

UNIVERSITY OF CALIFORNIA
College of Engineering
Department of Electrical Engineering
and Computer Sciences

Homework #7

EECS 140

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Due 10/15/99 before noon in 497 Cory

Fall 1999

Note: Use the device parameters given in the class handout “Device Parameters & SPICE Models” and also available on the EECS 140 website, <http://kowloon.eecs.berkeley.edu/~courses/140>

Grading: Problem 1: 15% Problem 2: 15% Problem 3: 30%% Problem 4: 40%

- 1) Consider the cascode current mirror shown in Fig.1. Assume equal W/L for all devices.
 - a) Use the small signal model of the circuit to derive an analytical expression for the systematic error $\epsilon = (I_{out}/I_{in}) - 1$ as a function of $\Delta V = (V_{out} - V_{in})$.
 - b) Show that compared to a simple current mirror (M2 and M4 removed), the systematic error is reduced by the factor $1/g_m r_o$.
 - c) Calculate the systematic error for a simple current mirror and for the shown cascode current mirror using $\Delta V = 2V$, $(W/L) = 10\mu m / 2\mu m$, and $I_{out} = 200\mu A$.

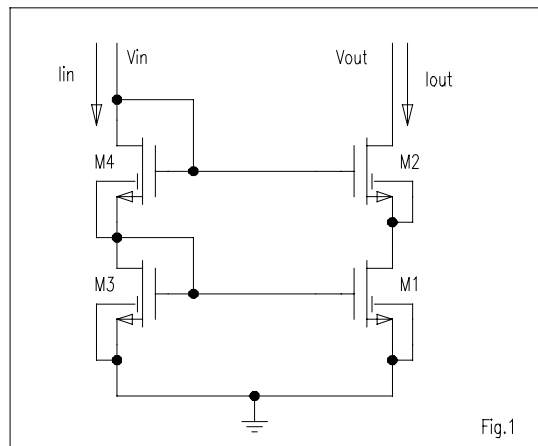
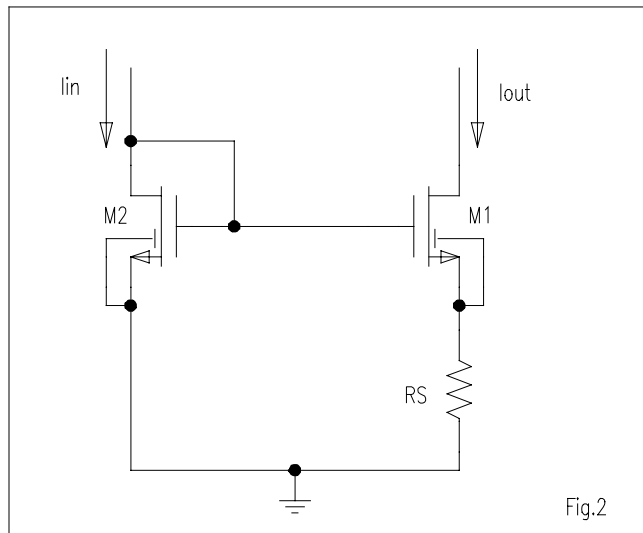
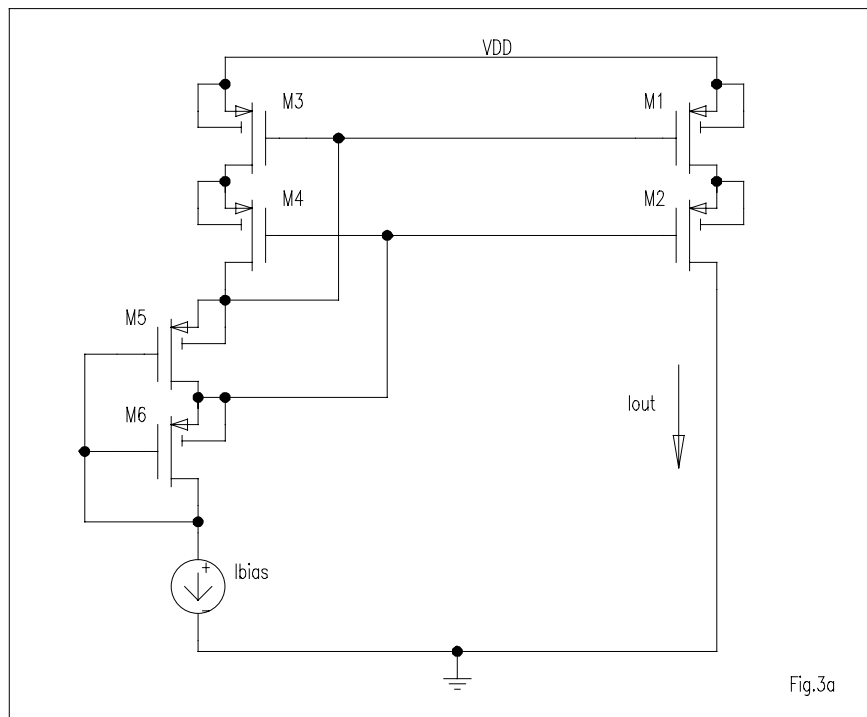


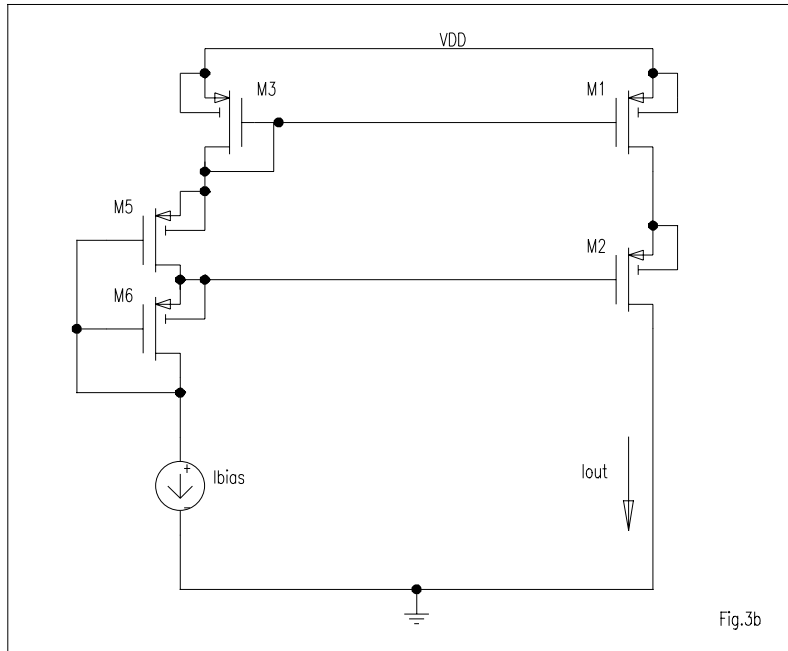
Fig.1

- 2) Consider the current mirror shown in Fig.2. In the layout of the circuit, the source of M2 had accidentally been connected to ground using polysilicon, resulting in a significant source resistance R_S . Use $(W/L) = 100$, $R_S = 100\Omega$ and $I_{in} = 200\mu A$ to calculate ϵ and $\Delta V_{GS} = V_{GS1} - V_{GS2}$. Assume $\lambda = 0$ and $(W/L)_1 = (W/L)_2$. Note how a small ΔV_{GS} can result in a fairly large mismatch in the branch currents!



- 3) Consider the high swing bias circuits shown in Fig. 3a.
- Design the circuit in Fig. 3a, i.e. find all (W/L) for $V_{\min}=400\text{mV}$ and $I_{\text{bias}}=100\mu\text{A}$. Assume $\lambda=0$ for this part of the analysis.
 - Draw the circuit and numerically label all node voltages and branch currents. Explicitly verify the region of operation of M4. Let $V_{\text{DD}}=10\text{V}$.
 - Calculate the systematic error $\epsilon=(I_{\text{out}}/I_{\text{bias}})$ with and without the presence of M4. (After the removal of M4, the circuit reduces to the one shown in Fig. 3b)





- 4) Fig. 4 shows a differential pair, biased through a current source I_{SS} that has a finite output resistance R_x . Assume $\lambda=0$ for the analysis of the circuit.
- Using the circuit's large signal model, find an analytical expression for the small signal differential gain $A_{dm0}=dV_{od}/dV_{id}$ as a function of the applied differential voltage V_{id} . Assume $R_x \rightarrow \infty$ for this part of the analysis.
 - Using your result from (a), calculate A_{dm0} for $V_{id}=0$ and for $V_{id}=100\text{mV}$ using the circuit parameters $(W/L)=50$, $R_D=30\text{k}$ and $I_{SS}=200\mu\text{A}$.
 - Use SPICE to generate a plot of $A_{dm0}=dV_{od}/dV_{id}$ for $V_{id}=-250\text{mV} \dots +250\text{mV}$, using the circuit parameters from part (b). Set $V_{i-}=3\text{V}$ and DC sweep $V_{i+} \pm 250\text{mV}$ with respect to V_{i-} . Use the "Awaves Expression Builder" to generate and display the expression for A_{dm0} . In the same diagram also show the differential output voltage V_{od} .
 - Verify your result from part (b). Mark your calculated value from (b) in the plot. What is the corresponding output voltage for $V_{id}=100\text{mV}$?
 - What are the minimum and maximum small signal gain values as V_{od} swings from -5.5 to $+5.5\text{V}$? Calculate the percent change $[A_{dm0}(\text{max}) - A_{dm0}(\text{min})] / A_{dm0}(\text{max})$. In practice, which gain should be quoted in the circuit's datasheet?
 - Using the circuit's small signal model, find an analytical expression for the common mode gain $A_{cm}=V_{ocm}/V_{icm}$ as a function of the current source resistance R_x . Recall that $V_{i+}=V_{icm}+V_{id}/2$ and $V_{i-}=V_{icm}-V_{id}/2$, $V_{o+}=V_{ocm}+V_{od}/2$, $V_{o-}=V_{ocm}-V_{od}/2$
 - Calculate A_{cm} using the circuit parameters from (b) and $R_x=100\text{k}\Omega$. Repeat for $R_x=10\text{M}\Omega$.

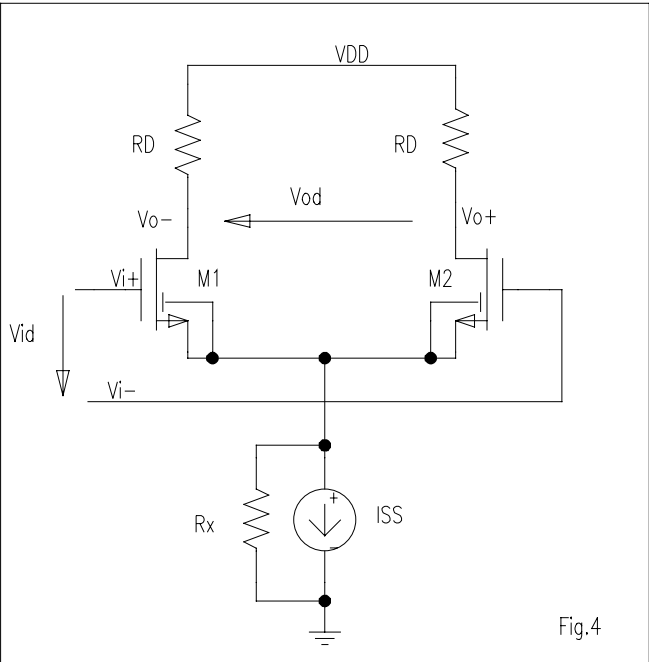


Fig.4