g_m r_o) \cdot (R_s C_{GD}) + R_s (C_1 + C_{GD}) + r_o (C_L + C_{GD})}$$

$$\omega_{p1} \approx \frac{1}{R_s C_{GD} g_m r_o}$$

$$\omega_{-3dB} = \frac{1}{R_s \cdot [C_{GD} \cdot \{1 + g_m \cdot (r_o \parallel R_L)\} + C_{GS} + C_{GB}]}$$

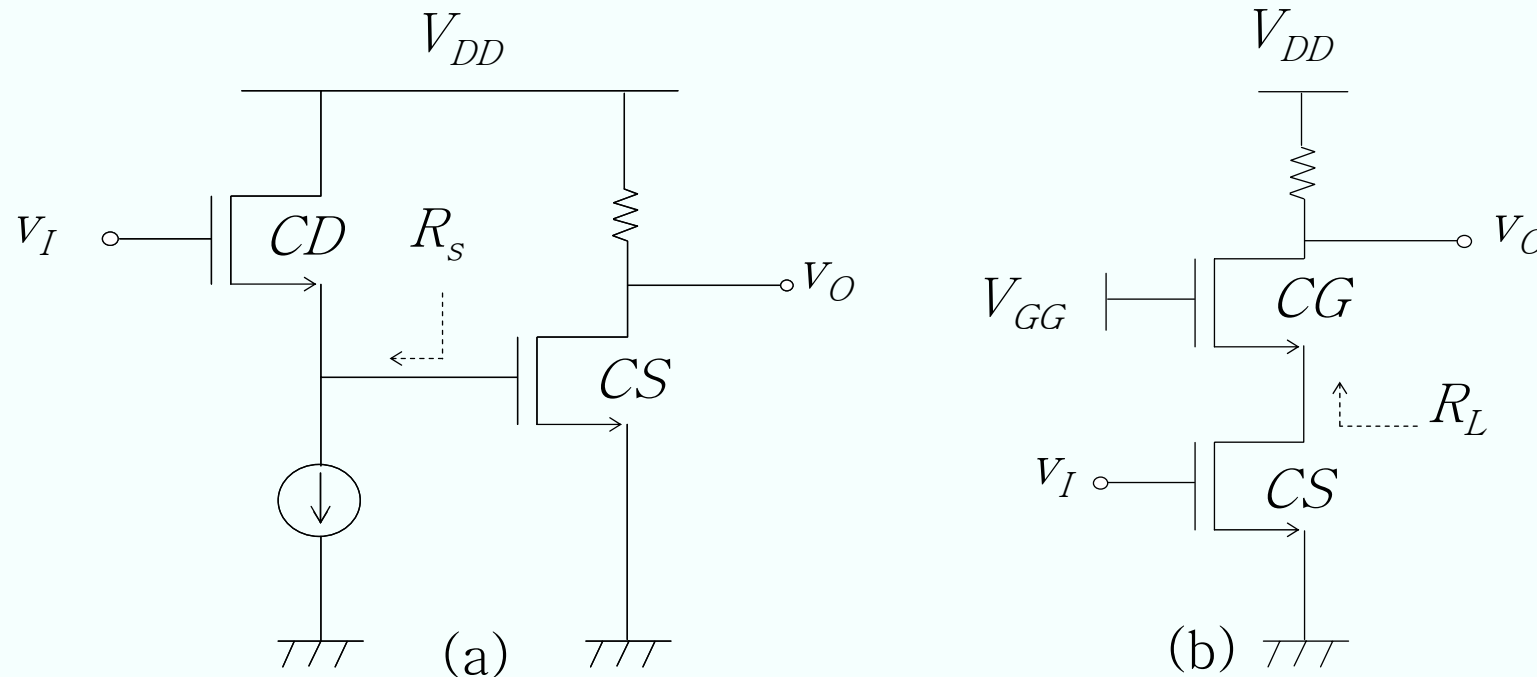


Fig 4.5.2 Circuits to improve the frequency response of CS amp

(a) CD-CS series connection (cascode) => reduce R_s

(b) CS-CG series connection (cascode) => reduce voltage gain by decreasing R_L

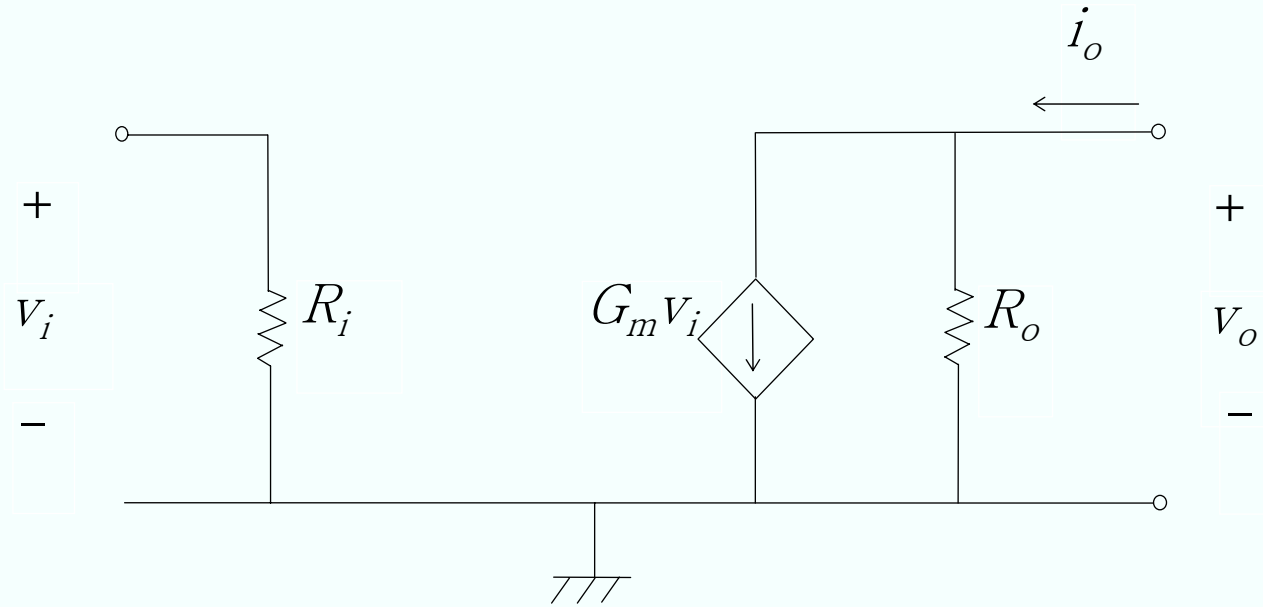


Fig 4.6.1 Low frequency small-signal equivalent circuit of general amplifier

$$G_m \triangleq \left. \frac{i_o}{v_i} \right|_{v_o = 0}$$

$$R_o \triangleq \left. \frac{v_o}{i_o} \right|_{v_i = 0}$$

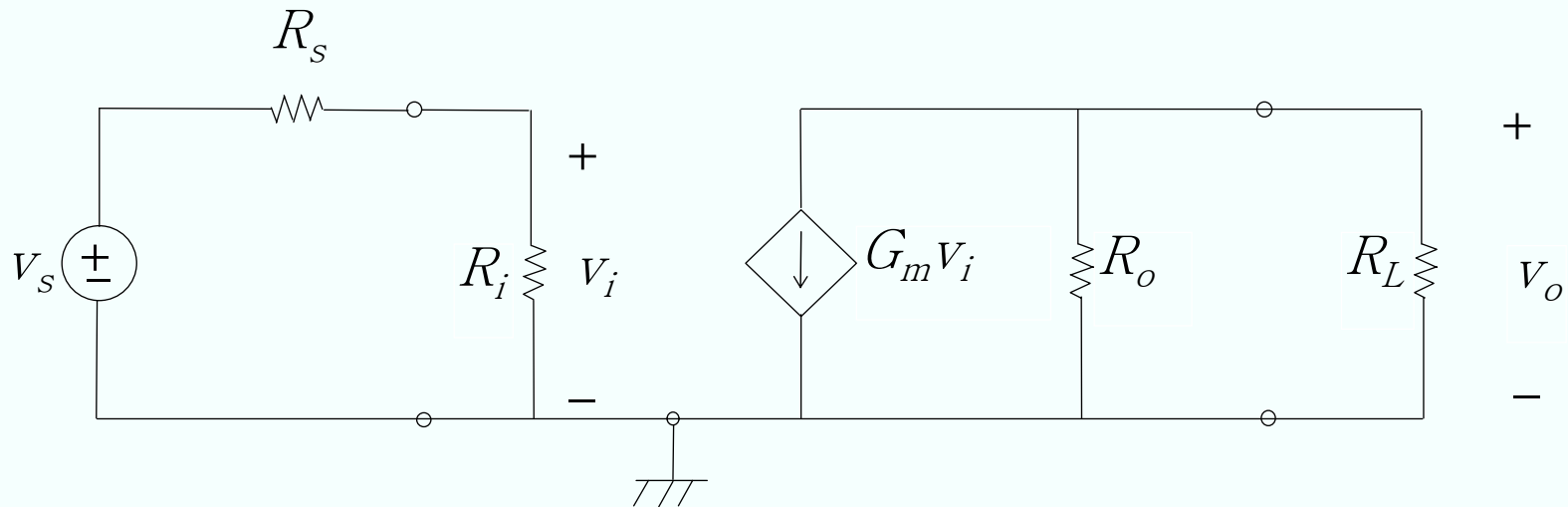
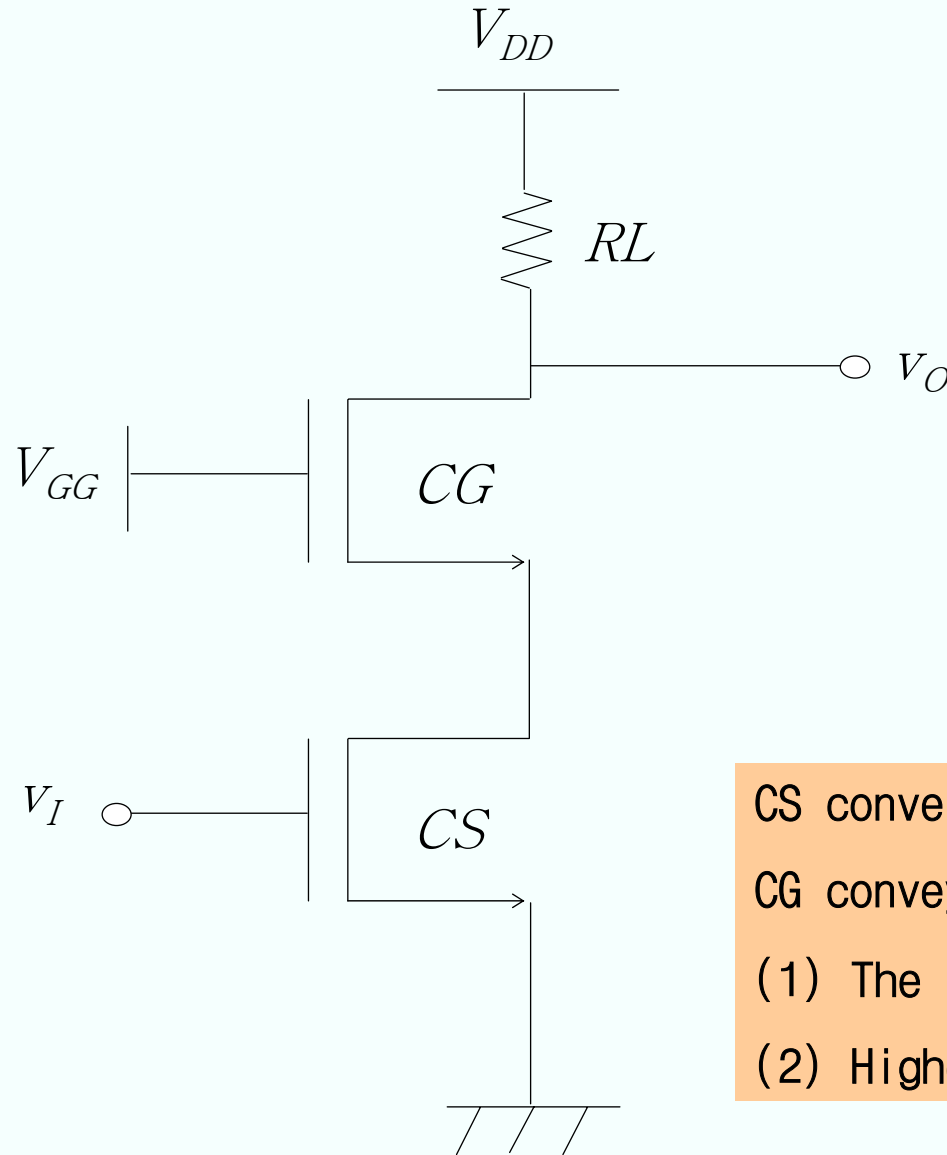


Fig 4.6.2 General amp with a voltage source input and a resistor load

$$A_v = \frac{v_o}{v_s} = -G_m \cdot (R_o \parallel R_L) \cdot \frac{R_i}{R_s + R_i}$$



CS converts voltage to current ($V \rightarrow I$),
CG conveys current to load ($I \rightarrow I$)

- (1) The same voltage gain as a single CS amp
- (2) Higher bandwidth than a single CS amp

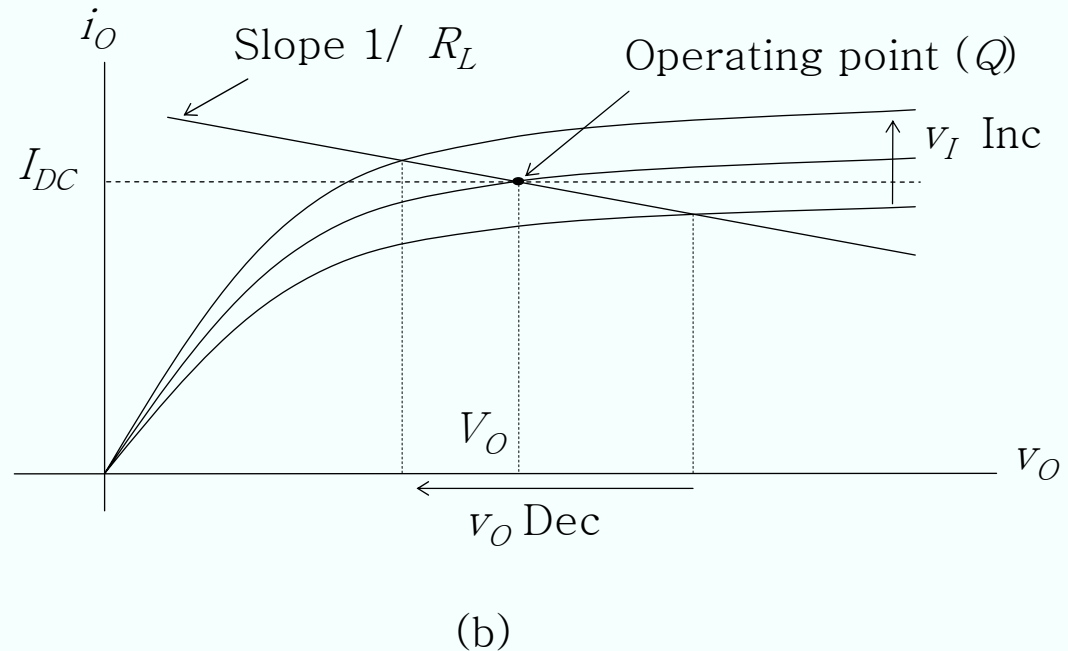
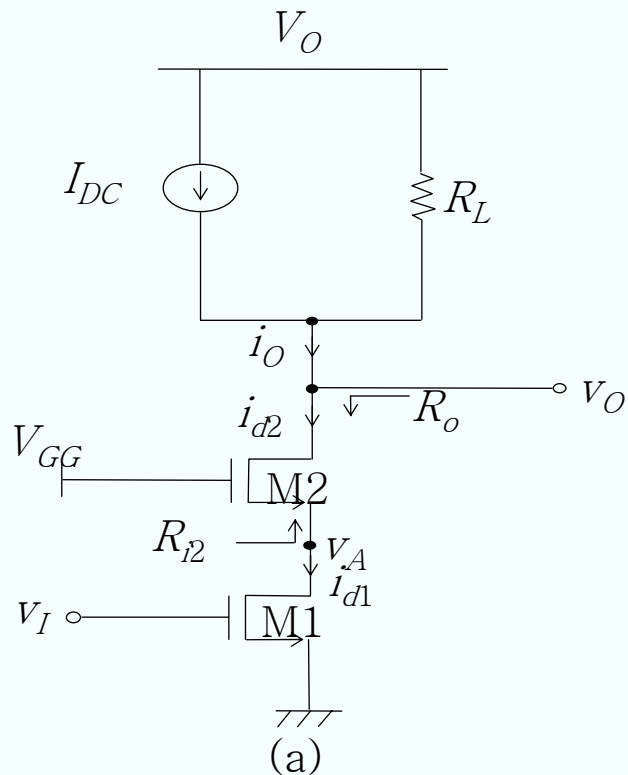
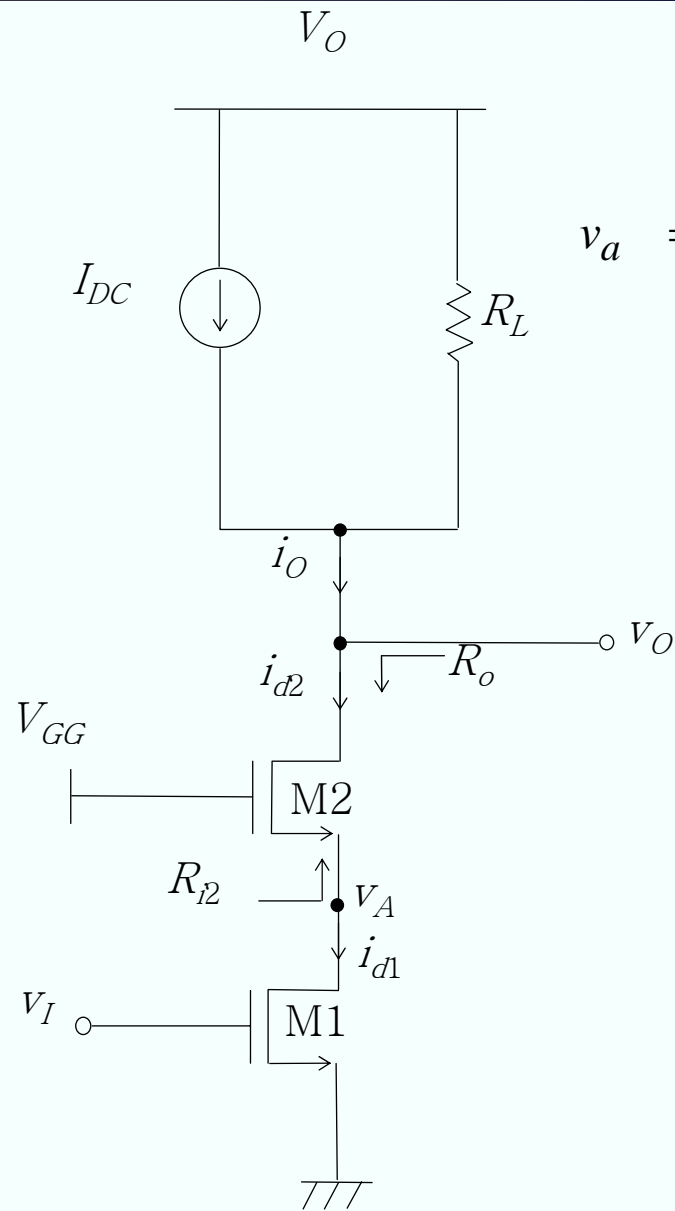


Fig 4.6.3 Cascode amplifier (a) circuit (b) operating point computation

$$v_A = V_{GG} - v_{GS2} \approx V_{GG} - V_{GS2} = V_{GG} - V_{DSAT2} - V_{TH2} = V_{GG} - \sqrt{\frac{2 \cdot I_{DC}}{\mu_n C_{ox} \cdot \left(\frac{W}{L}\right)_2}} - V_{TH2}$$

For max V_O headroom, set $V_{GG} = 2 V_{DSAT} + V_{TH}$, $V_A = V_{DSAT} (= V_I - V_{TH}) \Rightarrow V_O > 2 V_{DSAT}$, but little headroom for V_A (problem) \rightarrow Set $V_{GG} = 2 V_{DSAT} + V_{TH} + V_{margin}$



$$i_{d1} = g_{m1} v_i$$

$$v_a = -i_{d1} \cdot (r_{o1} \parallel R_{i2}) = -i_{d1} \cdot (r_{o1} \parallel r_{s2}) \approx -i_{d1} r_{s2} = -\frac{g_{m1}}{g_{m2}} \cdot v_i$$

$$v_a = -\frac{g_{m1}}{g_{m2}} \cdot v_i = -\sqrt{\frac{(W/L)_1}{(W/L)_2}} \cdot v_i$$

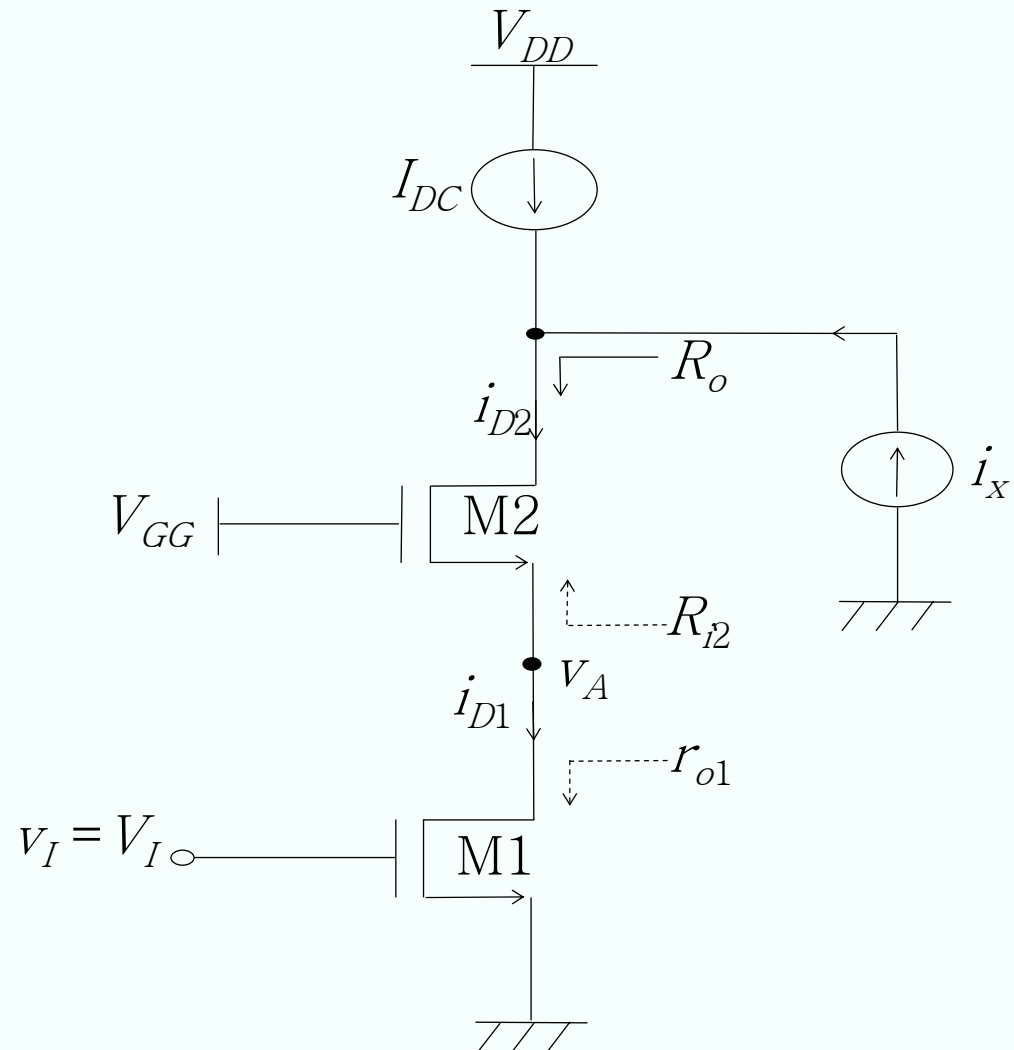
$$\text{if } (W/L)_1 = (W/L)_2 \quad v_a/v_i \approx -1$$

$$2C_{GD1} \quad \text{Due to Miller effect}$$

$$v_a = -i_{d1} \cdot (r_{o1} \parallel R_{i2}) = -i_{d1} \cdot r_{o1} = -g_{m1} r_{o1} \cdot v_i$$

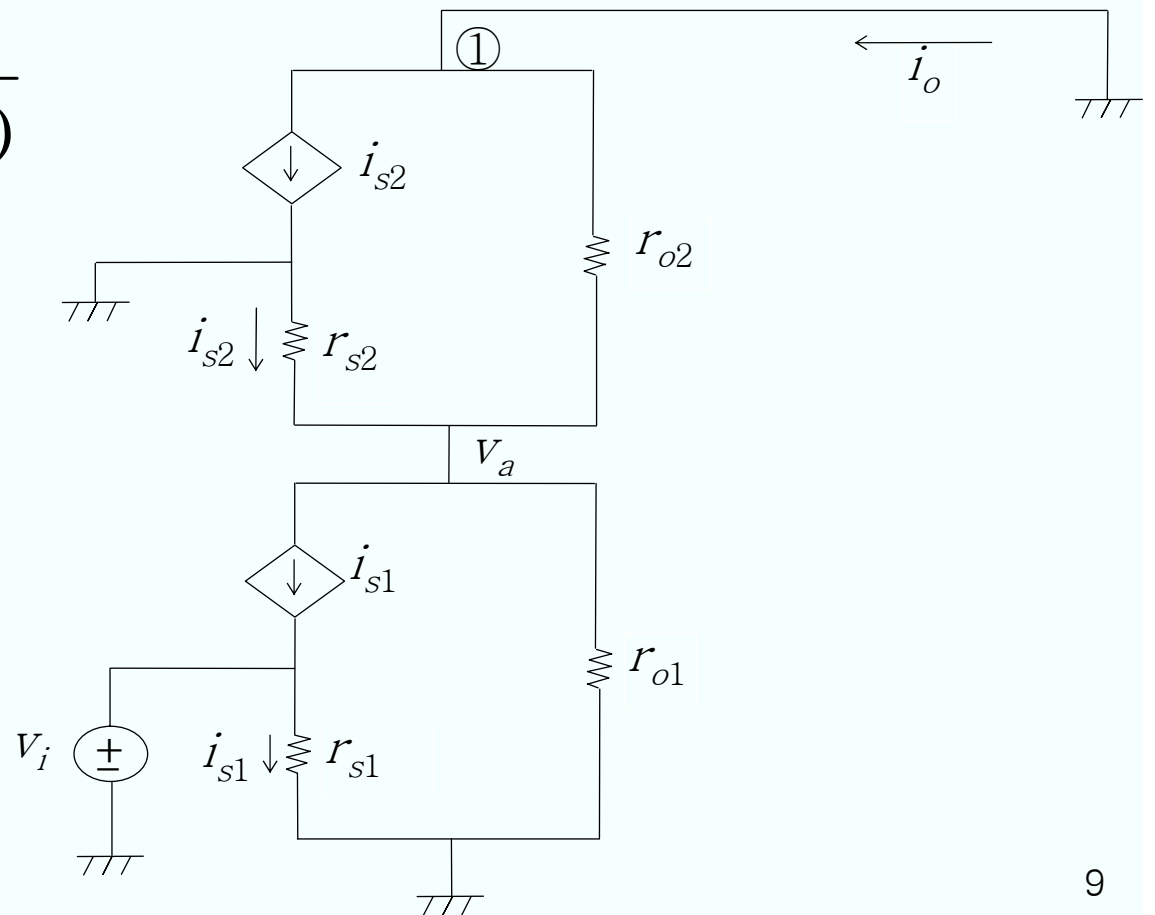
$$v_o = -G_m R_o \cdot v_i$$

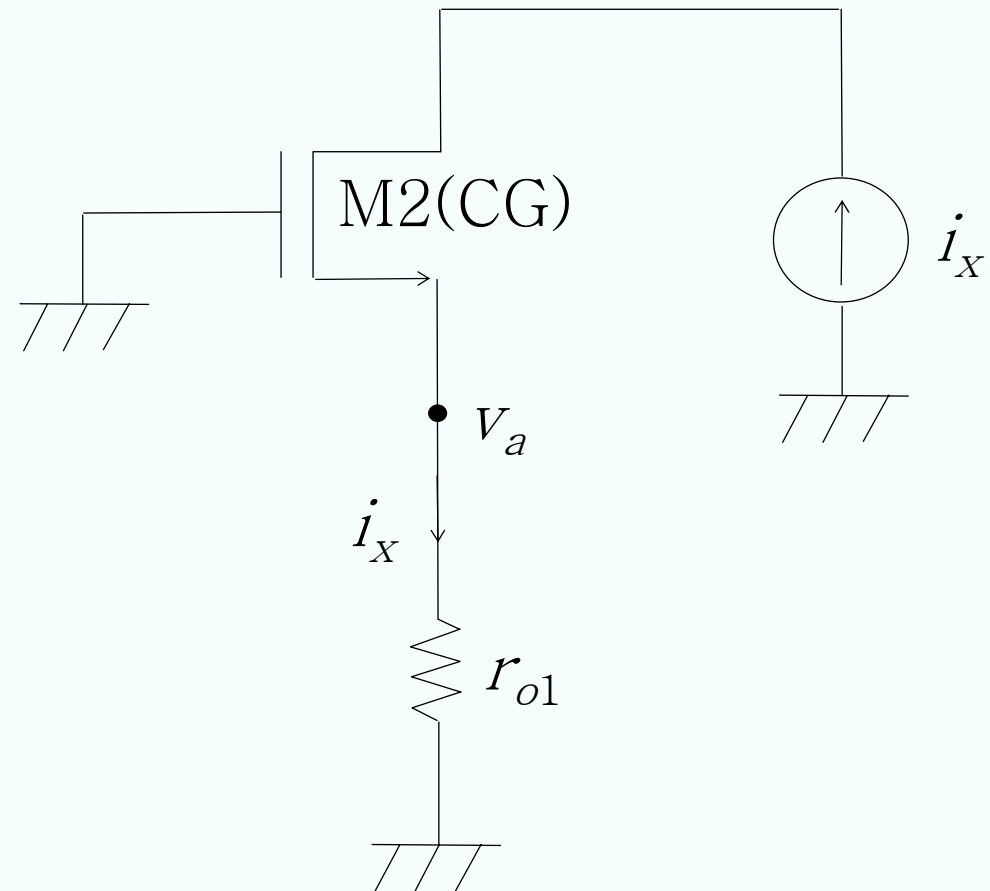
$$G_m \approx g_{m1}$$



$$G_m = \frac{i_o}{v_i} = g_{m1} \cdot \frac{r_{o1}}{r_{o1} + R_{i2}} \approx g_{m1}$$

$$R_{i2} = \frac{1 + (R_L / r_{o2})}{g_{m2} + g_{mb2} + (1 / r_{o2})}$$



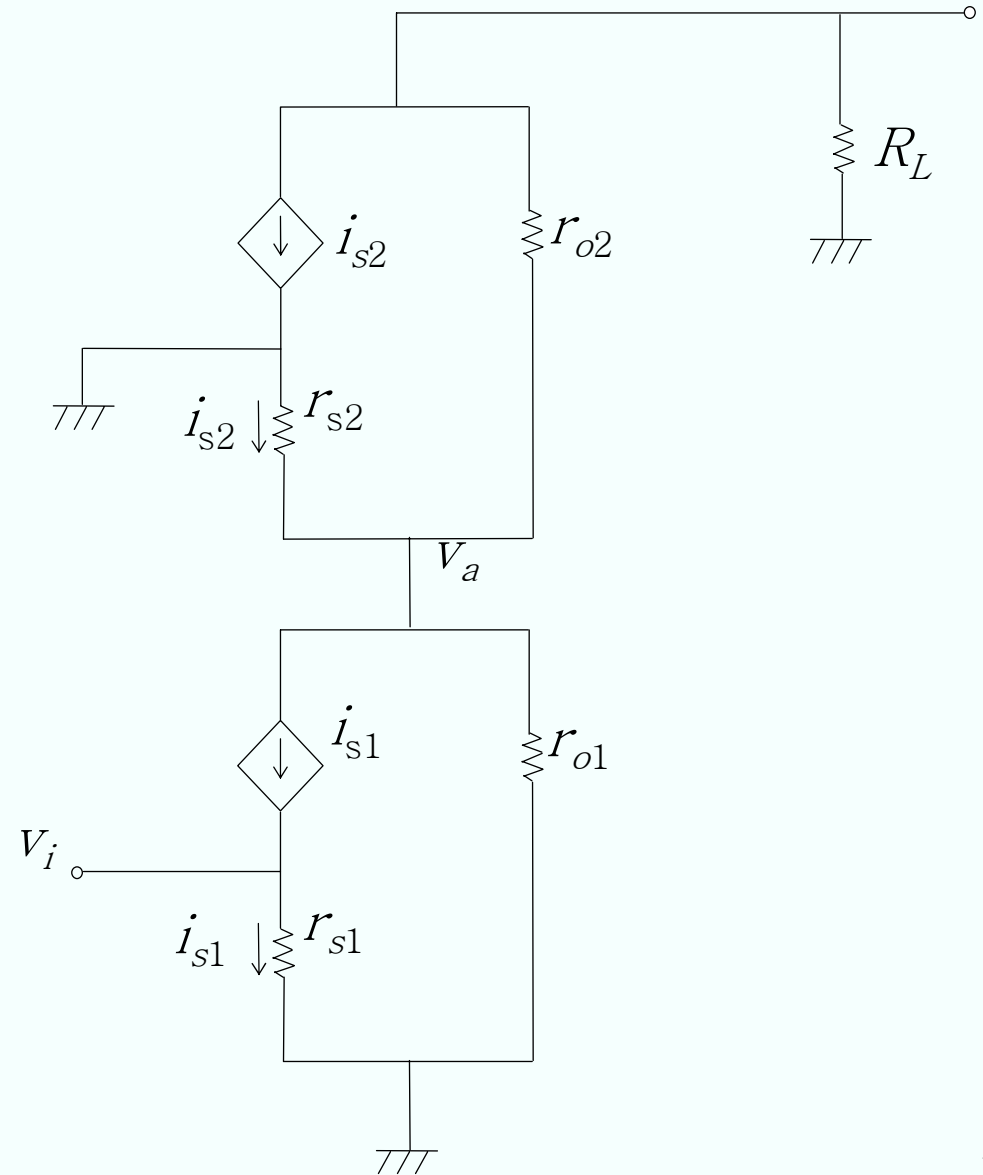
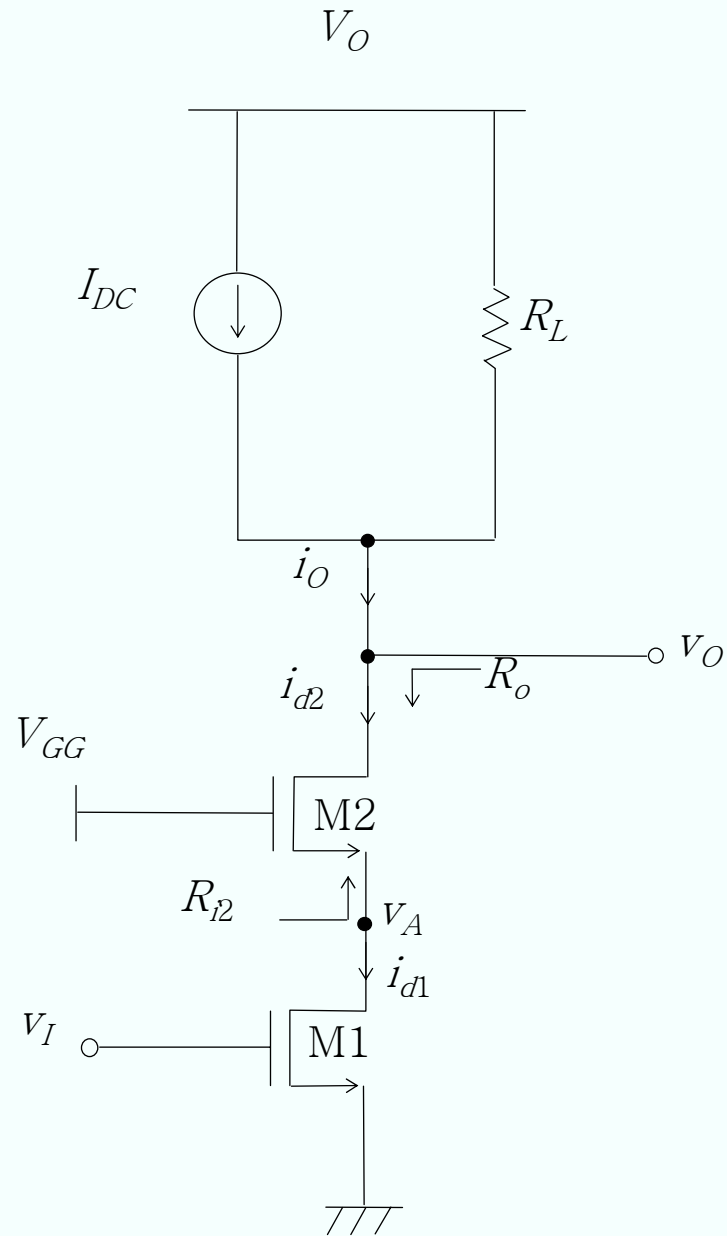


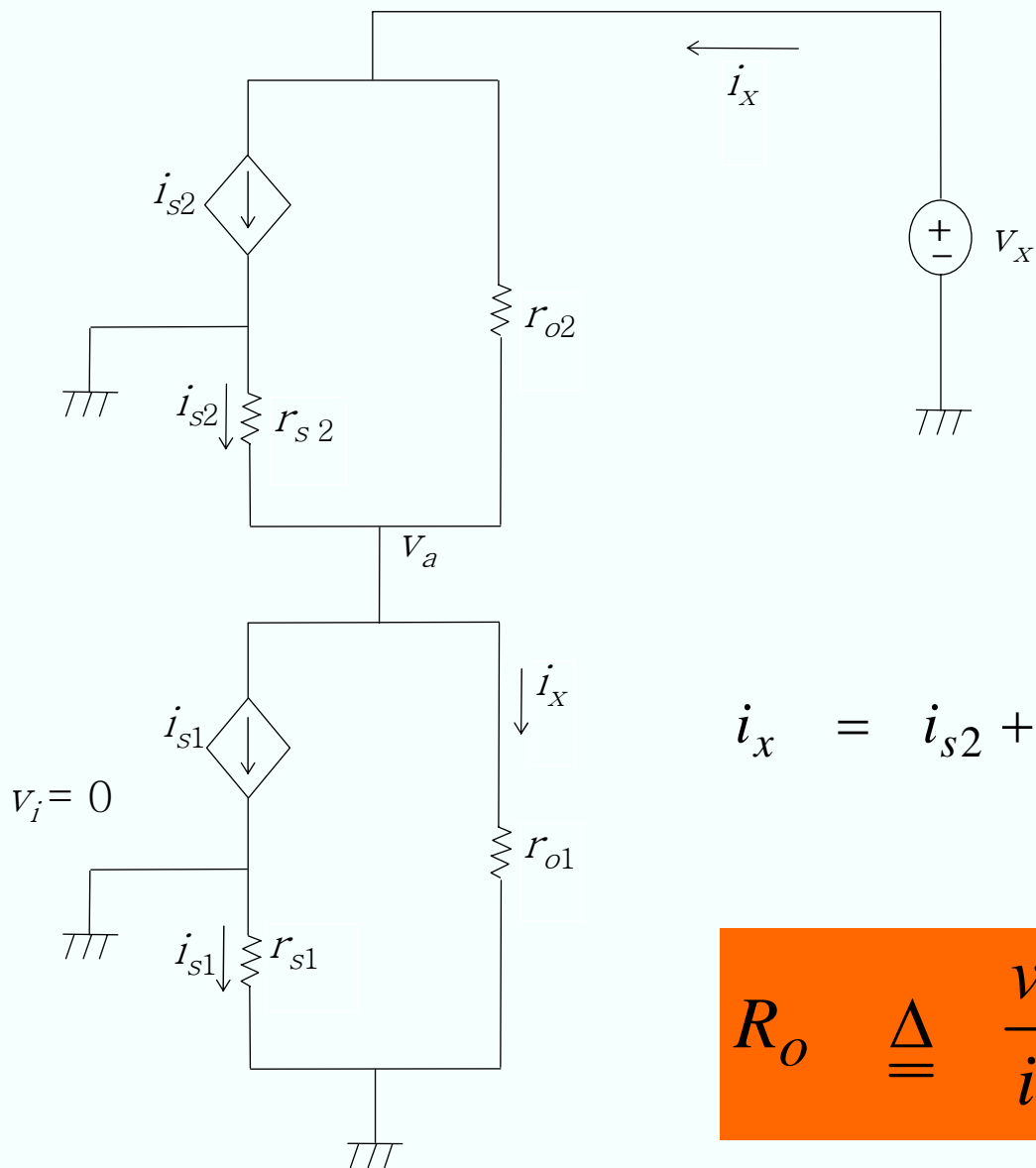
$$v_a = v_{ds1} = i_x \cdot r_{o1}$$

$$v_o \approx (g_{m2} + g_{mb2}) r_{o2} \cdot v_a = (g_{m2} + g_{mb2}) r_{o2} \cdot r_{o1} \cdot i_x$$

$$R_o = \frac{v_o}{i_x}$$

$$R_o \approx g_{m2} r_{o2} \cdot r_{o1}$$





$$i_x = i_{s2} + \frac{v_x - v_a}{r_{o2}} = -g_{m2}r_{o1} \cdot i_x + \frac{v_x - i_x r_{o1}}{r_{o2}}$$

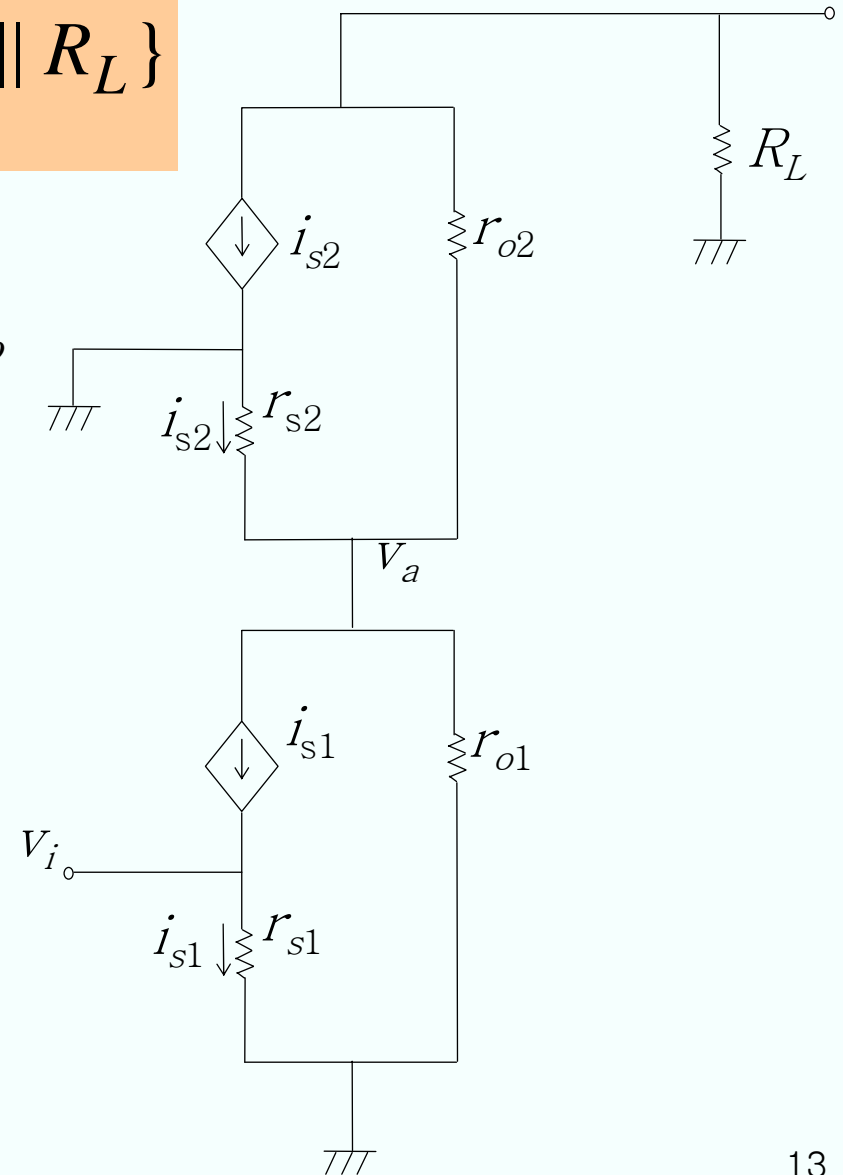
$$R_o \triangleq \frac{v_x}{i_x} = g_{m2}r_{o2} \cdot r_{o1} + r_{o1} + r_{o2}$$

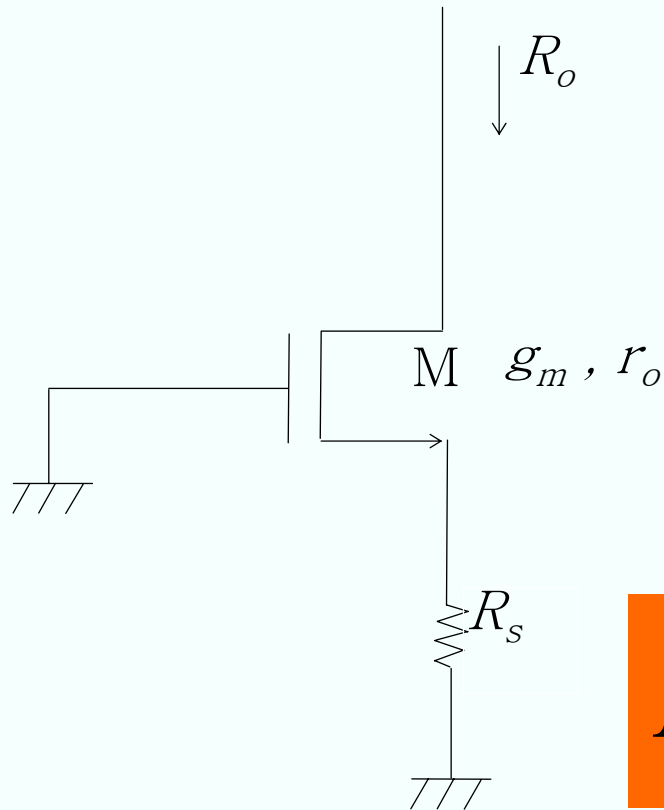
$$A_v = \frac{v_o}{v_i} \approx -g_{m1} \cdot \{ (r_{o1} g_{m2} r_{o2}) \parallel R_L \}$$

$$v_a = (R_{i2} \parallel r_{o1}) \cdot g_{m1} v_i \approx \frac{R_{i2} \parallel r_{o1}}{(g_{m2} r_{o2} r_{o1}) \parallel R_L} \cdot v_o$$

$$R_{i2} = \frac{1 + (R_L / r_{o2})}{g_{m2} + g_{mb2} + (1 / r_{o2})}$$

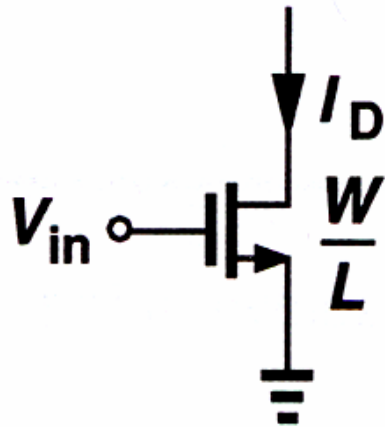
$$v_a \approx \frac{v_o}{g_{m2} r_{o2}} \quad \text{if } R_L = \infty$$



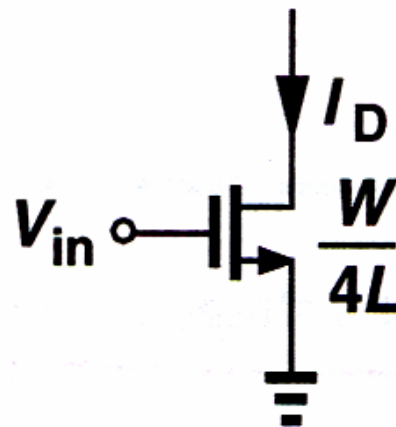


$$R_o = g_m r_o \cdot R_s + R_s + r_o$$

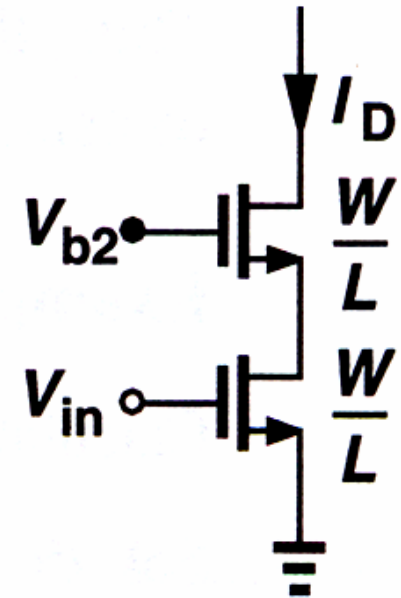
$$R_o = (g_m + g_{mb}) r_o \cdot R_s + R_s + r_o$$



(a)



(b)

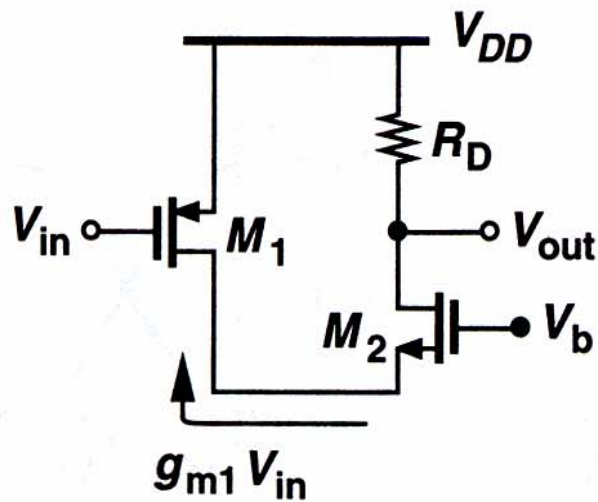


(c)

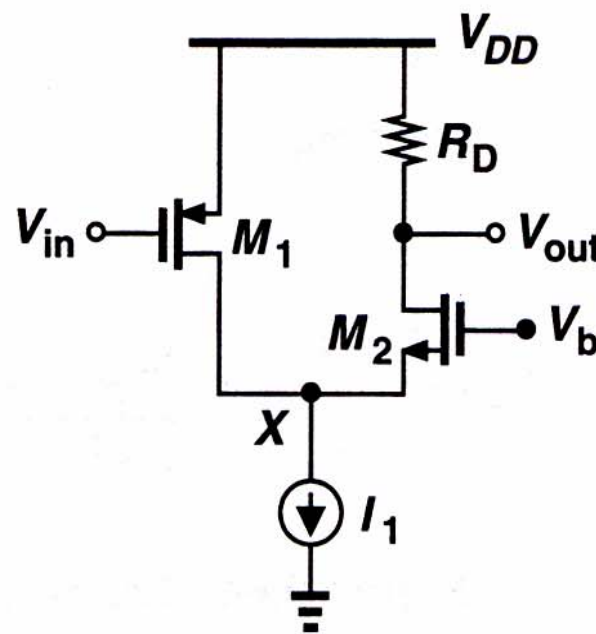
	(a)	(b)	(c)
Linear VOUT	$> V_{DSAT}$	$> 2 V_{DSAT}$	$> 2 V_{DSAT}$
Gm	1	0.5	1
Ro	1	4	gm x ro
Av (Gm x Ro)	1	2	gm x ro

Folded cascode:

- (1) relaxes voltage headroom problem
- (2) Larger current than the regular cascode

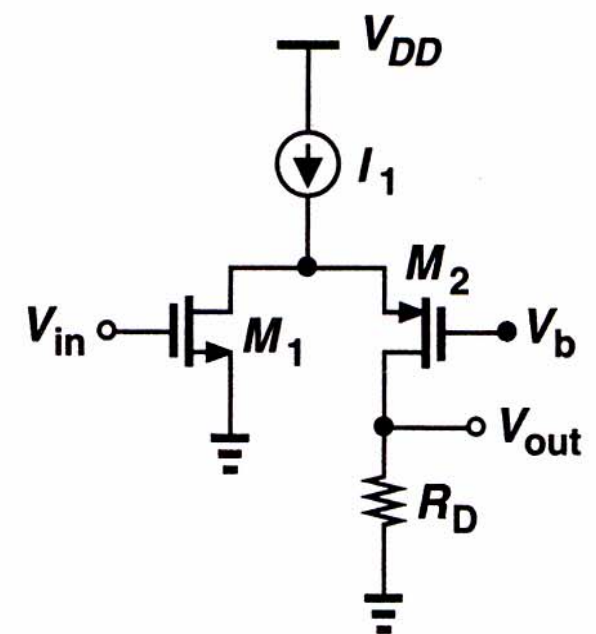


(a)



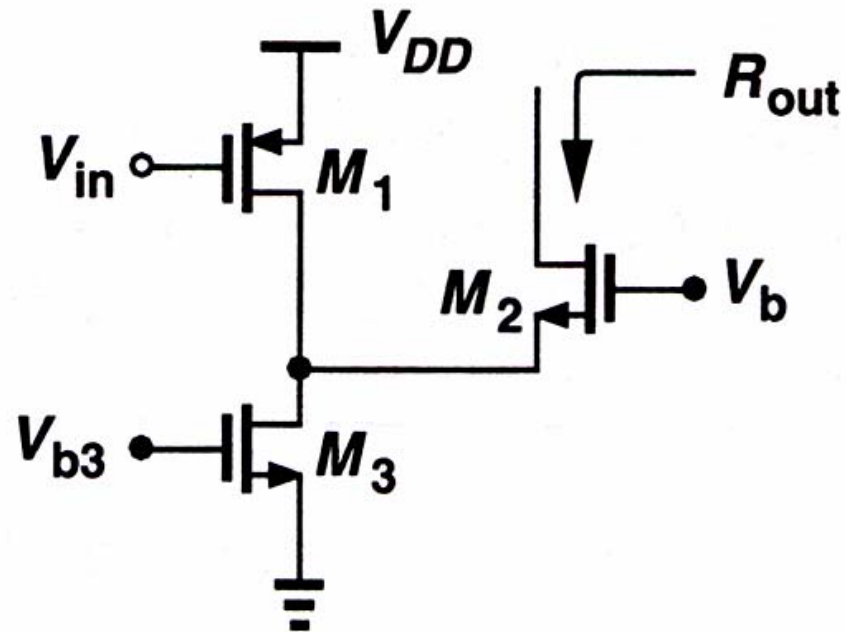
(b)

PMOS + NMOS

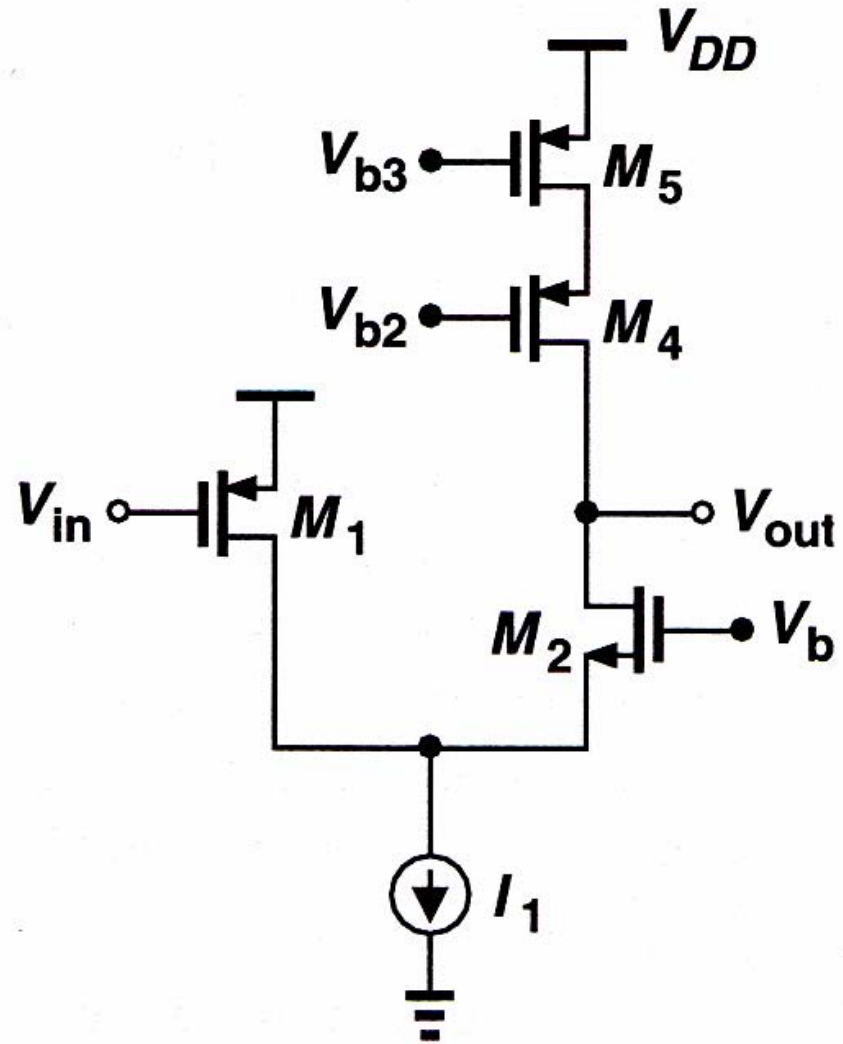


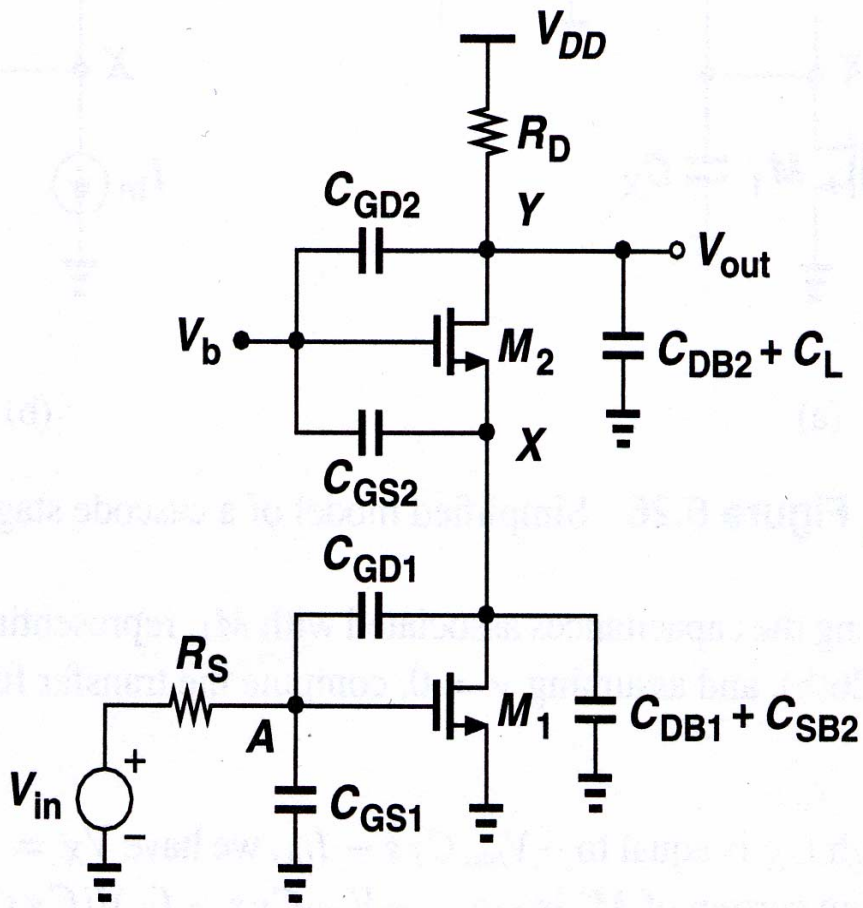
(c)

NMOS + PMOS



$$R_{out} = [1 + (g_{m2} + g_{mb2})r_{O2}](r_{O1} \parallel r_{O3}) + r_{O2}.$$



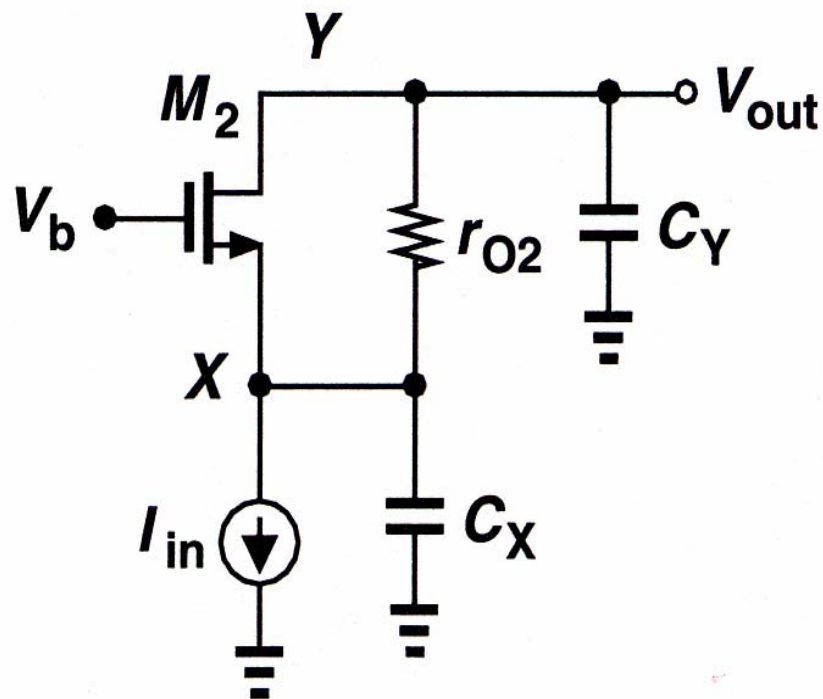


$$\omega_{p,A} = \frac{1}{R_S \left[C_{GS1} + \left(1 + \frac{g_{m1}}{g_{m2} + g_{mb2}} \right) C_{GD1} \right]}$$

$$\omega_{p,X} = \frac{g_{m2} + g_{mb2}}{2C_{GD1} + C_{DB1} + C_{SB2} + C_{GS2}}$$

Not much affected by RD increase

$$\omega_{p,Y} = \frac{1}{R_D (C_{DB2} + C_L + C_{GD2})}$$



$$Z_X = r_{o1} \parallel \frac{1}{sC_X} = \frac{r_{o1}}{1 + s r_{o1} C_X}$$

$$Z_{out} = (1 + g_{m2}r_{o2})Z_X + r_{o2}$$

Z_{out}: capacitive, pole at 1/(r_{o1}*C_x)