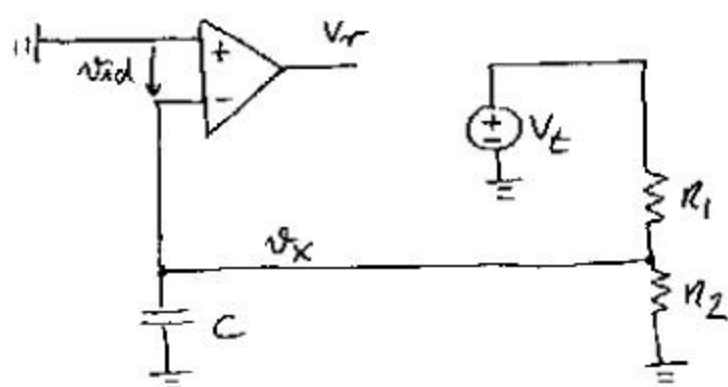


① a)



$$v_x = \frac{R_2}{R_1 + R_2} \frac{1}{1 + s/p_2} \cdot v_r \quad p_2 = -\frac{1}{(R_1 || R_2)C}$$

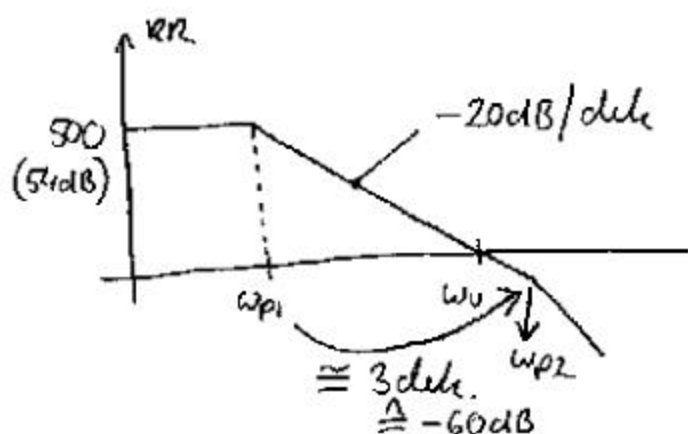
$$v_r = -v_x \frac{a_0}{1 + s/p_1}$$

$$RR = -\frac{v_r}{v_x} = \frac{a_0 R_2}{R_1 + R_2} \cdot \frac{1}{1 + s/p_1} \cdot \frac{1}{1 + s/p_2} //$$

b)

$$1 = \frac{1000 \cdot 30k}{60k} \cdot \left| \frac{1}{1 + s/p_1} \right| \cdot \left| \frac{1}{1 + s/p_2} \right| \quad p_2 = -2\pi \cdot 1.06 \text{ MHz}$$

$$500 = RR_{dc} \quad p_1 = -2\pi \cdot 1 \text{ kHz}$$



$\Rightarrow p_1$ is a dominant pole that \approx sets the f_u of RR

b) ctd.

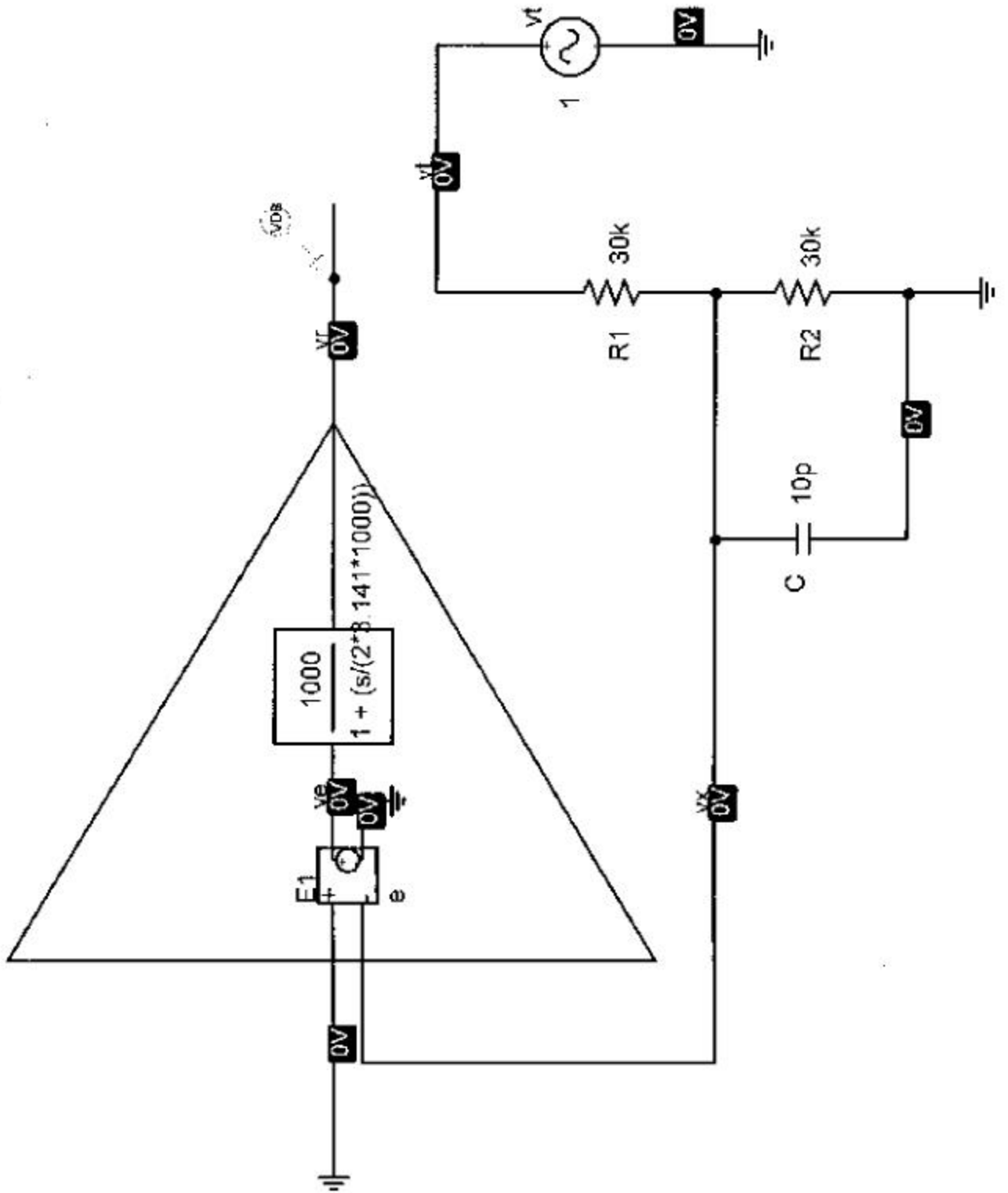
$$\Rightarrow \frac{f_u}{f_{p1}} \approx \frac{500}{1} \Rightarrow \underline{\underline{f_u \approx 500 \text{ kHz}}}$$

$$p_m = \arctan\left(\frac{f_{p2}}{f_u}\right) = \arctan\left(\frac{1060}{500}\right) = \underline{\underline{64^\circ}}$$

c) → see attached schematic, SPICE deck and Plot. Results:

	<u>SIM</u>	<u>CALC</u>	<u>ERROR</u>
f_u	462 kHz	500 kHz	7.6%
p_m	66.7°	64°	4%

→ errors due to dominant pole approximation
(note that you can only make this assumption when the poles are "far enough" apart s.t. a dominant pole basically sets the ω_0 of the RR)



Return-Ratio Simulation for Problem 1

* EECS140 - HW11 - Problem1

(NR Simulation)


* Schematics Netlist *

```
R_R1      vx vt  30k
R_R2      0 vx  30k
V_vt      vt 0 DC 0V AC 1
E_E1      ve 0 0 vx 1
C_C       0 vx  10p

E_LAPLACE1  vr 0 LAPLACE {V(ve)}
+ {(1000)/(1 + (s/(2*3.141*1000)))}

.ac 101 10 10meg
.probe
.END
```

LAPLACE
EXPRESSION

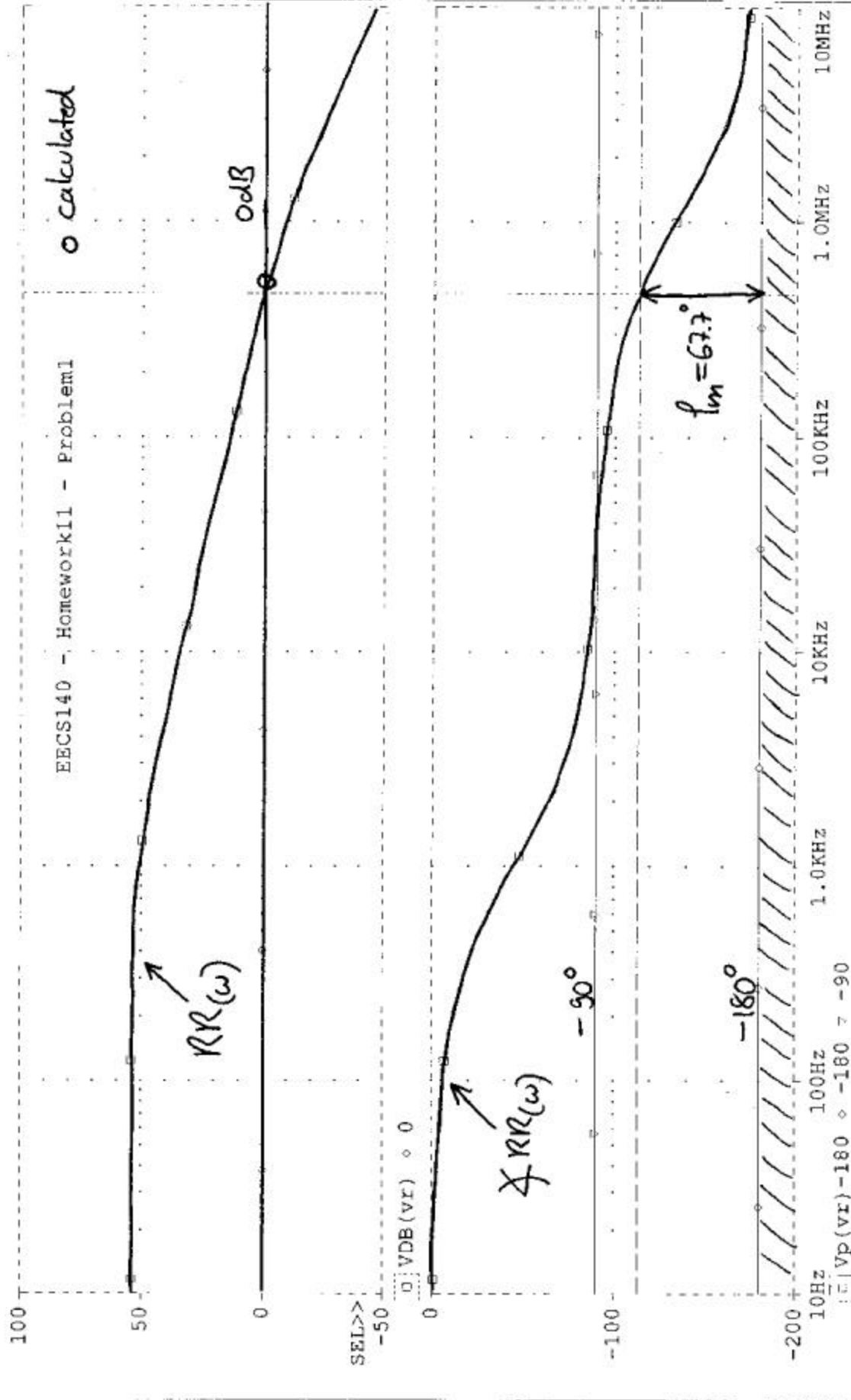


Date/Time run: 11/16/99 13:57:02

* U:\Eel140\spice\hwllsim1.sch

Temperature: 27.0

(A) hwllsim1.dat



A1: (461.725K, -39.520m) A2: (461.725K, -113.344) DIFF(A): (0.000, 113.304)

Date: November 16, 1999

Page 1

Time: 14:00:13

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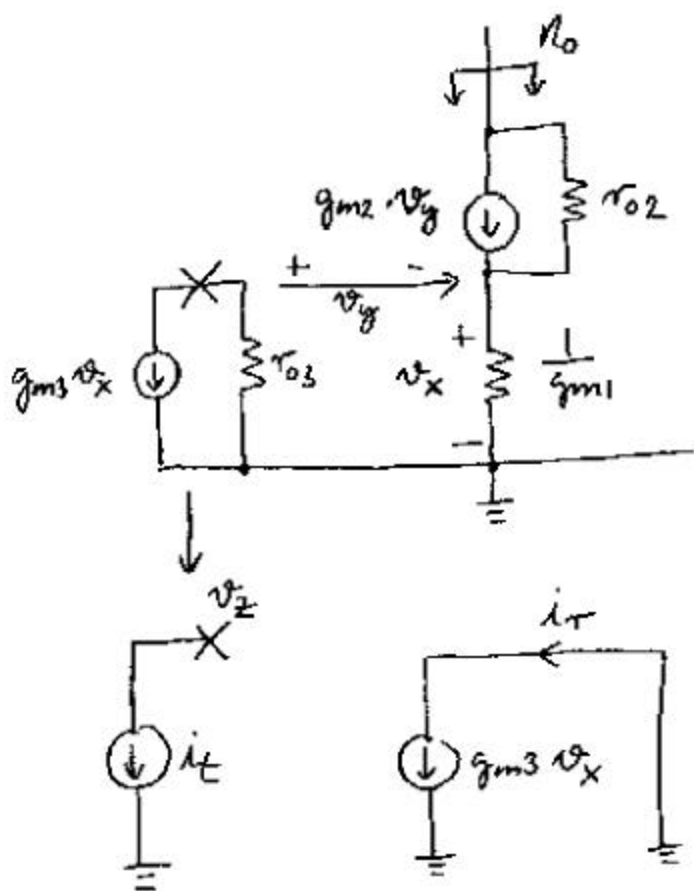
② Bias point: all $I_D = I_{ref}$

$$g_m = \sqrt{200\mu \cdot 200\mu \cdot 50}$$

$$g_m = 1.41 \text{ mS}$$

$$r_o = \frac{1}{0.05 \cdot 100\mu} = 200 \text{ k}\Omega$$

see also EX.
in text. (p. 8T.68)



(i) find $r_o(l_t=0) = r_o(g_{m3}=0)$

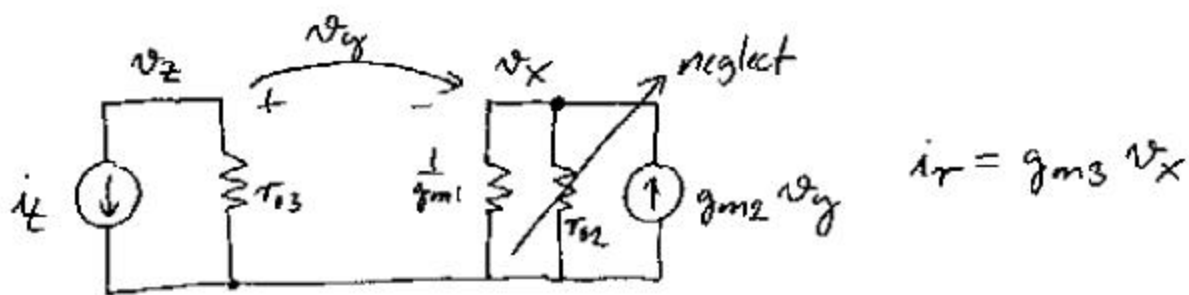
$$r_o(l_t=0) = r_{o2} + \frac{1}{g_{m1}} \approx \underline{\underline{r_{o2}}} = 200 \text{ k}\Omega$$

(ii) find r_r (open port)

→ for an open output port, g_{m2} can only supply current to $r_{o2} \Rightarrow v_x = 0$

$$i_r = g_{m3} v_x \Rightarrow \underline{\underline{r_r = 0}} \text{ (open port)}$$

iii) Find RR (shorted port)



$$v_x = \frac{g_{m2}}{g_{m1}} v_y \quad ; \quad v_y = v_z - v_x$$

$$v_z = -i_t r_{o3}$$

$$\Rightarrow v_x = \frac{g_{m2}}{g_{m1}} v_z - \frac{g_{m2}}{g_{m1}} v_x$$

$$v_x \left(1 + \frac{g_{m2}}{g_{m1}} \right) = -r_{o3} \frac{g_{m2}}{g_{m1}} i_t$$

$$v_x = - \frac{\frac{g_{m2} r_{o3}}{g_{m1}}}{1 + \frac{g_{m2}}{g_{m1}}} \cdot i_t$$

$$v_x = - \frac{g_{m2} r_{o3}}{g_{m1} + g_{m2}} i_t$$

$$\Rightarrow RR = - \frac{i_r}{i_t} = \frac{g_{m2} g_{m3} r_{o3}}{g_{m1} + g_{m2}}$$

$$RR = \frac{1}{2} \cdot 1.41 \text{ mS} \cdot 200 \text{ k} = \underline{\underline{141}} \quad (\text{shorted port})$$

$$\text{iv) } R_{out}(\text{closed loop}) = r_{o1}(i_t=0) \left[\frac{1 + RR(\text{shorted})}{1 + RR(\text{open})} \right]$$

$$= 200 \text{ k} \left[\frac{1 + 141}{1} \right] = \underline{\underline{28.4 \text{ M}\Omega}}$$

3 a)

$$I_{D1} = 10 I_{D2}$$

$$R I_{D2} = \sqrt{\frac{2 \cdot 10 \cdot I_{D2}}{k' \left(\frac{W}{L}\right)_1}} - \sqrt{\frac{2 I_{D2}}{k' \left(\frac{W}{L}\right)_2}} \quad \left(\frac{W}{L}\right)_2 = 10 \left(\frac{W}{L}\right)_1$$

$$\sqrt{I_{D2}} = \frac{1}{R} \left(\sqrt{\frac{20}{k' \left(\frac{W}{L}\right)_1}} - \sqrt{\frac{2}{k' \left(\frac{W}{L}\right)_2}} \right)$$

$$\sqrt{I_{D2}} = \frac{1}{10k} \left(\sqrt{\frac{20}{200\mu \cdot 10}} - \sqrt{\frac{2}{200\mu \cdot 100}} \right)$$

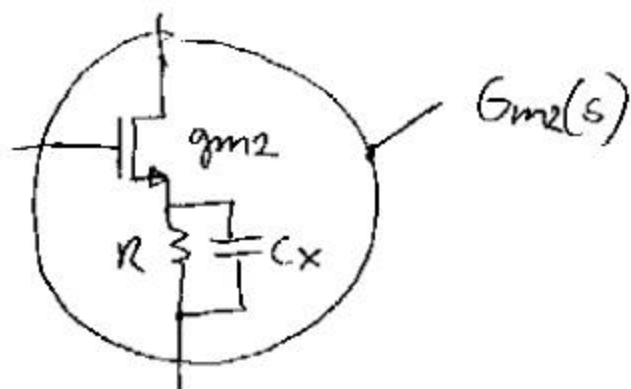
$$\underline{\underline{I_{D2} = 81 \mu A}}$$

$$g_{m1} = g_{m2} = 1.8 \text{ mS}$$

$$g_{m1} = 0.4 \text{ mS}$$

$$g_{m3} = 4 \text{ mS}$$

b)

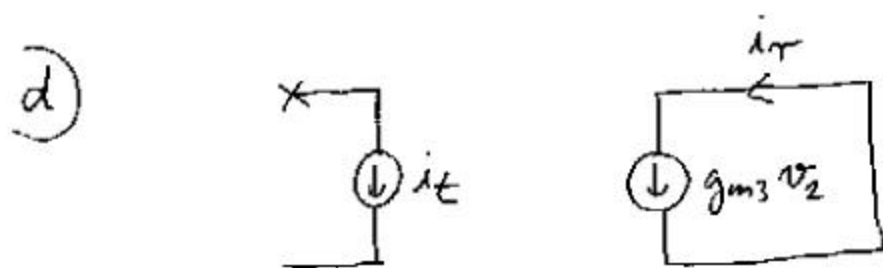
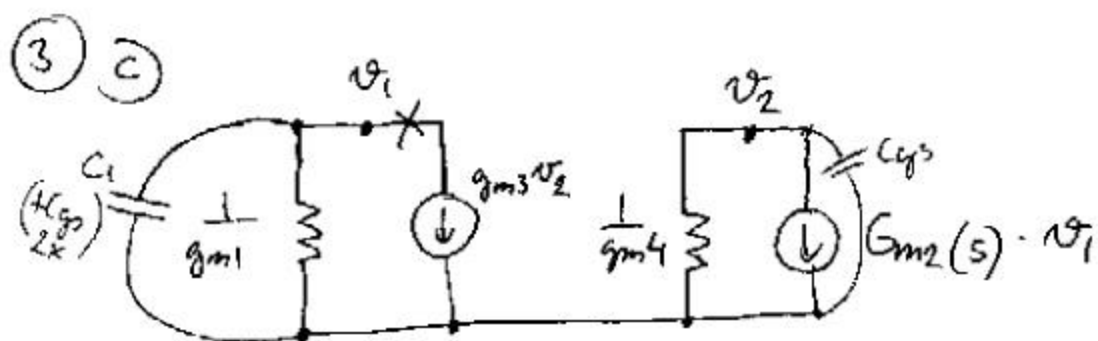


$$G_{m2}(s) = \frac{1}{\frac{1}{g_{m2}} + \left(R \parallel \frac{1}{sC_x}\right)} = \frac{1 + sRC_x}{1 + s \frac{RC_x}{1 + g_{m2}R}} \cdot \frac{1}{\frac{1}{g_{m2}} + R}$$

zero: $z_0 = -\frac{1}{RC_x}$

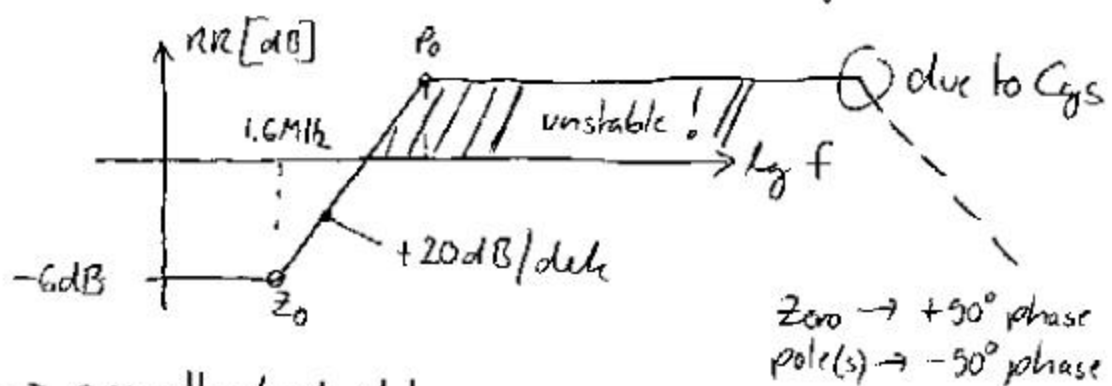
pole: $p_0 = -\frac{1 + g_{m2}R}{RC_x}$ zero-pole doublet

$$p_0 = (1 + g_{m2}R) \cdot z_0$$



$$\begin{aligned} \Rightarrow RR &= \frac{-iR}{iE} = - \frac{gm3}{gm4} \frac{Gm2(s)}{gm1} \\ &= - \frac{gm3}{gm4 gm1} \cdot \frac{1 - s/z_0}{1 - s/p_0} \cdot \frac{1}{\frac{1}{gm2} + R} \\ &= 0.53 \cdot \frac{1 - s/z_0}{1 - s/p_0} \\ &\cong -6dB \end{aligned}$$

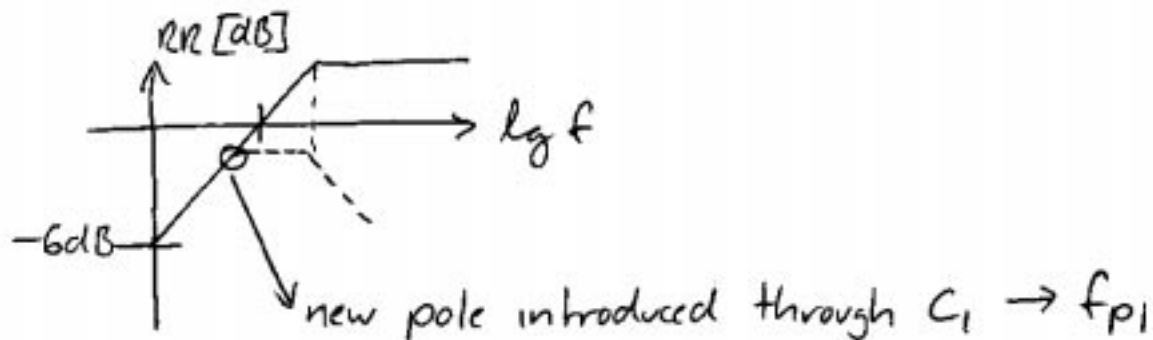
$z_0 = -\frac{1}{RC_x} \Rightarrow f_{z_0} = 1.6MHz$
 $p_0 = 1 + gm2R \Rightarrow f_{p_0} = 30.4MHz$



e) \rightarrow see attached plot

f) \rightarrow see attached plot. Note that the frequency of osc. can not be predicted from the linear analysis. This circuit acts like a "MULTIVIBRATOR". The source of M2 acts like a negative conductance seen by R, Cx. } SEE EE142

g) → need a pole before RR reaches unity:



→ zero at 0.8MHz , -6dB

→ slope is $+6\text{dB/octave} \Rightarrow f_{p1} \leq 1.6\text{MHz}$

$$f_{p1} = \frac{1}{2\pi} \frac{g_{m1}}{C_1}$$

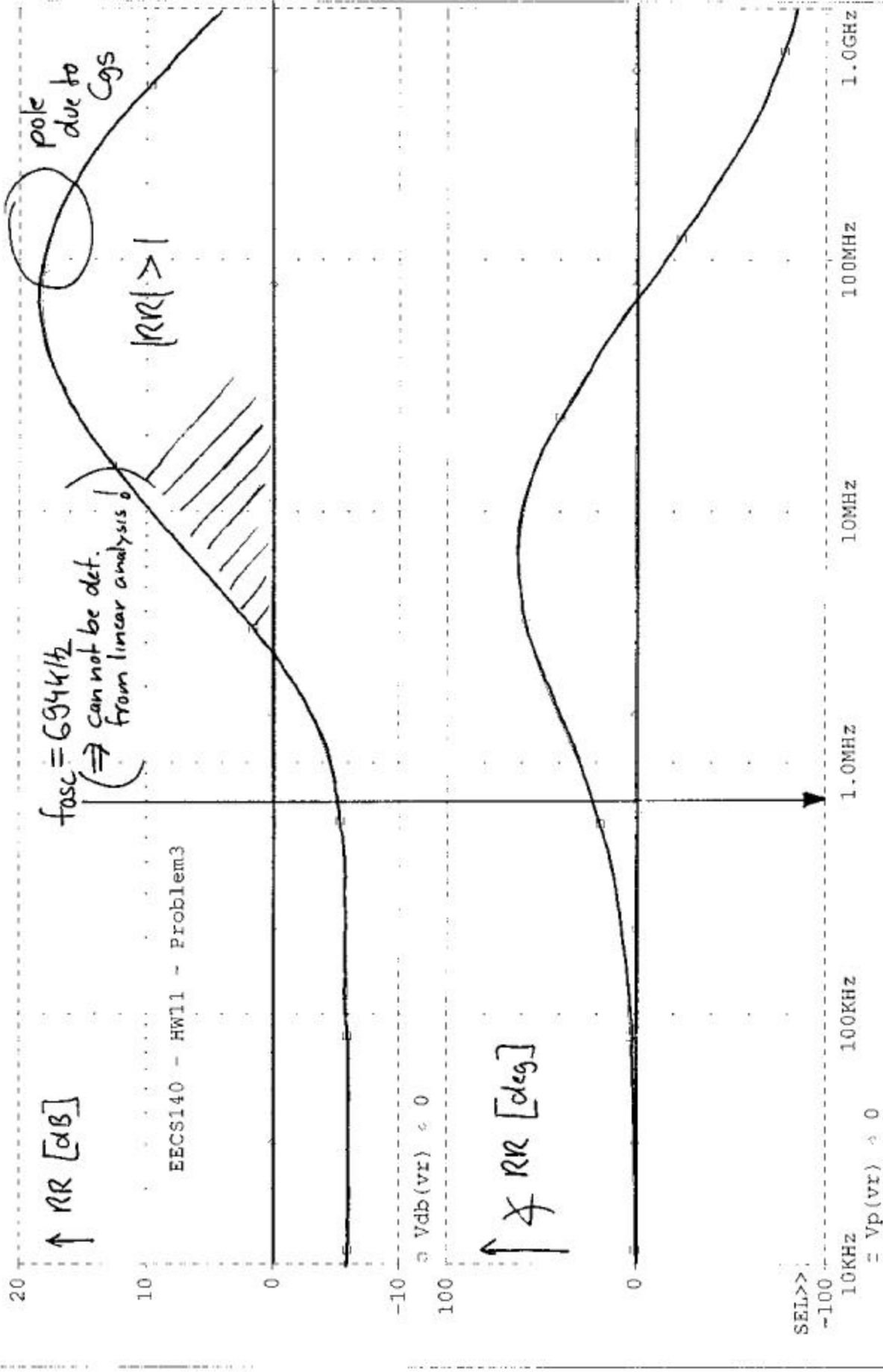
$$C_1 \geq \frac{g_{m1}}{2\pi f_{p1}} = \frac{1.8\text{mS}}{2\pi \cdot 3.2\text{MHz}} = \underline{\underline{179\text{pF}}}$$

→ say $C_1 \cong 200\text{pF}$ or larger

→ not very practical... → can find other ways to compensate

h) → see attached plots

(A) hw11_3ac.dat

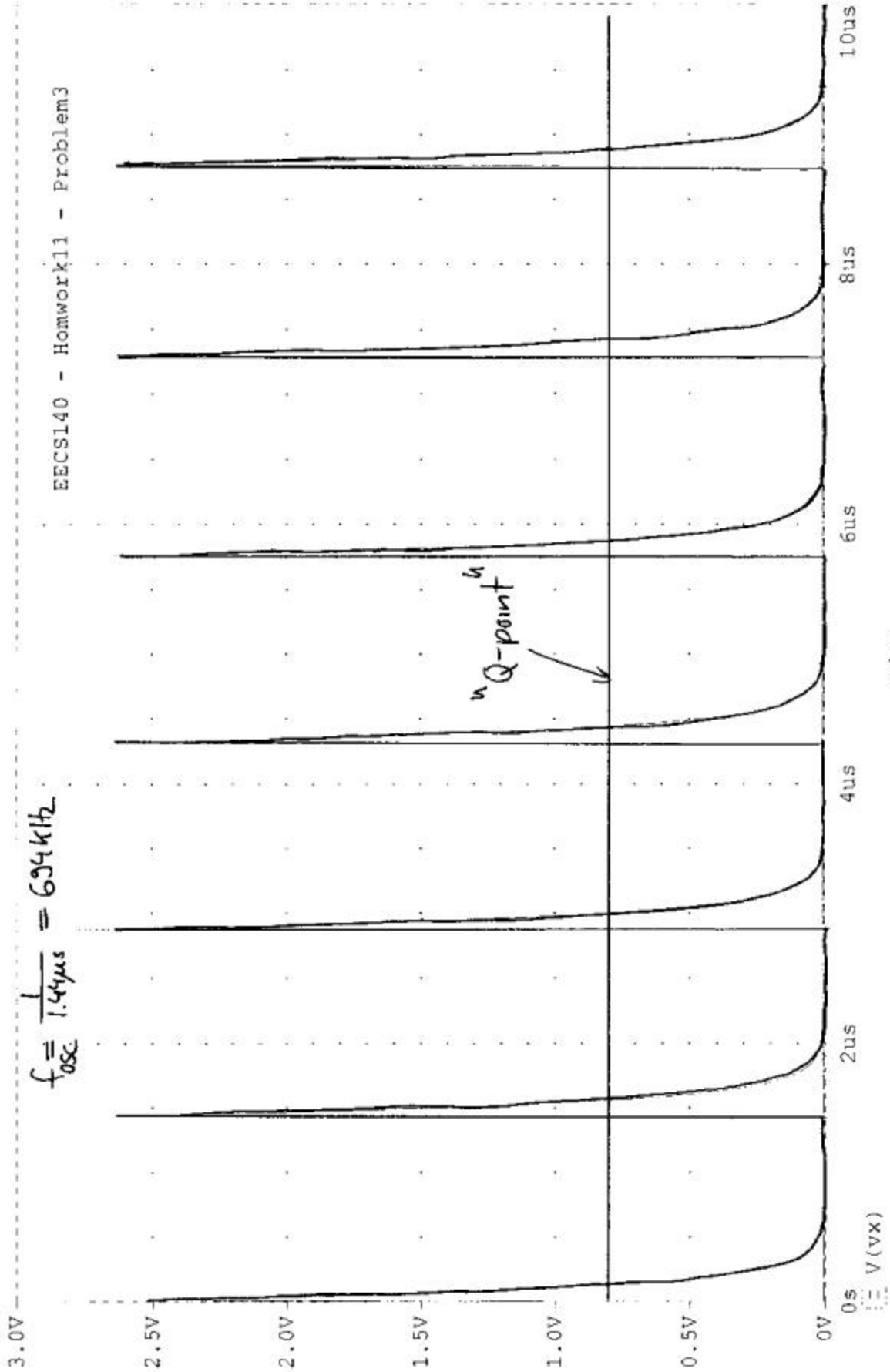


SEL>>
-100
10KHZ
= Vp(vr) < 0
100KHZ
1.0Mhz
10MHz
1.0GHz

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(A) hw11_3tr.dat

$$f_{osc} = \frac{1}{1.4\mu s} = 694 \text{ kHz}$$



Time

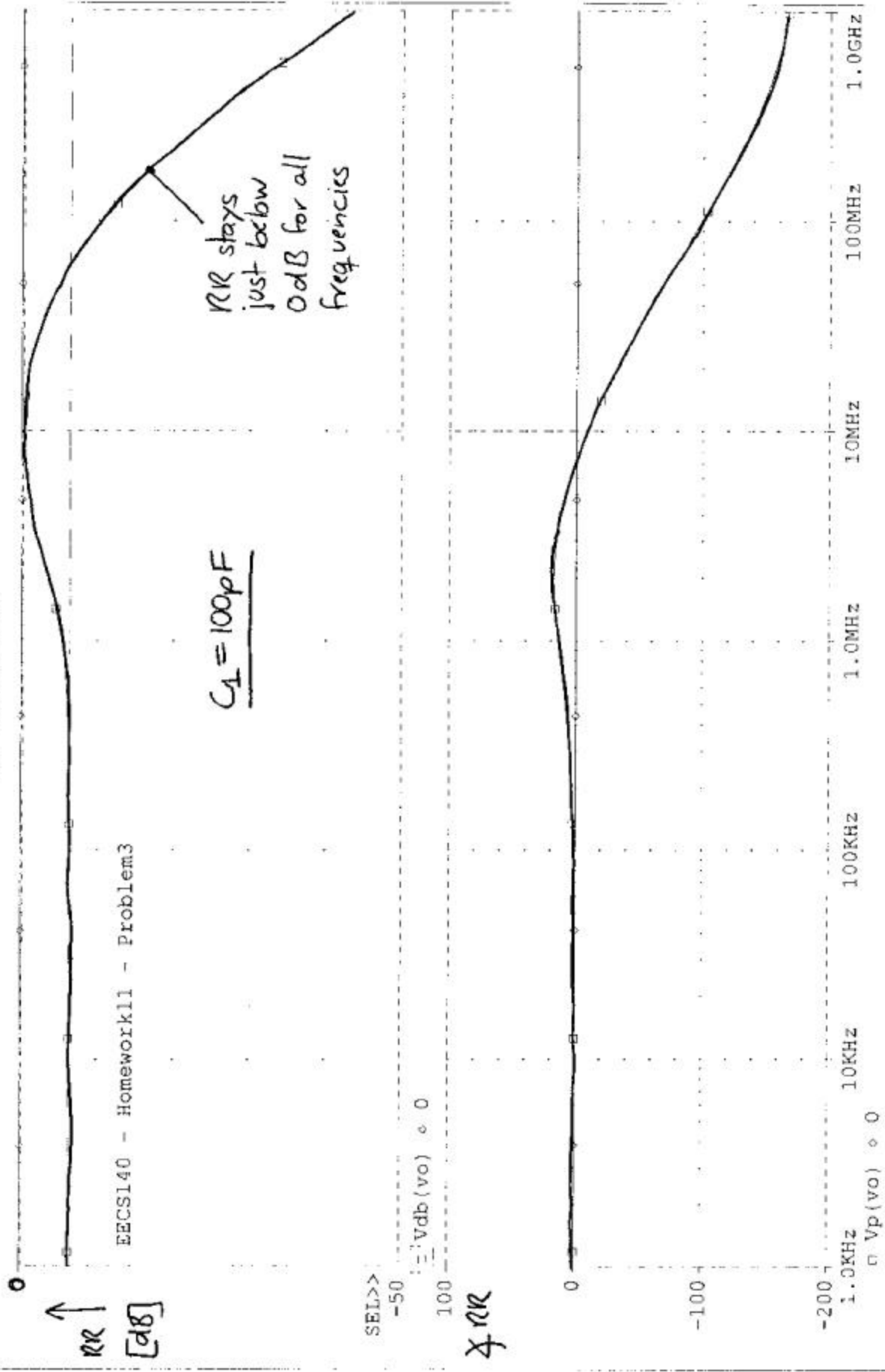
A1:(2.8687u,4.7310m) A2:(4.3094u,4.0941m) DIFF(A):(-1.4407u,636.903u)

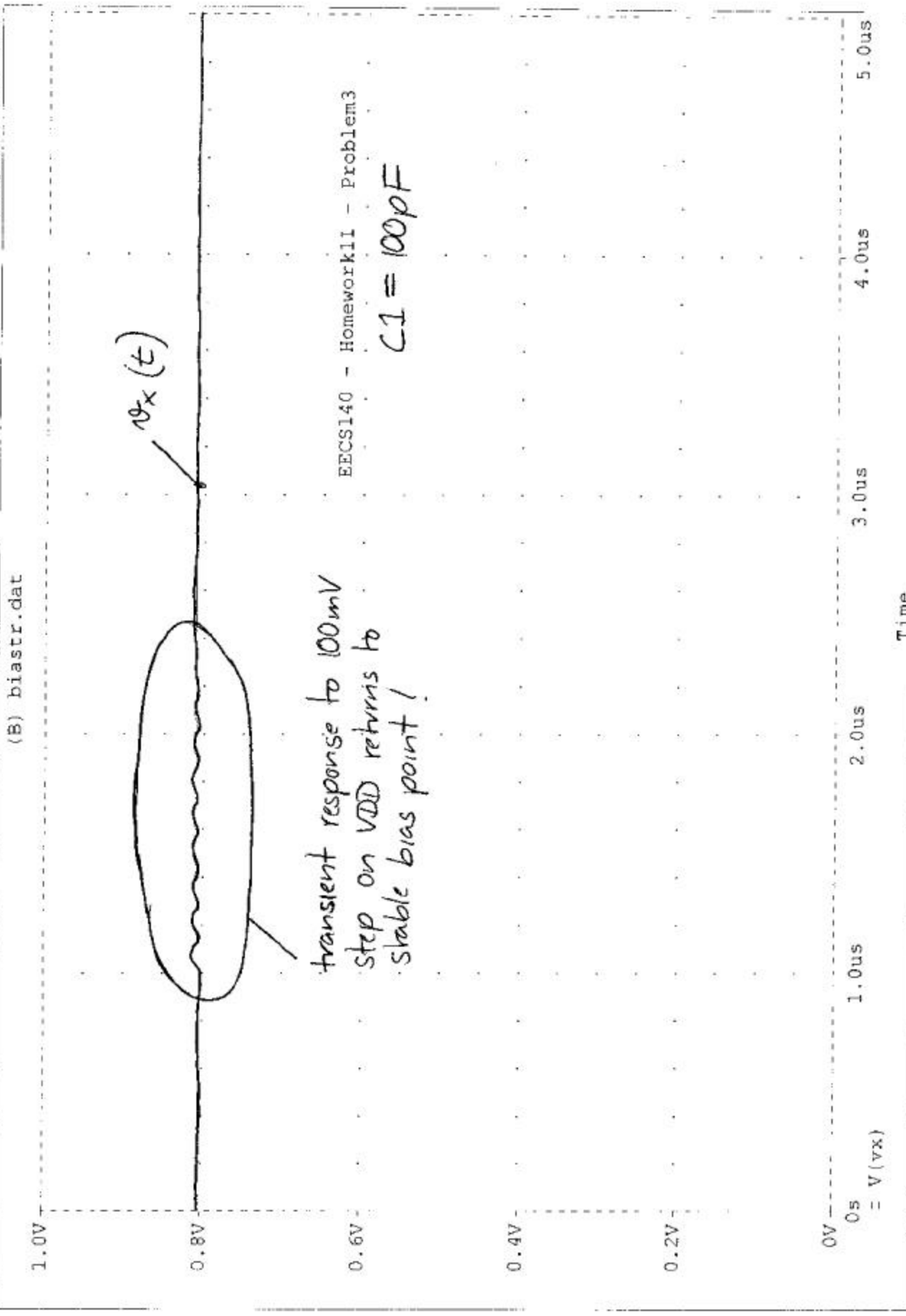
(A) biasac.dat

EECS140 - Homework11 - Problem3

$C_1 = 100\text{pF}$

RR stays just below 0dB for all frequencies





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*** EECS140 Homework11 Problem3
*** This HSPICE deck shows the oscillation

** Analysis setup **

```
.tran lns 10u
.OPTIONS post brief
.OP
.model nmos1 nmos vto=1 tox=6.9n kp=200u
+ lambda=0 gamma=0.5 phi=0.6
.model pmos1 pmos vto=-1 tox=6.9n kp=100u
+ lambda=0 gamma=0.5 phi=0.6
```

```
V_VDD      1 0 5V
R_R1      0 vx 10k
M_M1      v2 v2 0 0 NMOS1
+ L=1u
+ W=10u
M_M2      v3 v2 vx vx NMOS1
+ L=1u
+ W=100u
M_M4      v3 v3 VDD VDD PMOS1
+ L=1u
+ W=10u
V_V15     VDD 1
+PWL 0 0 1u 0 1.001u 100m
M_M3      v2 v3 VDD VDD PMOS1
+ L=1u
+ W=100u
```

*** dummy resistor to startup the circuit...

```
R_R7      v2 v3 100meg
C_Cx      vx 0 10p
```

.END

HELPS SPICE TO
FIND THE CORRECT
OPERATING POINT

